CDC ${ }^{\circ}$ CYBER 170<br>MODELS 175, 740, 750, 760, 865, 875 FUNCTIONAL UNITS

THEORY OF OPERATION
DIAGRAMS

| REMISIOR RECORD |  |
| :---: | :---: |
| REVISION | DESCRIPTION |
| 01 | Preliminary edition. |
| ( $12-74$ ) |  |
| 02 | Updated diagrams to reflect ECO 36000. This edition obsoletes previous edition. |
| (2-75) |  |
| 03 | Updated diagrams to reflect ECOs 35738, 35539, 35693, and 36121. This edition obsoletes pre- |
| (5-75) | vious editions. |
| A | Manual released. This edition obsoletes all previous editions. |
| (7-75) |  |
| B | No change to this manual (ECO 36403). |
| (9-75) |  |
| C | Updated manual to reflect ECO 36429. |
| (9-75) |  |
| D | Updated manual to reflect ECO 36183. |
| (9-75) |  |
| E | Updated manual to reflect ECO 36724. (ECO PD 1346 did not list this manual.) |
| (9-75) |  |
| F | Updated manual to reflect ECO/FCO 36194. |
| (9-75) |  |
| G | Updated manual to reflect ECO/FCO 36699. |
| 10-75) |  |
| H | Updated manual to reflect ECO 36185. |
| ( $11-75$ ) |  |
| J | Updated manual to reflect ECO 36849. |
| ( $12-75$ ) |  |
| K | Updated manual to reflect ECO/FCO 36197. Added secondary block and ESE diagrams. This |
| (2-76) | edition obsoletes all previous editions. |
| L | Manual revised; includes Field Change Order 37147. Pages vi, 5-1-4, 5-1-5 in volume 1, vi, viii, |
| ( $12-76$ ) | $5-11-11,5-11-13,5-11-29,5-11-33,5-11-34.8,5-11-35,5-11-53,5-11-71,5-11-147,5-12-3,5-12-17$, |
|  | $5-12-25,5-12-47,5-12-49,5-12-65$, and $5-14-3$ in volume 2 are revised. |
| M | Manual revised; includes Engineering Change Order 37731. Page 5-1-5 in volume 1 and pages 5-12-3, |
| (8-77) | $5-12-5,5-12-6.1,5-12-6.3,5-12-6.4,5-12-7,5-12-12.2,5-12-13$, and 5-14-1 in volume 2 are revised. |
| N | Manual revised; includes Field Change Order 37843 (ECO 37767). Pages vi, x , 5-1-5, 5-4-5, 5-5-1, |
| (8-77) | $5-7-5,5-7-15,5-10-9$, and 5-10-13 in volume 1 and pages vi, ix, 5-11-5, 5-11-13, 5-11-17, 5-11-19, |
|  | $5-11-29,5-11-33,5-11-35,5-11-93,5-11-153,5-12-1,5-12-3,5-12-5,5-12-7,5-12-9,5-12-10.2$, |
|  | $5-12-11,5-12-13,5-12-15,5-12-17,5-12-25,5-12-43,5-12-63,5-12-65,5-12-69,5-12-71,5-12-78.5$, |
|  | Part 13 divider, 5-13-1, and 5-13-9 in volune 2 are revised. Page 5-12-73.1 is added in volume 2 . |
| Publication No $60420300$ |  |

REVISION LETTERS I, O, Q, S, X AND Z ARE NOT USED.
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Publications and Graphics Division
4201 North Lexington Avenue
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St. Paul, Minnesota 55112
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or use Comment Sheet in the back of this manual.

|  | REVISION RECORD（CONTD） |
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## MANUAL TO EQUIPAENT LEVEL CORRELATION SHEET

This manual reflects the equipment configurations listed below.
EXPLANATION: Locate the equipment type and series number, as shown on the equipment FCO log, in the list below. Immediately to the right of the series number is an FCO number. If that number and all of the numbers underneait it match all of the numbers on the equipmeni FCO log, then this manual accurately reflects the equipment.

| EQUIPMENT TYPE | SERIES | WITH FCOs | COMMENTS |
| :---: | :---: | :---: | :---: |
| AA120 | A01 |  | Released |
|  | A02 | EC035091 |  |
|  | A03 | EC035467 |  |
|  | A04 | EC036623 |  |
|  | A05 | ECO36634 |  |
|  | A06 | EC036637 |  |
|  | A06 | EC036403 |  |
|  | A06 A06 | EC036429 |  |
|  | A06 | EC036724 |  |
|  | A07 | FC036579 |  |
|  | A08 | FCO36699 |  |
|  | A09 | FCO36194 |  |
|  | A09 A09 | ECO36185 ECO36849 |  |
|  | A10 | FCOPD 1407 |  |
|  | All | FCOPD1377 |  |
|  | A12 | FCOPD 1408 |  |
|  | Al3 | FCO36199 |  |
|  | A14 | FCOPD1511 |  |
|  | A16 | FCOPD 1392 |  |
|  | A17 | FCOPD 1454 |  |
|  | A18 | FCOPD1597 |  |
|  | A19 A20 | $\begin{aligned} & \text { FCOPD1562 } \\ & \text { FCO36197 } \end{aligned}$ |  |
|  | A21 | FCOPD938 |  |
|  | A21 | Included |  |
|  | ${ }_{\text {A22 }}$ | FCO36639 |  |
|  | A24 | FCO36624 |  |
|  | A25 | FC036641 |  |
|  | A26 | FC036999 |  |
|  | A27 | FC036646 |  |
|  | A28 | $\begin{aligned} & \text { FCO36645 } \\ & \text { FCO36642 } \end{aligned}$ |  |
|  | A30 | FCO36681 |  |
|  | A31 | FC036644 |  |
|  | A32 | FCO36656 |  |
|  | A33 | FCO36654 |  |
|  | A34 | FC036630 |  |
|  | A35 | FCO37073 |  |
|  | A36 | FCO37054 |  |
|  | A37 | FCO36631 |  |
|  | A39 | FC036651 |  |
|  | A40 | FCO36652 |  |
|  | A41 | FCO36875 |  |
|  | A42 A43 | FCO36626 |  |



| EQUIPMENT TYPE | SERIES | WITH FCOs | COMMENTS |
| :---: | :---: | :---: | :---: |
|  | B01 | FC036630 |  |
|  | B01 | FC037073 |  |
|  | B01 | FCO37054 |  |
|  | B01 | FCO36631 | - |
|  | B01 | FC036653 |  |
|  | B01 | FCO36651 |  |
|  | B01 | FCO36652 |  |
|  | B01 | FC036875 |  |
|  | B01 | FC036626 |  |
|  | B01 | FC036192 |  |
|  | B01 | FCO36854 |  |
|  | B02 | FC037362 |  |
|  | B02 | FCO37530 |  |
|  | B02. | FC037147 |  |
|  | B02 | FC036643 |  |
|  | B05 | EC037731 |  |
|  | B06 | FC037843 |  |
|  | B06 | FC037949 |  |
|  | B07 | FC038031 |  |
|  | B08 | FC038236 |  |
|  | B09 | FC038135 |  |
|  | B10 | FC037840 |  |
|  | B11 | FC038252 |  |
|  | B11 | FC038308 |  |
|  | B12 | FC038171 |  |
|  | B13 | FC038022 |  |
|  | B13 | EC037722 |  |
|  | B14 | FC037813 |  |
|  | B15 | FCO38338 |  |
|  | B16 | FCO38386 |  |
|  | B17 | FC038712 |  |
|  | B18 | FCO38418 |  |
| . | B18 | ECO38980 |  |
|  | B18 | EC038856 |  |
|  | B18 | EC038858 |  |
|  | B18 | EC038897 |  |
|  | B19 | ECO39044 |  |
|  | B20 | EC038584 |  |
|  | B20 | EC038670 |  |
|  | B21 | EC038764 |  |
|  | B22 | EC039041 |  |
|  | B22 | EC038781 |  |
|  | B22 | EC038871 |  |
|  | B23 | EC038941 |  |
|  | B24 | EC039411 |  |
|  | B25 | EC039368 |  |
|  | B25 | EC039319 |  |
|  | B26 | EC038867 |  |
|  | B26 | EC039978 |  |
|  | B26 | EC039735 |  |
|  | B26 | EC040383 |  |
|  | B27 | FC039396 |  |
|  | B27 | FC039678 |  |
|  | B28 | FCO39800/ |  |
|  |  | FC039778 |  |
|  | B29 | $\begin{aligned} & \text { FC039590/ } \\ & \text { FC039773 } \end{aligned}$ |  |
|  | B29 | FC039732 |  |
|  | B30 | FC039726 |  |
|  | B30 | EC040489 |  |
|  | B31 | $\begin{aligned} & \text { FC040748/ } \\ & \text { FC040843 } \end{aligned}$ |  |
|  | B31 | FC040479 |  |
|  | B31 | EC041276 |  |
|  | CO 2 | FC038135 |  |
|  | C02 | FC037840 |  |
|  | C02 | FC038252 |  |


| EQUIPMENT TYPE | SERIES | WITH FCOs | COMMENTS |
| :---: | :---: | :---: | :---: |
|  | C02 | FC038308 |  |
|  | C03 | FC038171 |  |
|  | C03 | FC038022 |  |
|  | C03 | ECO37722 |  |
|  | C03 | FC037813 |  |
|  | C04 | FC038338 |  |
|  | C04 | FC038386 |  |
|  | C05 | FC038712 |  |
|  | - C 06 | FCO38418 |  |
|  | C06 | EC038980 |  |
|  | C07 | EC038856 |  |
|  | C08 | EC038858 |  |
|  | C08 | EC038897 |  |
|  | C09 | EC039044 |  |
|  | C10 | EC038584 |  |
|  | C11 | ECO38670 |  |
|  | C11 | ECO38764 |  |
|  | C12 | EC039041 |  |
|  | C13 | EC038781 |  |
|  | C14 | EC038871 |  |
|  | C15 | EC038941 |  |
|  | C16 | EC039411 |  |
|  | C17 | EC039368 |  |
|  | C18 | EC039319 |  |
|  | C18 | EC038867 |  |
|  | C18 | EC039978 |  |
|  | C18 | EC039735 |  |
|  | C18 | EC040383 |  |
|  | C19 | FC039396 |  |
|  | C19 | FC039678 |  |
|  | C19 | $\begin{aligned} & \text { FC039800/ } \\ & \text { FC039778 } \end{aligned}$ |  |
|  | C20 | FC039590/ |  |
|  |  | FC039773 |  |
|  | C20 | FC039732 |  |
|  | C21. | FC039726 |  |
|  | C21 | EC040489 |  |
|  | C21 | $\begin{aligned} & \text { FC040748/ } \\ & \text { FC040843 } \end{aligned}$ |  |
|  | C22 | FC040479 |  |
|  | C22 | EC041276 |  |
|  | D01 | EC039978 |  |
|  | D01 | ECO39735 |  |
|  | D01 | EC040383 |  |
|  | D01 | FCO39396 |  |
|  | D01 | FC039678 |  |
|  | D01 | FC039800/ |  |
|  |  | FCO39778 |  |
|  | D01 | $\begin{aligned} & \text { FCO39590/ } \\ & \text { FC039773 } \end{aligned}$ |  |
|  | D01 | FCO39732 |  |
|  | D01 | FCO39726 |  |
|  | D01 | EC040489 |  |
|  | D01 | $\begin{aligned} & \text { FC040748/ } \\ & \text { FC040843 } \end{aligned}$ |  |
|  | D01 | FC040479 |  |
|  | D01 | EC041276 |  |
|  | D01 |  | Released |
|  | D02 | FC039944 |  |
|  | D03 | FCO40112 |  |
|  | D04 | FC040106 |  |
|  | D05 | $\begin{aligned} & \text { FC039936/ } \\ & \text { FC040225 } \end{aligned}$ |  |
|  | D06 | FC040205 |  |
|  | D07 | FC040221 |  |
|  | D08 | FC040432 |  |


| EQUIPMENT TYPE | SERIES | WITH FCOs | COMMENTS |
| :---: | :---: | :---: | :---: |
| AT364 | D09 | EC039947/ |  |
|  | D10 | FC040364 |  |
|  | D11 | FC039945 |  |
|  | D12 | FC040607 |  |
|  | D13 | FC040658 |  |
|  | D14 | FC040030 |  |
|  | D15 | FC040583 |  |
|  | D16 | FC040589A |  |
|  | D17 | FC040740 |  |
|  | D18 | FC040857 |  |
|  | D19 | FC040830 |  |
|  | D20 | FC040827 |  |
|  | D21 | FC040915 |  |
|  | D22 | FC040904 |  |
|  | D23 | FC04 1012 |  |
|  | D24 | FC04 1021 |  |
|  | D25 | FC04 1043 |  |
|  | D26 | EC041358 |  |
|  | D27 | FC041346 |  |
|  | D28 | EC04 1363 |  |
|  | D29 | FC041359 |  |
|  | D30 | FC041833 |  |
|  | D31 | EC041632 |  |
|  | D32 | EC041639 |  |
|  | D33 | FC041636 |  |
|  | D34 | EC041790 |  |
|  | D35 | FC042000 |  |
|  | D36 | EC041372 |  |
|  | D37 | FC042050 |  |
|  | D38 | FC041736 |  |
|  | D39 | FCO42323 |  |
|  | D40 | FCO42348 |  |
|  | D41 | FCO42913 |  |
|  | E01 | EC044495 |  |
|  | A01 | ECO38980 |  |
|  | A01 | EC038856 |  |
|  | A01 | EC038858 |  |
|  | A01 | EC038897 |  |
|  | A01 | EC039044 |  |
|  | A01 | EC038584 |  |
|  | A01 | EC038670 |  |
|  | A01 | EC038764 |  |
|  | A01 | EC039041 |  |
|  | A01 | EC038781 |  |
|  | A01 | EC038871 |  |
|  | A01 | EC038941 |  |
|  | A01 | EC039411 |  |
|  | A01 | EC039368 |  |
|  | A01 | EC039319 |  |
|  | A01 | EC038867 |  |
|  | A01 | EC039978 |  |
|  | A01 | EC039735 |  |
|  | A01 | EC040383 |  |
|  | A01 | FC039396 |  |
|  | A01 | FC039678 |  |
|  | A01 | $\begin{aligned} & \text { FC039800/ } \\ & \text { FC039778 } \end{aligned}$ |  |
|  | A01 | $\begin{aligned} & \text { FC039590/ } \\ & \text { FC039773 } \end{aligned}$ |  |
|  | A01 | FC039732 |  |
|  | A01 | FC039726 |  |
|  | A01 | EC040489 |  |
|  | A01 | $\begin{aligned} & \text { FC040748/ } \\ & \text { FC040843 } \end{aligned}$ |  |
|  | A01 | FC040479 |  |


| EQUIPMENT TYPE | SERIES | WITH FCOs | COMMENTS |
| :---: | :---: | :---: | :---: |
|  | A01 | EC041276 | - |
|  | A01 |  | Released |
|  | A01 | FC039944 |  |
|  | A01 | FC040112 |  |
|  | A01 | FC040106 |  |
|  | A01 | FC039936/ |  |
|  | A01 | FC040205 |  |
|  | A01 | FC040221 |  |
|  | A01 | FC040432 |  |
|  | A01 | $\begin{aligned} & \text { ECO39947/ } \\ & \text { FCO40364 } \end{aligned}$ |  |
|  | A01 | FC040379 |  |
|  | A01 | FC039945 |  |
|  | A01 | FC040607 |  |
|  | A01 | FC040658 |  |
|  | A01 | FC040030 |  |
|  | A01 | F6040583 |  |
|  | A01 | FC040589A |  |
|  | A01 | FC040740 |  |
|  | A01 | FC040857 |  |
|  | A01 | FC040830 |  |
|  | A01 | FC040827 |  |
|  | A01 | FC040915 |  |
|  | A01 | FC040904 |  |
|  | A01 | FC041012 |  |
|  | A01 | FC041021 |  |
|  | A01 | FC041043 |  |
|  | A01 | EC041358 |  |
|  | A01 | FC041346 |  |
|  | A01 | EC041363 |  |
|  | A01 | FC041359 |  |
|  | A01 | FC041833 |  |
|  | A01 | EC041632 |  |
|  | A01 | EC041639 |  |
|  | A01 | FC041636 |  |
|  | A01 | EC041790 |  |
|  | A01 | FC042000 |  |
|  | A01 | EC041372 |  |
|  | A01 | FC042050 |  |
|  | A01 | FCO41736 |  |
|  | A01 | FC042323 |  |
|  | A01 | FCO42348 |  |
|  | A01 | FC042913 |  |
|  | B01 | EC039978 |  |
|  | B01 | EC039735 |  |
|  | B01 | EC040383 |  |
|  | B01 | FC039396 |  |
|  | B01 | FCO39678 |  |
|  | B01 | $\begin{aligned} & \text { FCO39800/ } \\ & \text { FC039778 } \end{aligned}$ |  |
|  | B01 | $\begin{aligned} & \text { FC039590/ } \\ & \text { FC039773 } \end{aligned}$ |  |
|  | B01 | FCO39732 |  |
|  | B01 | FCO39726 |  |
| . | B01 | EC040489 |  |
|  | B01 | $\begin{aligned} & \text { FC040748/ } \\ & \text { FC040843 } \end{aligned}$ |  |
|  | B01 | FC040479 |  |
|  | R01 B01 | ECO41276 | Released |
|  | B01 | FC039944 |  |
|  | B01 | FC040112 |  |
|  | B01 | FC040106 |  |
|  | B01 | $\begin{aligned} & \text { FC039936/ } \\ & \text { FC040225 } \end{aligned}$ |  |
|  | B01 | FC040205 |  |



This manual contains theory of operation and diagrams for the functional unit portion of the CONTROL DATA ${ }^{(1)}$ AA120-A/B/C/D/E Central Computer and the AAl31-D Central Computer with an AT402-A Upgrade Option installed. Also included is information about the AT364-A/B Central Processor Enhancement which is used with the AA120-C/D Central Computer.

Models 740, 750, and 760 are defined as 7X0. Models 865 and 875 are defined as $8 \times 5$.
Comparable information on the central processing unit (CPU), central memory control (CMC), central storage unit (CSU), peripheral processor subsystem (PPS), and extended core storage (ECS) coupler is contained in other manuals. The system publication indexes on the following pages graphically list the related publications. Refer to the Literature and Distribution Services catalog for the latest revision of each manual.

This manual contains a manual to equipment level correlation sheet. The last line of this sheet indicates the latest equipment level (series code) that the manual covers. All manuals for the central computer indicate the latest series code even if the particular manual is not affected by the field change order.

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## SYSTEM PUBLICATION INDEX



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## general description

SECTION 2
OPERATION
(Information for sections 1 and 2 is contained in the CDC CYBER 170 Hardware Reference Manual.)

SECTION 3

INSTALLATION AND CHECKOUT

# (Information for section 3 is contained in the CDC CYBER 170 Installation and Checkout Manual.) 

SECTION 4

THEORY OF OPERATION
(Information for section 4 is combined with the diagrams in section 5 of this manual.)


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## GENERAL DESCRIPTION

## SECTION 2

OPERATION


#### Abstract

(Information for sections 1 and 2 is contained in the CDC CYBER 170 Hardware Reference Manual, publication number 60420000.)


## SECTION 3

 INSTALLATION AND CHECKOUT(Information for section 3 is contained in the CDC CYBER 170 Model 175 Installation and Checkout Manual, publication number 60420500.)

## SECTION 4

THEORY OF OPERATION
(Information for section 4 is combined with the diagrams in section 5 of this manual.)

## PART 1

INTRODUCTION

## GLOSSARY

The glossary is a list of terms, mnemonics, and abbreviations used on the diagrams.

## MODULE-TO-DIAGRAMS CROSS-REFERENCES

GENERAL

Table 5-1 provides a list of functional unit module types along with the following information on each type.

| Quantity | Number of modules of this type located on the mainframe |
| :--- | :--- |
| Location | All physical locations at which this module type can be found |
| Diagram | Secondary block and detailed-modules diagrams on which this module type <br> is depicted. |

USES

Table 5-1 can be used to locate a particular module logic diagram in the manual without knowing the functional entity to which it pertains. Another use is to determine the availability and the location of a substitute module during maintenance. The table also helps to locate all modules of the same type without having to scan all module placement diagrams.

KEY TO SYMBOLS (Figures 5-1 and 5-2)
The number of unfamiliar symbols on the block and detailed-modules diagrams has been kept to a minimum. The symbology has been chosen because it simplifies or clarifies and is therefore essential to the use and understanding of the diagrams. Time spent familiarizing oneself with these conventions is not wasted. Note particularly that the AND and OR symbols define functions, not gates. In some cases, hardware constraints caused the use of AND gates to perform OR functions. Therefore, the block and detailed-modules diagrams depict the $O R$ function, not the AND hardware.

## SYSTEM BLOCK AND INTERFACE DIAGRAMS

The system block diagrams (figures 5-3, 5-4, 5-5) provide a comprehensive view of the central computer. In addition to aiding one's understanding of the system, it relates physical to electrical data (for example, multiply functional unit located on chassis 7) and defines the boundaries of the central processor diagrams in the various parts of this section.

The CP interface diagrams (figures 5-6, 5-7) depict all data and control paths between the functional entities. Also, this diagram includes all paths between the CP and the other portions of the central computer which are covered in other manuals.

The status and control register interface diagram (figure 5-8) includes all status/control paths in the central computer.

## DIAGRAMS

Within each functional entity, the diagrams are arranged in the following order: primary, secondary, detailed-modules, and module logic. The diagrams are intended to ease the trainee gradually but quickly into the central processor or to allow the initiated to choose from four levels of drawings for maintenance assistance or memory jogging. The primary block diagram shows the basic relationship between the various logical components (registers, adders, and so on) comprising the functional entity. The next, more detailed level is the secondary block diagram. It depicts the physical building blocks (modules) and shows the data and control paths between them. The third and most explicit level is the detailed-modules diagram. This level is similar to the secondary level except that it also features a simplified block diagram of each unique module. From this level, the user must proceed to the module logic diagrams only for test point and connector information. However, when it is necessary to do so, the user arrives at the module logic diagram already familiar with its contents. Supplemental timing and troubleshooting information are provided at the end of each part of this section.

A test point chart is provided with each detailed-modules diagram. These charts list test points associated with signals and registers. Translator test points are not listed. An asterisk adjacent to the module type indicates that additional test points are available but not listed. Unless otherwise specified, all test points display the complement of the data indicated in the charts.

Maintenance personnel may find it helpful to reposition the pages in this section so the diagrams are not separated by theory of operation pages. Although the present arrangement is best for training purposes, contiguous diagrams may be easier to use during maintenance.

## MODULE PLACEMENT DIAGRAMS

The module placement diagrams for all CP chassis (5, 6, and 7) are included in the CPU/CMC manual.


any number of connections of either type may be made to the same bracket. brackets are nested when connections to all modules in the stack are not ioentical.
arackets along top or bottom of stack are equivalent to brackets adjacent to all modules in stack
individual inputs or outputs (not bracketed) may enter or leave individual modules in stack through gracket.

Figure 5-1. Key to Diagram Symbols


Figure 5-2. Key to Logic Symbols


Figure 5－3．System Block Diagram，Model 175






Figure 5-7. Interface Diagram, Central Processor, Models 740, 750, 760


Figure 5-8. Interface Diagram, Status/Control Register, Model 175

PART 2

BOOLEAN UNIT



PART 2

BOOLEAN UNIT

The boolean unit executes the CPU instructions requiring bit-by-bit data manipulation. This includes both the logical operations and the transmissive operations. The instructions providing logical operations are:

| 11 | Logical product of Xj and Xk to Xi |
| :--- | :--- |
| 12 | Logical sum of Xj and Xk to Xi |
| 13 | Logical difference of Xj and Xk to Xi |
| 15 | Logical product of Xj and $\overline{\mathrm{Xk}}$ to Xi |
| 16 | Logical sum of Xj and $\overline{\mathrm{Xk}}$ to Xi |
| 17 | Logical difference of Xj and $\overline{\mathrm{Xk}}$ to Xi |

The instructions providing transmissive operations are:

```
Transmit Xj to Xi
Transmit \overline{Xk}}\mathrm{ to Xi
Unpack Xk to Bj and Xi
Pack Xk and Bj to Xi
```


## INPUT REGISTERS

Three data input registers exist for the $B j, X j$, and $X k$ operands. These registers are cleared and entered with new data each clock period. The contents of the $\mathrm{Bj}, \mathrm{Xj}$, and Xk registers are transmitted to the boolean unit each clock period, without regard to the instruction in the CIW register. These operands are then available in the boolean unit in the following clock period.

## CONTROL

Several bits of control information enter the boolean unit. Instruction designators f bit 0 and m bits 0 through 2 are sent to the boolean unit from the CIW register each clock period. These bits define the particular boolean instruction being executed. The instruction control translator decodes the bits to determine the type of logical operation and to select the data paths through the boolean unit for the various instructions. The control signals generated by the instruction
control translator are:

| Extend X: | 26 - Xk bit 59 |
| :--- | :--- |
| Gate Xk upper: | $12+13+14+16+17$ |
| Comp Xk: | $14+15+17+17$ |
| Gate Xj: | $10+12+13+16+17$ |
| Gate LP: | $11+12+15+16$ |
| Gate B: | 27 |
| Comp B: | $27+\mathrm{Xk}$ bit 59 |
| Gate Xk lower: | $12+13+14+16+17+26+27$ |

Go boolean is received by the boolean unit during the clock period following issue of a boolean instruction. Data is transmitted to the destination registers only during a clock period in which go boolean is set.

## OPERATION

If go boolean is set during a given clock period, a boolean instruction issued during the previous clock period and the data in the boolean unit input registers corresponds with the data described by the $j$ and $k$ designators in that instruction. Data in the input registers is merged in a static network for transmission to the destination $B$ and/or $X$ registers. The type of logical operation and the selection of data paths in this static network are determined by the control information.

## LOGICAL PRODUCT

The boolean unit forms the logical product (AND function) of 60 -bit words from the Xj and Xk registers (or its complement) and places the product in the Xi register. The operation is controlled by the gate LP signal. Bits in Xi are set to one when the corresponding bits in Xj and Xk (or its complement) are one as in the following examples.

$$
\begin{array}{ll}
X j=0101 & X j=0101 \\
X k=1100 & X k=0011 \\
X i=0100 & X i=0001
\end{array}
$$

The boolean unit forms the complemented result, which is recomplemented so that the true result is transmitted to the $X$ register.

## LOGICAL SUM

The boolean unit forms the logical sum (OR function) of 60 -bit words from the Xj and Xk registers (or its complement) and places the logical sum in the Xi register. This operation is performed in two steps and is controlled by the gate LP, gate Xk upper, gate Xk lower, and gate Xj signals.

Bits in Xi are set to one if the corresponding bits in Xj and Xk (or its complement) are a one as in the following examples.

$$
\begin{array}{ll}
X_{j}=0101 & X j=0101 \\
X k=1100 & X k=0011 \\
X i=1101 & X i=0111
\end{array}
$$

The boolean unit forms the complemented result, which is recomplemented so that the true result is transmitted to the X register.

## LOGICAL DIFFERENCE

The boolean unit forms the logical difference (exclusive OR function) of the quantity from the Xj and Xk registers (or its complement) and places the difference in the Xi register. This operation is controlled by the gate Xk upper, gate Xk lower, and gate Xj signals.

Bits in Xi are set to one if the corresponding bits in Xj and Xk (or its complement) are unlike as in the following examples.

$$
\begin{array}{ll}
\mathrm{Xj}=0101 & \mathrm{Xj}=0101 \\
\mathrm{Xk}=\frac{1100}{1001} & \mathrm{Xk}=001 \\
\mathrm{Xi}=1 & X i=0110
\end{array}
$$

The boolean unit forms the cornplemented result, which is recomplemented so that the true result is transmitted to the X register.

TRANSMIT Xi or $\overline{\mathrm{Xk}}$

Except that it complements Xk , the boolean unit executes both the transmit Xj and transmit $\overline{\mathrm{Xk}}$ instruction in essentially the same manner. Xj or $\overline{\mathrm{Xk}}$ enter an equivalence circuit in which the other operand is zero. The result is either $\overline{\mathrm{Xj}}$ or Xk. This result is recomplemented upon being sent to the Xi register. Transmit Xj is controlled by the gate Xk upper, gate Xk lower, and comp Xk signals.

## UNPACK

The boolean unit unpacks the floating-point quantity from the Xk register. It sends the 48 -bit coefficient to the Xi register and the 11 -bit exponent to the Bj register. The exponent bias is removed during the unpack operation so that the quantity in Bj is the true ones complement representation of the exponent. The sign bit of the exponent is extended to fill the 18 -bit B register. The true exponent result is gated to the Bj register by the go boolean signal. B register access control allows entry from the boolean unit for only the unpack instructions.

Except that Xk is not complemented, the Xk coefficient bits ( 0 through 47) follow the same paths as for the transmit $\overline{X k}$ instruction. Gate Xk lower controls this part of the operation. The extend $X$ signal causes the coefficient sign to be extended to bits 48 through 59 to fill the 60 -bit result in the $X$ register.

## PACK

The boolean unit packs a floating-point number in Xi. The coefficient of the number is obtained from Xk and the exponent from Bj . The boolean unit adds bias to the exponent before merging it with the Xk operand. If Xk is positive, the packed exponent occupying positions 48 through 58 of Xi is obtained from bits 0 through 10 of Bj by complementing bit 10 ; if Xk is negative, bit 10 is not complemented, but bits 0 through 9 are complemented. The comp $B$, gate $B$, and gate Xk lower signals control the pack operation.




The boolean unit executes the CPU instructions requiring bit-by-bit data manipulation. This includes the logical operations for instructions 11, 12, 13, 15, 16, and 17 plus the transmissive operations for instructions $10,14,26$, and 27.

## INPUT REGISTERS

The three input registers in the boolean unit receive data from the $\mathrm{Bj}, \mathrm{Xj}$, and Xk registers each clock period, without regard to the instruction in the CIW register. Data is transmitted to the input registers concurrent with the instruction issue. During the following clock period, data moves from the input registers through the static selection network and back to the operating registers. Thus, each instruction is executed in 2 clock periods. The boolean unit is free to begin executing a new instruction every clock period. If a boolean-type instruction does not issue in a given clock period, the data in the input registers is not used. New data enters the input registers in the following clock period.

## CONTROL

The boolean unit also receives bits of the $f$ and $m$ designators from the CIW register each clock period. Instruction designators $f$ bit 0 and $m$ bits 0 through 2 are held in registers and are then translated into control signals that determine the type of logical operation and select data paths required by the instruction.

The boolean unit receives the bo boolean signal in the clock period following issue of a boolean instruction. The go boolean signal enables the output of the boolean unit to the destination registers.

The data path to the destination X register for bits 48 through 59 on the 4 KN 7 module is shared by the various boolean instructions. Control signals from the 4KM7 module prevent conflicts by ensuring that only one data path is active at any one time. The active data path, containing the complemented result, merges with all ones on the other two paths. The result is recomplemented upon being sent to the destination $X$ register.

## LOGICAL PRODUCT

The logical product for instructions 11 and 15 is formed on the 4 KL 7 and 4 KN 7 modules. Xk is complemented if this operation is being performed for a 15 instruction. Xj is then ANDed with the corresponding Xk bits and the gate LP signal.

The complement of the logical product goes to another circuit where it is ANDed with the results of an equivalence circuit. For instructions 11 and 15 , the results of the equivalence operation are all ones because the Xj and Xk inputs to the equivalence circuit are not enabled. Thus, this second AND circuit for the logical product instructions acts like a merge. From the second AND circuit, the complemented logical product is recomplemented and sent to the destination X register.

## LOGICAL SUM

The logical sum instructions (12 and 16) use the same circuitry as the logical product instructions except that the inputs to the equivalence circuit are enabled. Instead of ones, meaningful data is ANDed with the complement of the logical product.

For example, if Xj equals 0101 and Xk equals 1100, the following logical operations take place.

1. The first AND circuit forms the logical product of Xj and Xk .

$$
\begin{aligned}
& X j=0101 \\
& X k=1100 \\
& L P=0100
\end{aligned}
$$

2. Both the Xj and Xk inputs to the equivalence circuit are enabled by gate Xj , gate Xk upper, and gate Xk lower so the equivalence circuit produces:
$\mathrm{Xj}=0101$
$\mathrm{Xk}=\underline{1100}$
$\mathrm{EQ}=0110$
3. The complement of the logical product $(\overline{\mathrm{LP}})$ is ANDED with EQ to produce:

$$
\begin{aligned}
& \mathrm{LP}=1011 \\
& \mathrm{EQ}=\underline{0110} \\
& \mathrm{Xi}=0010
\end{aligned}
$$

4. This result is recomplemented and sent to the destination $X$ register as 1001.

## LOGICAL DIFFERENCE

The logical difference instructions (13 and 17) use the equivalence circuit to form the complement of the result. The complemented result is ANDed with all ones from the logical product circuit, which is not enabled for these two instructions. The result of this merge operation is recomplemented and sent to the destination X register.

## TRANSMIT X

The boolean unit executes both transmit instructions (10 and 14) in essentially the same way. Either Xj or Xk is enabled into the equivalence circuit with the other operand a zero. The result is the complement of whichever operand was input. The result is then recomplemented upon being sent to the destination X register.

## UNPACK

The unpack operation for instruction 26 requires three separate data paths: one for the exponent to the destination B register, another for coefficient bits 0 through 47 to the destination X register, and a third for extending Xk sign to bits 48 through 59 of the destination X register.

Unpacking of the exponent takes place on the 4 KM 7 module. To remove bias and provide the proper sign, bits 48 through 57 of Xk enter a complement control circuit. This circuit complements bits 48 through 57 if the coefficient is positive
(bit 59 is not set). The output of the circuit is recomplemented to become bits 0 through 9 upon being gated to the destination B register. The complement of the sign of the exponent ( $\overline{\left.\mathrm{BJ}_{\mathrm{J}} \mathrm{sign}\right)}$ is determined by the logical difference of Xk bits 58 and 59. $\overline{\mathrm{Bj} \mathrm{sign}}$ is complemented and sent to the B register as bits 10 and 11 . This sign extension is repeated on the 4 KN 7 module for extending Bj sign to bits 12 through 17 of the B register. Bj and extend Bj sign bits are gated by the go boolean signal. $B$ register access control prevents entry to the destination $B$ register from the boolean unit for all instructions except 26.

Bits 0 through 47 of Xk are gated into the equivalence circuit on the three 4 KL 7 modules by the gate Xk lower signal. Since the Xj input to the equivalence circuit is not enabled, the result of the equivalence is the complement of Xk . This is recomplemented to become bits 0 through 47 in the destination X register.

Note that gate Xk upper is not set for this instruction, thereby preventing transmission of Xk bits 48 through 59 to the destination X register. Instead, the coefficient sign is extended to bits 48 through 59 by the extend X signal on the 4 KN 7 module. If the coefficient is positive, $\overline{\text { extend } X}$ is set and is ANDed with the go boolean signal. When ANDed with all ones from the other two paths and recomplemented, the signal causes all zeros to enter bits 48 through 59 of the destination X register. When the coefficient is negative, $\overline{\text { extend } \mathrm{X}}$ is clear, and the result is all ones to bits 48 through 59 of the destination X register.

## PACK

Just as unpack requires separate data paths to the $B$ and $X$ registers, packing for instruction 27 requires separate data paths from the $B$ and $X$ registers.

Packing of the exponent takes place on the 4 KN 7 module. Bj bit 11 is discarded because it is merely an extension of the exponent sign bit (bit 10). Bit 10 is complemented to add bias, and bits 0 through 10 are then gated into a complement control circuit by the gate $B$ signal. This circuit complements bits 0 through 10 if the coefficient is positive ( Xk bit 59 is not set). The output of the circuit is Xi bits 48 through 58 . These bits and the complement of the comp B signal (Xi bit 59) are ANDed with ones from the other paths and are recomplemented upon being sent to the destination X register.
Packing of the coefficient is a straightforward transmission of Xk bits 0 through 47 to the X register gated by gate Xk lower. Data paths for Xk bits 48 through 59 are blocked because gate Xk upper is not set for this instruction.

BOOL 3.0 TEST POINTS



| CONTROLOATA | SECONDARY BLOCK DIAGRAM BOOLEAN UNIT | $\left\|\begin{array}{l} \text { cest kent } \\ 34010 \end{array}\right\|$ | T006 6042030 | $00 \mid Y .$ |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { DEVELOPMENT } \\ & \text { OIVISION } \\ & \hline \end{aligned}$ |  | 6 | B00L $2.0 \mid$ | 5-2-i |

The boolean unit executes the C:PU instructions requiring bit-by-bit data manipulation. This includes the logical operations for instructions 11, 12, 13, 15, 16, and 17 plus the transmissive operations for instructions $10,14,26$, and 27 .

## INPUT REGISTERS

The three input registers in the boolean unit receive data from the $\mathrm{Bj}, \mathrm{Xj}$, and Xk registers each clock period, without regard to the instruction in the CIW register. Data is transmitted to the input registers concurrent with the instruction issue. During the following clock period, data moves from the input registers through the static selection network and back to the operating registers. Thus, each instruction is executed in 2 clock periods. The boolean unit is free to begin executing a new instruction every clock period. If a boolean-type instruction does not issue in a given clock period, the data in the input registers is not used. New data enters the input registers in the following clock period.

## CONTROL

The boolean unit also receives bits of the $f$ and $m$ designators from the CIW register each clock period. Instruction designators $f$ bit 0 and $m$ bits 0 through 2 are held in registers and are then translated into control signals that determine the type of logical operation and select data paths required by the instruction.

The boolean unit receives the go boolean signal in the clock period following issue of a boolean instruction. The go boolean signal enables the output of the boolean unit to the destination registers.

The data path to the destination X register for bits 48 through 59 on the 4 KN 7 module is shared by the various boolean instructions. Control signals from the 4KM7 module prevent conflicts by ensuring that only one data path is active at any one time. The active data path, containing the complemented result, merges with all ones on the other two paths. The result is recomplemented upon being sent to the destination $X$ register.

LOGICAL PRODUCT
The logical product for instructions 11 and 15 is formed on the 4 KL 7 and 4 KN 7 modules. Xk is complemented if this operation is being performed for a 15 instruction. $X j$ is then ANDed with the corresponding $X k$ bits and the gate LP signal.

The complement of the logical product goes to another circuit where it is ANDed with the results of an equivalence circuit. For instructions 11 and 15 , the results of the equivalence operation are all ones because the $\mathrm{Xj}_{\mathrm{J}}$ and Xk inputs to the equivalence circuit are not enabled. Thus, this second AND circuit for the logical product instructions acts like a merge. From the second AND circuit, the complemented logical product is recomplemented and sent to the destination $X$ register.

## LOGICAL SUM

The logical sum instructions (12 and 16) use the same circuitry as the logical product instructions except that the inputs to the equivalence circuit are enabled. Instead of ones, meaningful data is ANDed with the complement of the logical product.

For example, if Xj equals 0101 and Xk equals 1100 , the following logical operations take place.

1. The first AND circuit forms the logical product of $X j$ and $X k$.

$$
\begin{aligned}
& X_{j}=0101 \\
& X_{k}=\frac{1100}{0100} \\
& L P=
\end{aligned}
$$

2. Both the Xj and Xk inputs to the equivalence circuit are enabled by gate $X j$, gate Xk upper, and gate Xk lower so the equivalence circuit produces:
$X_{j}=0101$
$X k=1100$
$\mathrm{EQ}=\overline{0110}$


## PART 3

SHIFT UNIT

The shift unit executes CPU instructions 20, 21, 22, 23, and 43. These instructions shift the entire 60 -bit field of data within the operand word. Input operands are of two types. A 60 -bit word is read from either the Xi or the Xk register, depending upon the type of instruction. A second operand is read from either the CIW or the Bj register. This operand determines the shift count for the 60 -bit word. Shifted operands are sent from the shift unit to the Xi register.

## INSTRUCTIONS EXECUTED

## LEFT SHIFT Xi BY jk (20ijk INS'TRUCTION)

This instruction causes the shift unit to read one operand from the Xi register, shift the 60 -bit word left circularly by jk bit positions, and write the resulting 60 -bit word back into the same Xi register.

The shift unit does not actually shift data in a left circular mode. Instead, it simulates the shift. The shift count is complemented, a three-bit left circular shift correction is performed, and the shift is made in a right circular mode.

## RIGHT SHIFT Xi BY jk (21ijk IN§TRUCTION)

This instruction causes the shift unit to read one operand from the Xi register, shift the 60 -bit word to the right with sign extension by jk positions, and write the resulting 60 -bit word back into the same Xi register.

## LEFT SHIFT Xk BY Bj TO Xi (22ijk INSTRUCTION)

This instruction normally causes the shift unit to read one operand from the Xk register, shift the 60 -bit word left circularly by an amount determined by bits 0 through 5 of the Bj register, and write the resulting 60 -bit word into the Xi register. However, if the 18 -bit operand in the Bj register is negative (bit 17 is a one), the shift is to the right with sign extension, and the shift count is deter-
mined by the complement of Bj bits 0 through 5 .
The shift unit always complements Bj bits 0 through 5 during the execution of this instruction. The left shift is simulated by making a three-bit left circular shift correction and shifting in a right circular mode.

## RIGHT* SHIFT Xk BY Bj TO Xi (23ijk INSTRUCTION)

This instruction normally causes the shift unit to read one operand from the Xk register, shift the 60 -bit word to the right with sign extension by an amount determined by Bj bits 0 through 5, and write the resulting 60 -bit word into the Xi register. However, if the 18 -bit operand in the Bj register is negative (bit 17 is a one), the shift is left circular, and the shift count is determined by the complement of Bj bits 0 through 5.

The shift unit does not actually shift data in a left circular mode when Bj is negative; it simulates the shift. A three-bit left circular shift correction is performed and the shift is made in a right circular mode. The shift count is complemented because the Bj register contains a negative number.

## MASK UPPER jk PLACES OF Xi (43ijk INSTRUCTION)

This instruction causes the shift unit to generate a 60 -bit word containing ones in the upper $j k$ bit positions and zeros in the remaining bit positions. The resulting 60-bit word is written into the Xi register.

## INSTRUCTION TRANSLATION

The instruction translator determines what instruction has been issued by translating thee bits from the instruction designators in the CIW register. These are f bit 1 and m bits 0 and 1 .

## OPERAND SELECTION

The operand holding register holds the data which is to be shifted. This 60-bit operand is selected in the CPU from either the Xi or Xk register. Before entering the holding register, the selected operand passes through a second selection network. For shift instructions which require a right shift of data, the operand enters the register unchanged. For shift instructions which require a left shift of data, the operand is left circularly shifted by three bit positions before entering the register. For the mask instruction, the operand is discarded and the register contains all zeros. Before entering the first shift rank, the operand is complemented in a complement control network if a 43 instruction (mask) has been issued or if the selected operand is negative.

## SHIFT COUNT

## SHIFT COUNT SELECTION

The amount of shift to be performed on the selected 60 -bit operand is determined by translating a six-bit operand from one of two sources. For 20, 21, and 43 instructions, the j and k designators from the CIW register are used. These designators are treated as a single six-bit quantity. For 22 and 23 instructions, the lower six bits from the Bj register are used. If a 20 or 22 instruction has been issued, the selected shift count bits are complemented in a complement control circuit.

## SHIFT COUNT TRANSLATORS AND FLAGS

The selected shift count bits are translated in three groups. Bits 0 and 1 determine which one of four rank 1 shift count flags will be set. These flags allow the 60 -bit operand to be shifted zero, one, two, or three bit positions in the first shift rank. Bits 2 and 3 determine which one of four rank 2 shift count flags will be set. These flags allow the 60 -bit operand to be shifted $0,4,8$, or 12 bit positions in the second shift rank. Bits 4 and 5 determine which one of four rank 3 shift count flags will be set. These flags allow the 60 -bit operand to be shifted $0,16,32$, or 48 bit positions in the third shift rank.

SHIFT COUNT GREATER THAN 63 TEST

Prior to the execution of a 22 or 23 instruction which requires a right shift, the shift count greater than 63 test network determines if the shift will be greater than 63 bit positions. If so, the enable shift signal is not generated, causing none of the rank 3 shift count flags to be set. With no flags set, the third shift rank outputs all zeros, and the shift unit outputs all ones to the X registers. This causes all zeros to enter the Xi registers.

## LEFT CIRCULAR SHIFT SIMULATION

For instructions requiring a left circular shift (20, 22 with $\operatorname{Bj}$ positive and 25 with Bj negative), the shift unit actually performs a right circular shift. The left circular shift signal enables the three right-shift ranks to shift circularly. The amount of right shift is determined by the complement of the shift count. For the left-shift instructions (20 and 22), the shift count is complemented as it enters the shift unit. For the right-shift instruction which results in a left shift ( 23 with Bj negative), the shift count is not complemented because it is already expressed as a negative number. When a left circular shift is simulated by doing a right circular shift using the complemented shift count, a three-bit position error is introduced. The error is caused because the maximum shift count specifies three bit positions more than the 60 -bit word. This error is corrected by the three-bit left circular shift network which shifts the operand before it enters the operand holding register.

## SHIFT RANKS

The three shift ranks shift the operand by an amount determined by the corresponding rank 1, 2, and 3 shift count flags. Each rank shifts the operand right circularly if the left circular shift signal is present. If this signal is absent, the shift is to the right with sign extension. Since the operand is always positive in the shift unit, the sign extension bits are always zeros.

A 43 instruction (mask) forces all ones into shift rank 1. These ones are rightshifted with zeros filling the vacated bit positions. The shift count is determined in the same manner as previously.

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The third rank of the shift unit outputs the shifted operand (complemented) to the X registers whenever a go shift signal is present. If this signal is not present, all ones are sent to the X registers. If the original operand was positive, the shifted operand is recomplemented before entering the Xi register. Negative operands are complemented twice in the shift unit. Therefore, they are not recomplemented.



The shift unit executes CPU instructions $20,21,22,23$, and 43. These instructions shift the entire 60-bit field of data within the operand word. Input operands are of two types. A 60 -bit word is read from either the Xi or the Xk register, depending upon the type of instruction. A second operand is read from either the CIW or the Bj register. This operand determines the shift count for the 60 -bit word. Shifted operands are sent from the shift unit to the Xi register.

## INSTRUCTIONS EXECUTED

## LEFT SHIFT Xi BY jk (20ijk INSTRUCTION)

This instruction causes the shift unit to read one operand from the Xi register, shift the 60 -bit word left circularly by $j k$ bit positions, and then write the resulting 60 -bit word back into the same Xi register.

The shift unit does not actually shift data in a left circular mode. Instead, it simulates the shift. The shift count is complemented, a three-bit left circular shift correction is performed, and the shift is made in a right circular mode.

## RIGHT SHIFT Xi BY jk (21ijk INSTRUCTION)

This instruction causes the shift unit to read one operand from the Xi register, shift the 60 -bit word to the right with sign extension by jk bit positions, and write the resulting 60 -bit word back into the same Xi register.

## LEFT SHIFT Xk BY Bj TO Xi (22ijk INSTRUCTION)

This instruction normally causes the shift unit to read one operand from the Xk register, shift the 60 -bit word left circularly by an amount determined by Bj bits 0 through 5, and write the resutling 60 -bit word into the Xi register. How-
ever, if the 18 -bit operand in the Bj register is negative, the shift is to the right with sign extension, and the shift count is determined by the complement of Bj bits 0 through 5.

The shift unit always complements Bj bits 0 through 5 during the execution of this instruction. The left shift is simulated by making a three-bit left circular shift correction and shifting in a right circular mode.

## RIGHT SHIFT Xk BY Bj TO Xi (23ijk INSTRUCTION)

This instruction normally causes the shift unit to read one operand from the Xk register, shift the 60 -bit word to the right with sign extension by an amount determined by Bj bits 0 through 5 , and write the resulting 60 -bit word into the Xi register. However, if the 18 -bit operand in the Bj register is negative, the shift is left circular and the shift count is determined by complemented bits 0 through 5.

The shift unit does not actually shift data in a left circular mode when Bj is negative; it simulates the shift. A three-bit left circular correction is performed and the shift is made in a right circular mode. The shift count is complemented because the Bj register contains a negative number.

MASK UPPER jk PLACES OF Xi (43ijk INSTRUCTION)

This instruction causes the shift unit to generate a 60 -bit word containing ones in the upper $j k$ bit positions and zeros in the remaining bit positions. The resulting 60 -bit word is written into the Xi register.

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## INSTRUCTION TRANSLATION

The 3KO7 module determines what instruction has been issued by translating three bits from the instruction designators in the CIW register. These are f bit 1 and $m$ bits 0 and 1 .

## OPERAND SELECTION

The operand holding register (4KP7 modules) holds the data which is to be shifted. This 60 -bit operand is slected from either the Xi or Xk register. The selection is made in the CPU (4RF7 and 4RG7 modules) and is based on the value of $m$ in the CIW register. The Xi data. path is selected if the m designator has an octal value of $0,1,4$, or 5 . The 3 Kk data path is selected when this value is 2,3 , 6 , or 7 .

Before entering the operand holding register, the selected operand passes through a second selection network ( $4 \mathrm{~K}: 7$ modules). For 2 X series instructions which require a right-shift of data, the operand enters the register unchanged. For 2 X series instructions which require a left-shift of data, the operand is left circularly shifted by three bit positions before entering the register. For 43 instructions, the operand is discarded. In this case, the register contains all zeros.

Before entering the first shift rank, the operand is complemented in a complement control network (4KP7 modules) if the complement control flag is set. This flag is set when a 43 instruction has been issued or if the selected operand is negative (operand must be positive). A 43 instruction is indicated when the f designator from the CIW register has a zero in the middle bit (XOX). A negative operand is indicated when bit 59 (sign) of the selected operand is a one.

## SHIFT COUNT

## SHIFT COUNT SELECTION

The shift count to be performed on the selected 60 -bit operand is determined by translating a six-bit operand from one of two sources. The selection of one of these sources is made in the 3 KO 7 module. For 20,21 , and 43 instructions, the j and k designators from the CIW register are used. These designators are treated as a single six-bit quantity. For 22 and 23 instructions, the lower six bits from the Bj register are used. If a 20 or 22 instruction has been issued, the selected shift count bits are complemented.

## SHIFT COUNT TRANSLATORS AND FLAGS

The selected shift count bits are translated in three groups. Bits 0 and 1 are sent to a translator in the 4KP7 modules. This translator determines which one of four flags will be set. These flags allow the 60 -bit operand to be shifted zero, one, two, or three bit positions in the first shift rank (4KP7 modules). Shift count bits 2 and 3 are translated in the 3 KO 7 module. This translator determines which one of four flags in the 4KP7 modules will be set. These flags allow the 60 -bit operand to be shifted $0,4,8$, or 12 bit positions in the second shift rank (4KQ7 and 4KW7 modules). Shift count bits 4 and 5 are also translated in the 3 KO 7 module. This translator determines which one of four flags in the 4KP7 modules will be set. These flags allow the 60 -bit operand to be shifted 0,16 , 32 , or 48 bit positions in the third shift rank ( 4 KQ 7 and 4 KW 7 modules).

## SHIFT COUNT GREATER THAN 63 TEST

Prior to the execution of a 22 or 23 instruction, a network in the 3 KO 7 module determines if the right-shift is greater than 63 bit positions. If so, the enable shift signal is not generated. This causes none of the third shift rank flags to set. With no flags set, the third shift rank outputs all zeros and the shift unit outputs all ones to the X registers. This causes all zeros to enter the Xi register.


## LEFT CIRCULAR SHIFT SIGNAL AND FLAG

For instructions requiring a left circular shift, the shift unit actually shifts the operand in a right circular mode. The left circular shift signal generated by the 3KO7 module enables the shift ranks to simulate this shift. This signal enables a three-bit left circular shift correction in the 4 KP 7 modules. It also sets the left circular shift flag (lower 4KP7 module). When this flag is set, the shift ranks perform a right circular shift.

## SHIFT UNIT OUTPUT

The third rank of the shift unit (4KQ7 and 4 KW 7 modules) outputs the shifted operand (complemented) to the X registers whenever a go shift signal from the 4LE7 module is present. If this signal is not present, all ones are sent to the $X$ registers. If the original operand was positive, the shifted operand is recomplemented before entering the Xi register under control of the X register sign control. Negative operands are complemented twice in the shift unit. Therefore, they are not recomplemented.

SHIFT 3.0 TEST POINTS





PART 4

## NORMA LIZE UNIT

The normalize unit executes CPU instructions 24 and 25. These two instructions are identical except instruction 25 adds a round bit to the coefficient.

The normalize unit operates on operands in positive (true) form so it complements negative operands before operating on them. It then left-shifts the coefficient by an amount that causes a one jit to appear in the most significant position. The normalize unit adjusts the exponent by subtracting the shift count.

## OPERATION

The normalize instructions require 3 clock periods for execution. Data moves from the operating registers to the normalize unit in the same clock period in which the instruction issues from the CIW register. Input registers hold the information necessary to complete the instruction for use during the second clock period. During the second clock period, the normalize unit complements the operand if the sign was negat:ve, and a static network determines the number of shifts necessary to normal.ze the coefficient. The shift count determination network sends a six-bit shift count to the first stage of the exponent adder for exponent correction, to a register that holds the shift count for transmission to the $B$ register, and to the shift count translator. The shift count translator translates the six-bit binary shift count into control signals: shift $0,2,4,8$, 16, 32,48 , and no shift. The decoded shift count register holds this decoded shift count for use during the third clock period.

The shift count determination network also sends bit zero of the binary shift count to shift rank 1 where it left-shifts the coefficient one position if bit 0 is a one. During the third clock period, shift ranks 2, 3, and 4 complete the shifting of the coefficient, and the second stage of the exponent adder completes the subtraction of the shift count from the exponent. Go normalize gates the shift count to the $B$ register, and i.f there are no special cases, gates the corrected exponent and shifted coefficierit to the destination $X$ registers.

## SPECIAL CASES

After the normalize unit corrects the exponent for sign, it tests for an overflow or indefinite quantity. If the exponent is overflow or indefinite, the unit blocks the shift count, the operand passes through the unit unaltered. The quantity delivered to the B register is zero and a bit is set in the exit condition register (CPU).

If the coefficient portion of the operand is all zeros after the correction for sign, the shift count is 48 . If this situation occurs in execution of a 25 instruction, the round bit results in a normalized coefficient. If this situation occurs in execution of a 24 instruction, the shift count of 48 is a special case and the result entered in the destination X register is all zeros. The shift count delivered to the $B$ register is 48 in either of these cases.

The subtraction of the six-bit shift count from the exponent may result in an underflow of the floating-point exponent range. This situation causes all zeros to enter the destination $X$ register. The shift count delivered to the $B$ register is not affected by this situation.



## NORMALIZE UNIT

The normalize unit executes CPU instructions 24 and 25 . These two instructions are identical except instruction 25 adds a round bit to the coefficient.

The normalize unit operates on operands in positive (true) form so it complements negative operands before operating on them. It then left-shifts the coefficient by an amount that causes a one bit to appear in the most significant position (bit 47). The normalize unit adjusts the exponent by subtracting the shift count.

## EXPONENT AND CONTROL

## 4EE7 MODULE

Xk bits 48 through 59 and instruction designators $m$ bit 0 enter registers in this module when the instruction issues from the CIW register. A static network complements bits 48 through 58 if bit 59 is a one to place the exponent in biased positive format. A ones check is then performed on exponent bit 0 through 9 . Bits 0 through 9 are all ones if an exponent overflow condition (3777) or an exponent indefinite condition (1777) exists. If either of these conditions exists, the ones check detects it and blocks enable shift. Blocking enable shift causes the operand to pass unchanged through the normalize unit and a zero shift count to be sent to the $B$ register.

This module also sends bit 59 ( X sign) to the CPU X register sign control. This controls the complement of the result at the input to the destination $X$ register.

If the instruction is a round normalize (25), $m$ bit 0 is a one and sets the round flag. The round flag sends a round signal to the 3EF7 and 4EC7 modules. Round causes a bit to be added to the coefficient in a position immediately below the bit zero position of the coefficient before it is shifted. The shift ranks then the round bit along with the coefficient.

3EF7 MODULE

Exponent bits 0 through 10 enter this module from the 4EF7 module during the second clock period, where the first stage of subtraction of shift count bits 0 through 5 from the exponent takes place. Enable shift gates shift count bits 0 through 5 into the first stage of the adder. Registers hold the partial difference for use during the third clock period. The $3 E F 7$ module also tests for a zero coefficient and a complete underflow.

In the case of a 24 instruction (no round bit) with a coefficient equal to zero, a normalized result is impossible. The zero coefficient results in a shift count of 48 , and combined with round and enable shift, blocks gate output. Blocking gate output causes all ones to be sent to the X registers. A normalize underflow signal is sent to the $X$ register input control ( 5 CB 7 module) which complements the word of ones causing all zeros to enter the destination X register.

If subtraction of the shift count causes the exponent to exceed its negative range, a complete underflow results. This is detected by testing the bit enables and bit borrows. A complete underflow blocks gate output which causes all ones to be sent to the $X$ registers. A normalize underflow signal is sent to the $X$ register input control (5CB7 module) which complements the word of ones causing all zeros to enter the destination X register.

Go normalize enters the module during the second clock period and the go normalize flag-2 holds it for use during the third clock period. Go normalize enables gate output under normal conditions and also causes the extension of zero bits in B register bits 6 through 17.

This module tests for indefinite or infinite operands. For indefinite operands (exponent equals 1777), the complement of exponent bit 10 and the complement of enable shift are ANDed with the go normalize signal. For infinite operands (exponent equals 3777 ), exponent bit 10 and the complement of enable shift are ANDed with the go normalize signal. The indefinite and infinite signals are sent to the CPU to set the corresponding bit in the exit condition register (CPU).

## 4EH7 MODULE

This module performs the second stage of subtraction of the shift count from the exponent. Gate output gates the output of the adder to the $X$ registers. The result is complemented as it is sent to the $X$ registers. If the result is positive, it is recomplemented by the $X$ register input control (5CB7 module) as it enters the destination $X$ register.

## COEFFICIENT

4EA7 MODULE

Xk bits 0 through 47 enter this module when the instruction issues from the CIW register. The module contains the coefficient input register and complements the coefficient if bit 59 is a one. This changes the coefficient to positive format in the same way the $4 \mathrm{EE7}$ module corrected the exponent for sign. The 48 coefficient bits go to the 4 EC 7 and 4 EB 7 modules.

## 4EB7 MODULE

This module contains the static shift count determination network. its output is a six-bit binary shift count equal to the number of zero bits from the most significant one bit on the unshifted corfficient up to and including bit 47. All six bits of this shift count go to the 3 EF7 and 4 EG 7 modules. Shift count bit 0 goes to the 4EC7 modules.

4EG7 MODULE

Enable shift gates the shift count into these modules during the second clock period. The shift count translator translates the six-bit shift count into shift control signals of shift $0,2,4,8,12,16,32,48$, and no shift. Registers hold them for use in the 4ED7 modules during the third clock period. A register also holds shift count bits 0 through 5 for use during the third clock period. Extend Bj gates the shift count to $B$ register bits 0 through 5 .

## FC7 MODULF

The coefficient bits 0 through 47 enter these modules during the second clock period. A left-shift of one takes place in shift rank 1 if shift count bit 0 and enable shift are both ones. A register holds the output of shift rank 1 for use in the 4ED7 modules during the third clock period.

4ED7 MODULE

Coefficient bits 0 through 47 enter these modules during the third clock period Shift ranks 2, 3, and 4 complete the left shifts needed to normalize the coefficient. Gate output gates coefficient bits 0 through 47 to the destination $X$ register. The result is complemented as it is sent to the $X$ registers. If the result is positive, it is recomplemented by the X register input control as it enters the destination X register.

NORM 3.0 TEST POINTS




PART 5

## FLOA TING-ADD UNIT

## DIAGRAM LAYOUT

The floating-add unit primary block diagram is drawn with the registers and flip-flops arranged in vertical columns according to clock periods.

There are three columns of registers. The first column on the left has the m 1 input flip-flop on top and the borrow register for the first stage of Xk minus Xj exponent on the bottom. The second column of registers has the m 1 flip-flop on top and the reference sign flip-flop on the bottom. The third column has the DP flip-flop on top and the borrow register for the first stage of shifted operand plus reference operand on the bottom.

Because of this arrangement, this diagram shows everything that happens in each clock period. Everything happening on the left side of the first column of registers, including the setting of the registers, occurs during the first clock period. Everything happening between the first and second columns of registers, including the setting of the second column of registers, occurs during the second clock period. Everything happening between the second and third columns of registers, including the setting of the third column of registers, occurs during the third clock period. Everything happening to the right of the third column of registers, including the setting of the result into the Xi register in the CPU, occurs during the fourth clock period.

The exponent takes a path from left to right across the upper half of the diagram, and the coefficient takes a path across the lower half of the diagram. Shift count determination and shift control are across the middle. The remaining control is across the top of the diagram.

## GENERAL OPERATION

The floating-add unit performs CPU instructions 30 through 35. They are:
Floating sum (30)
Floating difference (31)
Floating double-precision sum (32)

Floating double-precision difference (33)
Round floating sum (34)
Round floating difference (35)
The $m$ designator of the instruction controls the type of operation the unit performs. Bit 0 controls the mode of operation (add or subtract). Bit 1 controls single or double precision. Bit 2 controls the rounding operation.

Execution time is 4 clock periods.

The unit receives two $60-$ bit operands in floating-point format from the X registers specified by the $j$ and $k$ designators of the instruction.

If the instruction is a 31,33 , or 35 (subtract), $m$ designator bit 0 causes the unit to complement the Xk operand. Thus, the unit subtracts by complementary addition.

The unit tests both exponents for overflow and indefinite. If a special case is detected, the output of the unit is blocked and the appropriate special case flag is sent to the CPU.

Before the two coefficients can be added, they must be aligned in a manner that causes the exponents to be equal. Thus, each bit in the adder has equal significance with the bit to which it is added. The unit aligns coefficient bits of equal significance by right-shifting the coefficient having the smaller exponent by an amount equal to the difference between the two exponents. The coefficient selected for shifting is called the shifted operand. The coefficient having the larger exponent is called the reference operand. The difference between the two exponents is determined by an 11 -bit adder. The output of this adder translates into shift control signals. These signals control the shift ranks that shift the shift operand. This adder also determines which exponent is the larger and outputs a signal called sign of difference. Sign of difference is equal to one when Xj is the larger exponent. It controls the selection of the exponent and which coefficient becomes the shifted operand.

If the two exponents are equal so that no shifting is required, all the shift control signals are zero and the shifted operand passes through the shift ranks unchanged. Both coefficients occupy the upper half of the 97 -bit shifted and reference operands with the signs of the two coefficients filling the lower half. The binary points are in the middle of the 97 -bit operands (between bits 47 and 48). If the exponents are different so that alignment is required, the coefficient selected for shifting is right-shifted within the 97-bit shifted operand. The sign of this coefficient fills in above and below as it is shifted.

After the amount of the difference between the two exponents is determined, the smaller exponent is discarded, under control of sign of difference. The larger one is selected to continue through the unit to become the exponent of the result.

When the command is a 34 or 35 (round) instruction, $m$ designator bit 2 causes the unit to add a round bit to one or both operands. If both operands are normalized or if the signs of the two operands are different, the round bit is added to both operands. If the above conditions are not met, the round bit is added only to the reference operand. The round bit added to the reference operand is always in bit position 47. This is immediately to the right of the binary point. If the round bit is also added to the shifted operand, it starts out in bit position 47 but is right-shifted with the rest of the coefficient. If no shifting takes place, the round bit remains in bit position 47 . Round bits are equal to the complement of the sign of the operands to which they are added. After the alignment, the two operands are added in a 99 -bit ones complement adder.

Instructions $30,31,34$, and 35 (single precision) cause the unit to deliver the upper half of the 99 -bit result to Xi location 0 through 47. The result exponent and sign bit are packed in floating-point format into Xi locations 48 through 59.

Instructions 32 and 33 (double precision) cause the unit to deliver the lower half of the 99 -bit result to Xi locations 0 through 47 . These 48 bits are to the right of the binary point, so the unit subtracts 48 from the result exponent to correct for double precision before sending it to the Xi register.

When the coefficient sum overflows the highest order bit of the result coefficient, the unit detects it by testing the signs of the two operands and determining whether or not there was a borrow (carry) into bit 96. If the unit determines that there was a coefficient overflow, the 48 -bit coefficient delivered to Xi is taken from the 99 -bit adder result one bit higher than if no overflow occurs. If overflow occurs, the exponent is increased by one prior to sending it to Xi .



INPUT: SPECIAL CASE TEST: SHIFT COUNT DETERMINATION;
COEFFICIENT OUTPUT SELECTION CONTROL; EXPONENT SELECTION

## INPUT

The 60 -bit Xj and Xk operands enter the 4 FF 7 module during the clock period in which the instruction issues from the CIW register. Instruction designator $m$ bits 0 and 2 also enter the 4 FF 7 module at the same time. Designator m bit 1 enters the 4RD7 module (which will be discussed later).

Designator $m$ bit 0 differentiates between an add and a subtract command. If $m$ bit 0 is a one, the command is a subtract so it complements all 60 bits of the Xk operand. The 4 FF7 module contains registers which hold both 60 -bit operands for use during the second clock period.

## ROUND BIT CONTROL

Designator $m$ bit 2 controls the rounding of the coefficients. A flip-flop holds m bit 2 for use during the second clock period. The output of the m bit 2 input flip-flop is round reference. Round reference causes the round bit to be added to the operand with the largest exponent (reference operand) or the Xk operand if the exponents are equal. A round bit is also added to the operand with the smaller exponent (shift operand) if both operands are normalized or if the signs of the operands are not equal. The 4FF7 module tests for normalized operands by comparing bit 47 to the sign bit (59). The result of these tests is ANDed with round reference and causes the shift operand to be rounded.

The sign of the operand controls the binary state of the round bit for that operand. The state of the round bits is always different than the sign bit. This round bit preparation takes place on the 4 FC 7 modules

The 4 FC 7 modules prepare round bits before the determination of which operand will be the reference and which will be the shift operand. Therefore, if the round reference signal is a one, the 4 FC 7 modules output a reference round bit for both the Xk and Xj operands. If the round shift signal is a one, shift round bits are output for both operands.

## BIAS REMOVAL

The 4 FC7 modules remove the bias, complement the exponent for negative coefficients, and test for overflow and indefinite. The sign bit (59) controls the removal of bias by complementing exponent bits 0 through 9 if the sign is negative. The sign bit controls bit 10 (bias bit) opposite to bits 0 through 9 . For example, if the sign is negative, exponent bits 0 through 9 are complemented, and bit 10 is not complemented. If the sign is positive, exponent bits 0 through 9 are not complemented, and bit 10 is complemented.

## OVERFLOW AND INDEFINITE TEST

The tests for overflow and indefinite take place after the removal of bias. $\mathrm{Be}-$ fore removal of bias, an exponent of 3777 or 4000 indicates overflow. After removal of bias, an overflow is 1777. An indefinite exponent before bias removal is 1777 or 6000. Unbiased, indefinite is 3777.

## SHIFT COUNT DETERMINATION

The exponents from the two floating-point format operands enter the 4 FA7 module during the clock period that the instruction issues from the CIW register. Static networks remove the bias from both exponents and complement the Xk exponent. The exponents are also complemented if their coefficients are negative. The exponents then enter the first stage of an adder that subtracts the Xj exponent from the Xk exponent by complementary addition. Registers store the partial result for use in the $4 \mathrm{FB} 7,4 \mathrm{FD} 7$, and 3 FE 7 modules during the second clock period.

The 4FB7, 4FD7, and 3FE7 modules each contain a second stage adder which forms results that are shift control signals.

The second stage adder in the 4 FB 7 module receives enables and borrows from the 4 FA7 module and forms a result which is shift $1,2,4$, and sign of difference. Sign of difference is a one when the Xj exponent is larger than the Xk exponent. In this case, the shift 1,2 , and 4 signals are in complement form. The 4 FB 7 module sends shift 1,2 , and 4 signals to shift ranks 1,2 , and 3 where they control the shifting of the shift operand. Sign of difference is used for operand selection.

The second stage adder in the 4 FD 7 module receives group enables and group borrows from the 4 FA 7 module. The adder also receives Xk and Xj exponent bits 3 through 5 from the 4 FC 7 modules. From these inputs, it forms results which are shift 8,16 , and 32 control signals. These shift terms are always in true form. A register holds them for use during the third clock period on the 3 FH 7 modules.

The second stage adder in the 3 FE7 module receives group enables and group borrows from the 4 FA 7 module. The adder also receives Xk and $\mathrm{X} . \mathrm{i}$ exponent bits 6 through 10. From these inputs, it forms results which are shift 64 and 128 control signals. A register holds them for use during the third clock period on the 3 FH 7 module.

## EXPONENT SELECTION

During the second clock period of execution, the unit makes a selection between the operands. One operand becomes the reference operand, and the other becomes the shift operand. The coefficient for this shift operand goes to a shift network where it is right-shifted by an amount equal to the difference between the two exponents. The unit discards the exponent of the shift operand. The sign of the shift operand becomes the shifted sign, and the sign of the reference operand becomes the reference sign.

The selection of which exponent will be the reference exponent takes place in the 4 FD7 and 3 FE7 modules. The sign of difference makes the choice and gates the larger exponent into the selected exponent register. Sign of difference is equal to one when $\mathrm{X}_{\mathrm{j}}$ is the larger exponent.

Sign of difference gates the sign of the operand with the smaller exponent into the shifted sign flip-flop and the sign of the operand with the larger exponent into the reference sign flip-flop. These registers and flip-flops hold the selected exponent, shifted signs, and reference signs for use during the third clock period. The 4 FD 7 module handles the selection of exponent bits 0 through 5 . The $3 F E 7$ module handles the selection of exponent bits 6 through 10 , shifted signs, and reference signs.

## EXPONENT CORRECTION FOR DOUBLE PRECISION

Designator $m$ bit 1 differentiates between a single-precision and a double-precision instruction. If $m$ bit 1 is a zero, the instruction is single-precision and the unit selects the upper 48 bits of the 97 -bit result as the result coefficient. If m bit 1 is a one, the instruction is double-precision and the unit selects the lower 48 bits of the 97 -bit result register as the result coefficient. Since the binary point is effectively right-shifted 48 วlaces, it is necessary to adjust the exponent by subtracting 48 from it.

Designator m bit 1 enters the 4FD7 module during the clock period in which the instruction issues from the CIW register. Flip-flops hold it for 2 clock periods for use during the third clock period. The output of the two flip-flops is doubleprecision and goes to the 3 FM7 module.

During the third clock period, exponent bits 0 through 10 enter the 3 FM7 module from the selected exponent register in the 4 FD 7 and 3 FE 7 modules. A static network in the 3FM7 module complements bits 0 through 10 and feeds them into an adder that subtracts 60 (octal) if $D P$ is a one. If $D P$ is a zero, exponent bits 0 through 10 pass through the adder unaltered. The exponent always leaves this adder in true form so it needs no complementing. A register holds the result for use during the fourth clock period.

## SPECIAL CASE TEST

The floating-add unit senses overflow, underflow, and indefinite special cases. The 4 FC 7 modules test for overflow and indefinite for each operand and send the results to the 4 FD7 module. The 4 FD7 module stores Xk and Xj overflow for use during the third clock period. It also ORs Xk and Xj indefinite and stores the result in the indefinite flip-flop for use in the 3FM7 module during the third clock period.

Static networks in the 3 FM 7 module test these conditions and send the proper special case flag to the X sign control translator during the third clock period.

The following table shows which flag is sent to the X sign control translator for different conditions of overflow and indefinite. For example, if either operand is indefinite, the indefinite flag is set. If both operands are negative overflow, the overflow flag is set.
$\mathrm{W}=$ Operand with no special cases
$-\infty=$ Overflow with negative sign
$+\infty=$ Overflow with postive sign
IND = Indefinite flag
OVF = Overflow flag

|  |  | Xk |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | W | + | $-\infty$ | IND |
| Xj | W |  | OVF | OVF | IND |
|  | $+\infty$ | OVF | OVF | IND | Ind |
|  | - - | OVF | Ind | OVF | IND |
|  | IND | IND | IND | IND | IND |

Signs are reference and shifted.

An underflow occurs when the unbiased selected exponent is a number mor negative than 2060 (octal) and the instruction is double-precision. The correction for double-precision subtracts 60 (octal) from the exponent causing it to exceed the most negative end of its range. If this happens, the $3 F M 7$ module sends the underflow signal to the X sign control translator.

During the second clock period, the 4FD7 module receives the go floating-add signal from the CPU if the instruction code is 30 through 35 . The go holding flip-flop holds it for use during the third clock period in the 3 FM 7 module. During the third clock period, the go signal enters a flip-flop in the 3FM7 module and enables output data to the destination X register during the fourth clock period. If go floating-add is not set, the data is discarded.

If the go floating-add signal is received and a special case flag is set, go add special case is set. This signal, together with the special case flag and the sign of the reference operand, goes to the $X$ sign control translator and causes the $X$ register input circuits to generate the proper special case result.

The 3FM7 module generates indefinite-1 and infinite signals. When these signals occur, the corresponding bit is set in the exit condition register (CPU). The indefinite -1 signal is generated when go and indefinite signals are present. The infinite signal is generated when go and either Xk or Xj overflow signals are present.

## COEFFICIENT OUTPUT SELECTION CONTROL

During the fourth clock period, the floating-add unit transmits the results to the destination $X$ register. There are four possible data paths from the coefficient portion of the unit to the $X$ registers. The unit selects one of these four paths on the basis of instruction mode and if coefficient overflow occurred.

There are two possible output data paths from the 97 -bit result for single-precision commands, one for normal single precision (bits 48 through 95) and one for coefficient overflow (bits 49 through 96).

There are also two possible output data paths from the 97-bit result for doubleprecision commands, one for normal double precision (bits 0 through 47) and one for coefficient overflow (bits 1 through 48).

The coefficient output selection control on the 3 FM 7 module controls the selection of which half of the 97 -bit result is transmitted from the unit.

If the instruction is single precision with no special cases and a go floating-add is received from the CPU, the sample upper half and sample center signals are ones. These signals are held for use during the fourth clock period in the 4 FK 7 module.

If the instruction is double precision with no special cases and go floating-add is received from the CPU, the sample lower half and sample center signals are ones. These signals are held in the 3 FM7 module for use during the fourth clock period by the 4 FK 7 module.

EXPONENT SAMPLE CONTROL

The exponent for the result coefficient is transmitted to the destination X register during the fourth clock period. Control for enabling the exponent output takes place on the 3FM7 module. If there are no special cases and go floating-add is received from the CPU, the sample exponent signal is sent to the $4 F L 7$ module where it enables the output of the exponent.

FAD 3.0 TEST POINTS

| Module | Location | Test Point | Description | Module | Location | Test Point | Description | Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 74 <br> 73 <br> 71 <br> 66 <br> 35 <br> 43 <br> 53 <br> 65 <br> 72 <br> 76 <br> .36 <br> 42 <br> 52 <br> 61 75 <br> 25,42 <br> 51, 62 <br> 76 <br> 71 <br> 73 <br> 06 <br> 01 <br> 11 <br> 16 <br> 21 <br> 64-66 <br> 63 <br> 32,33 <br> 41,46 <br> 51, 56 <br> 61 <br> 41, 73 <br> 25 <br> 05 <br> 06 <br> 11 <br> 16 <br> 23 <br> 31, 36 <br> 51, 55 <br> 75, 76 <br> 76 <br> 72 <br> 73 <br> 02 <br> 01 <br> 22 <br> 21 |  |  |  | 35 35 36 55 56 56 65 63 04 06 06 24 26 31 33 53 51 51 61 62 71 56 66 62 33 33 05 04 03 25 24 23 23 35 32 06 01 02 |  |  |  |  |  |

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Coefficient selection takes place during the second clock period on the 4 FG7 and 4 FO 7 modules. The coefficient with the larger exponent becomes the reference operand and the other becomes the shifted operand. A register on the 4 FG 7 and 4 FO7 modules holds the reference operand for use during the third clock period. The coefficient having the smaller exponent becomes the shift operand. The shift operand enters a shift network which right-shifts it by an amount equal to the difference between the two exponents:

Sign of difference gates the coefficient having the larger exponent into the reference operand register. Sign of difference is a one when Xj is the larger exponent. Sign of difference also gates the proper round bits. If the Xk coefficient becomes the shift operand, the shift $X k$ round bit is selected for the shift operand. If the Xk coefficient becomes the reference operand, the reference Xk round bit is selected for the reference operand. The round bit is inserted one bit position to the right of bit 0 of the original coefficient. Sign of difference gates the coefficient having the smaller exponent into shift rank 1.

Both coefficients expand into 96 -bit operands before entering the adder. Since the reference operand is not shifted, the original coefficient becomes bits 48 through 95 of the 96 -bit reference operand with the round bit in position 47. Reference sign fills bits 0 through 46 of the reference operand.

## SHIFT

If the exponents are equal, there are no shifts so the shifted operand occupies bits 47 through 95 with bit 47 being the shift round bit. If the exponents are not equal, this coefficient is right-shifted by the amount of the difference. The bits on either end of the original coefficient in the 96 -bit shifted operand are filled with shifted sign.

The shifting takes place in the 4 FG7, 4FO7, and $3 F H 7$ modules. The $4 F G 7$ and 4 FO7 modules contain shift ranks 1, 2, and 3. The control signals for these three shift ranks are shift 1,2 , and 4 which come from the 4 FB 7 module. If sign of difference is a one, the 4 FB 7 module sends these signals in complementary form. To correct for this, the 4 FG 7 and 4 FO 7 modules recomplement them if sign of difference is a one. Therefore, when the shift signals enter the shift ranks, they are always in true form.

Shift rank 1 performs a right shift of zero or four bit positions under the control of shift 4. If shift 4 is a one, the shift rank performs a right shift of four bit positions. If shift 4 is a zero, the operand moves through the shift rank unchanged. Shift ranks 2 and 3 function in the same way under control of shift 2 and shift 1.

These first three shift ranks perform any combination of right shifts up through seven bit positions. If a shift of seven is performed, the round bit which was inserted in bit position 47 is shifted to bit position 40 . Therefore, the output of shift rank 3 is shifted operand bits 40 through 95.

Shift ranks 1 and 2 perform their shifts during the second clock period. The shifted operand register holds the result for use during the third clock period. Shift rank 3 performs its shift during the third clock period and sends the results to the 3 FH 7 modules for further shifting.

Shifted operand bits 40 through 95 enter the 3 FH 7 module during the third clock period where the remainder of the shifting takes place in shift ranks 4 and 5.

Control for these shift ranks comes from the 4FD7 and 3 FE7 modules during the third clock period. These control signals are shift $8,16,32,64$, and 128 . Shift $8,16,32$, and 64 control signals enter the shift count translator on the 3 FH7 module. The translator outputs two groups of shift control signals.

The first group of signals is shift $0,8,16$, and 24 . These signals go to shift rank 4 where they cause this shift rank to right-shift the shifted operand by 0 , 8 , 16, or 24 bit positions.

The second group of signals is shift $0,32,64$, and 96 . These signals are ANDed with shift 128 and go to shift rank 5 . Shift 128 goes directly to shift rank 5.

The output of shift rank 5 is shifted operand bits 0 through 95. The 3 FH7 module sends this shifted operand to the 4 FJ 7 and 4 FI 7 modules for the first stage of addition.

FAD 3.1 TEST POINTS



## ADDER

Shifted operand bits 0 through 95 and reference operand bits 49 through 95 enter the 4 FJ 7 and 4 FI 7 modules during the third clock period. Reference sign fills bits 0 through 46 of the reference operand. Bit 96 of the reference operand is filled with reference sign, and bit 96 of the shifted operand is filled with shifted sign. This makes room for coefficient overflow and controls the section borrow from section 11 (end-around borrow). Since the upper two bits of the adder are sign bits, if both operands are positive the adder always has an end-around borrow. If both operands are negative, the adder never has an end-around borrow. If the signs are not alike, the remaining bits control the end-around borrow.

Each 4 FJ 7 module handles nine bits, and the 4 FI 7 module handles seven bits. Both operands are complemented as they enter the 4 FJ 7 and 4 FI 7 modules and then enter the first stage of the adder. The first stage of the adder divides the operands into bits, groups, and sections and forms a partial sum consisting of nables and borrows.

## ADDER FORMAT

Each module is a section; therefore, sections 1 through 10 are nine bits long because they are on 4 FJ7 modules, and section 11 is seven bits long because it is on a 4 FI 7 module. A section borrow sets when there is a borrow-out of that section. A section enable sets when all bit enables in that section are set.

A group is three bits long. There are two groups per section. These two groups form the lower six bits of each section. A group enable sets when all bit enables in that group set. A group borrow sets when there is a borrow-out of that group. This partial sum is held in registers for use during the fourth clock period.

A second stage of the adder is on 4 FK 7 modules. The second stage receives the partial sum inputs and forms the sum. The sum at this point is in ones complement form.

COEFFICIENT OVERFLOW DETECTION

The result of an addition or subtraction instruction may be one bit longer than the operand's input into the adder. Therefore, since the adder receives 96 -bit operands, it is possible to overflow into the 97 th bit. When this happens, it is detected in the 4 FL7 module which generates a coefficient overflow signal. Coefficient overflow causes the result coefficient to be right-shifted one bit position to retain this overflow bit. Because of this right shift, the exponent is corrected by adding one to it.

The coefficient overflow detecting network is on the $4 F L 7$ module and is divided into two parts. The first part receives section and group enables and borrows from the first stage of the adder. It tests to see if a borrow will be propagated to bit 96. The second part receives the output from the first part and the signs of the two operands. It determines if there was an overflow of the coefficient.

To have a coefficient overflow, the signs of the operands must first be alike. If the signs are not alike, they actually subtract and cause a smaller result.

If the signs of both operands are positive, the coefficient overflow determination network receives borrow to 96 . If the signs are both negative, the network receives borrow to 96 .

## COEFFICIENT OUTPUT SELECTION

There are four coefficient data paths from the 97 -bit result coefficient. The 4 FK 7 modules output the upper 48 bits for single-precision instructions and the lower 48 bits for double-precision instructions. If coefficient overflow occurred, the 4 FK 7 modules select a different set of outputs that perform a wired right-shift of one bit position to the upperor lower half of the 97-bit result.

Normal output for single precision is bits 48 through 95 of the 97 -bit result. This data path is coefficient bits 48 through 95 and becomes bits 0 through 47 in the X register. If coefficient overflow occurred, the 4 FK 7 modules output bits 49 through

96 of the 97 -bit result. This coefficient overflow data path is coefficient bits 49 through 96 shifted which becomes bits 0 through 47 in the X register.

Normal output for double precision is bits 0 through 47 of the 97 -bit result which become bits 0 through 47 in the X register. If coefficient overflow occurred, the 4 FK 7 modules output coefficient bits 1 through 48 shifted, which become bits 0 through 47 in the X register.

Control of which half of the 97 -bit result coefficient is sent to the X register comes from the coefficient output selection control on the 3FM7 module. These control signals are sample upper half, sample center, and sample lower half. The double precision (CP) also comes from the 3 FM7 module.

The 4 FK 7 module in location 5 E 06 handles bits 45 through 53 of the 97 -bit result. Because some of its bits are in the upper half and some in the lower half, this module is the only 4 FK 7 module that has four possible outputs.

The DP, sample center, and coefficient overflow signals control the output of bits 45 through 53 in the following manner.

If DP, sample center, and coefficient overflow signals are all present, the data path for coefficient bits 45 through 48 shifted is enabled.

If only sample center and coefficient overflow are present, the data path for coefficient bits 49 through 53 shifted is enabled.

If only DP and sample center are present, the data path for coefficient bits 45 through 47 is enabled.

If only sample center is present, the data path for coefficient bits 48 through 53 is enabled.

If sample center is not present, all ones are sent to the X register.
Bits 54 through 96 are controlled by sample upper half and coefficient overflow signals. The 4 FK 7 module inputs for DP and $\overline{\mathrm{DP}}$ are both forced to a constant one. Thus, if sample upper half is present, data is gated from the upper half
of the 97-bit result. Coefficient overflow determines which set of outputs (shifted or unshifted) is used.

Bits 0 through 44 are controlled by sample lower half and coefficient overflow signals in the same manner as the upper half.

The upper unused bits, coefficient bits 96 through 98 and coefficient bits 97 , 98 shifted, are terminated. Coefficient bit 0 shifted is also terminated because the right shift for coefficient overflow is an end-off type shift and bit 0 is discarded.

EXPONENT CORRECTION FOR COEFFICIENT OVERFLOW, EXPONENT OUTPUT CONTROL

The exponent enters the 4 FN 7 module during the fourth clock period and has a choice of two paths through the module. The path taken by the exponent is under control of sample exponent, the signs of the two operands, and borrow to 96 .

Sample expunieni aliows the 4 FN 7 module to output the exponent. If sample exponent is not present, both exponent paths are blocked and all ones are output to the X register.

If the signs of the two operands are alike, borrow to 96 determines if there was a coefficient overflow. If coefficient overflow occurred, the exponent is corrected by the addition of one. This correction is necessary because the coefficient was right-shifted one bit position. If the signs of the two operands are different, borrow to 96 indicates which operand was the larger so the correct sign can be given the result.

Three combinations of signs and borrow to 96 cause the exponent to take the path through output 2. They are listed with the sign given the result coefficient, the form (true or ones complement) of the result exponent, and if there was coefficient overflow.


Both Signs Positive
Borrow to 96

+ sign of the result
Output in true form

No coefficient overflow

Both Signs Negative

## Borrow to 96

- sign of the result

Output in ones complement form

No coefficient overflow

Unlike Signs
Borrow to 96

+ sign of the result Output in true form

No coefficient overflow

When both signs are positive and borrow to 96 is a one, there is no coefficient overflow. Since both signs are positive, the result is positive and expressed in true form. Complement control 2 controls the form of the result exponent and adds bias, sign 2 gives it its sign, and output control 2 gates the exponent to the X register.

When the signs of the two operands are not alike, borrow to 96 indicates which operand is larger. In the case of output 2 , borrow to 96 is a zero so the positive operand is the larger of the two. Thus, the sign given the result exponent by sign 2 is positive, and complement control 2 causes the exponent to be output in true form.

Three combinations of signs and borrow to 96 cause the exponent to take the path through the exponent plus $1 / 0$ adder and output 1 . They are listed with the sign given the result coefficient, the form of the result exponent, and if there was coefficient overflow.

| Both Signs Positive | Both Signs Negative | Unlike Signs <br> Borrow to 96 |
| :--- | :--- | :--- |
| Borrow to 96 | Borrow to 96 |  |
| Output in true form | -sign of the result <br> Output in ones comple- <br> ment form | Output in ones comple- <br> ment form |
| Coefficient overflow | Coefficient overflow | No coefficient overflow |

Borrow to 96

- sign of the result ment form No coefficient overflow

When both signs are positive and borrow to 96 is a zero, there is a coefficient overflow. The exponent takes the path through the adder where plus 1 causes the adder to add ones to the exponent. Complement control 1 causes the exponent to be output in true form and adds bias, sign 1 gives the result coefficient a positive sign, and output control 1 gates the result exponent to the X register.

Complement control 1 also ensures that the exponent correction for coefficient overflow does not result in the unit sending out a negative 0 exponent. If both signs are positive and the exponent takes the path through output 1, it is because coefficient overflow and the adder add one to the exponent. In this case, if the exponent is negative 1 , the addition of one results in a negative 0 exponent. Complement control 1 senses this and causes the 4 FN 7 module to output a positive 0 .

When the signs of the two operands are not alike and borrow to 96 is a one, there is no coefficient overflow. The exponent takes the path through the adder but plus 1 is a zero because the signs are not alike. This causes the exponent to pass through the adder unchanged. Complement control 1 causes the exponent to be output in ones complement form, and sign 1 gives a negative sign to the result coefficient. In this case, borrow to 96 indicates that the negative operand was the larger of the two

The resultant leaving output 1 of the 4 FN7 module is exponent positive $1 / 0$ bits 48 through 59 and goes to the destination X register

The result leaving output 2 of the 4 FN 7 module is exponent bits 48 through 59 and goes to the destination X register.

FAD 3.2 TEST POINTS

| Module | Location | Test Point | Description | Module | Location | Test Point | Description | Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & 34 \\ & 36 \\ & 56 \\ & 45 \\ & 46 \\ & 52 \\ & 25 \\ & 74 \\ & 75 \\ & 76 \\ & 73 \\ & 24 \\ & 66 \\ & 61 \\ & 43 \\ & 35 \\ & 71 \\ & \hline 42,63 \\ & 31,53 \\ & 22 \\ & 32 \\ & 33 \\ & 45 \\ & 44 \\ & 65 \\ & 66 \\ & 55 \\ & 54 \\ & 56 \\ & 71 \\ & 74 \\ & 72 \\ & 75 \\ & 73 \\ & 76 \\ & 62 \\ & 61 \\ & \\ & 31 \\ & 46 \\ & \\ & 43,63 \\ & 34,53 \end{aligned}$ |  | 4FK7* <br> 4FL7* <br> 4FN7 | $\begin{aligned} & 5 \mathrm{E} 01-11 \\ & 5 \mathrm{D} 12 \\ & 5 \mathrm{C} 07 \end{aligned}$ | - | (No signal/rgtr TPs) <br> (No signal/rgtr TPs) <br> (No signal/rgtr TPs) |  |  | \% |  |

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PART 6

LONG ADD UNIT

The long add unit executes CPU instructions 36 and 37 . It performs a 60 -bit integer addition to $X$ register operands specified by the $j$ and $k$ portions of the instruction and delivers the 60 -bit sum to the X register specified by the i portion of the instruction. Prior to addition, the unit complements the Xk and Xj operands for the 36 instruction (integer sum) and only the Xj operand for the 37 instruction (integer difference).

The long add instructions require 2 clock periods for execution. The operands move from the X registers to the long add unit in the same clock period in which the instruction issues from the CIW register. The result moves from the long add unit to the destination X register during the following clock period. A new instruction may issue from the CIW register for execution in the long add unit each clock period.

The long add unit complements the operands as specified by the instruction code and forms a partial sum in the first stage of addition. The partial sum enters the second stage of the adder where the result is formed from the partial sum. At this point, the result is in ones complement form. The go long add flag, sent from the CPU, gates this sum to the result $X$ register through a static network which complements it to return it to true form.

Adder Format:


## FIRST STAGE

The first stage of addition forms a partial sum which consists of bit enables, bit borrow generates, group borrow generates, section borrow generates, and section enables. Group enables are also generated but are not sent to the second stage of the adder.
Bit enables $=\frac{1}{0} \quad$ or $\quad \frac{0}{1} \quad 1$

1
Bit borrow generates $=\frac{1}{0}$

Group borrow generates = borrow-out of three-bit group
Example: 110101
$010 \quad \underline{110}$

000 or 011

Group enable $=$ all bit enables in group equal to one Section borrow generates = borrow-out of 12 -bit section Section enable $=$ all group enables in section equal to one

## SECOND STAGE

The second stage of the adder forms group borrow inputs and bit borrow inputs from the partial sum. An exclusive $O R$ of the bit borrow inputs and the bit enables forms the sum. The sum at this point is in ones complement form.

Group borrow inputs $=$ borrow into group
Bit borrow inputs $=$ borrow into bit

Example：

The following example uses a 12 －bit adder performing a 36 instruction．In this case，the section borrow generate is the end－around borrow．In the case of the long add unit，it is the section borrow from the highest－order section．

36 Instruction Example：



| CONTROL DATA | PRIMARY BLOCK DIAGRAM | $\left[\begin{array}{l} \text { Coot IDRAT } \\ 34010 \end{array}\right.$ | $6$ | K'K |
| :---: | :---: | :---: | :---: | :---: |
| DEVELOPMENT DIVISION | LONG ADD UNIT | C |  | -1 |


note :
(1) section i enable goes to 6c02, ec03, 6c04, scos
 SECTION 3 ENABLE GOES TO 6CO1,6CO2,6CO4,6COS section 4 enable goes to 6col, 6c02, 6co3,6cos section 5 enable goes to 6 col, $6 C 02,6003,6 c 04$

## 017012

| CONTROL DATA | SECONDARY <br> LONG ADD | BLOCK | DIAGRAM | $\left\lvert\, \begin{aligned} & \text { coot loent } \\ & 34010 \end{aligned}\right.$ | $6$ | K |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { DEVELOPMENT } \\ & \text { DIVISION } \end{aligned}$ |  |  |  | 0 | \|scti LG ADD 20, | 6-3 |

The long add unit executes CPU instruction 36 and 37 . It performs a 60 -bit integer addition to $X$ register operands specified by the $j$ and $k$ portions of the instruction and delivers the 60 -bit sum to the X register specified by the i portion of the instruction. Prior to addition, the unit complements the Xk and Xj operands for the 36 instruction (integer sum).

The long add instructions require 2 clock periods for execution. The operands move from the X registers to the long add unit in the same clock period in which the instruction issues from the CIW register. The result moves from the long add unit to the destination X register during the following clock period. A new instruction may issue from the CIW register for execution in the long add unit each clock period.

During the first clock period, the 4 KC 7 modules complement both operands. If the instruction is integer difference (37), m bit 0 is a one and the 4 KC 7 modules recomplement the Xk operand. After complementing, the operands enter the first stage of the adder.

The first stage of the adder is divided into two parts. The first part forms the bit enables, bit borrow generates, and group borrow generates. Registers hold them for use during the second clock period. The second part of the first stage of addition takes place during the second clock period. It forms the section borrow generates and section enables.

The 4KC7 modules send the partial sum to the 4KD7 modules where the second stage of addition takes place. Go long add gates the result from the second stage to the result X register.

LG ADD 2.0 TEST POINTS

| Module | Location | Test Point | Description | Module | Location | Test Point | Description | Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\begin{aligned} & 35 \\ & 33 \\ & 31 \\ & 42 \\ & 44 \\ & 46 \\ & 52 \\ & 54 \\ & 56 \\ & 65 \\ & 64 \\ & 61 \\ & 36 \\ & 34 \\ & 41 \\ & 43 \\ & 51 \\ & 53 \\ & 66 \\ & 63 \\ & 32 \\ & 45 \\ & 55 \\ & 72 \\ & 71 \end{aligned}$ $71,72$ | $\begin{aligned} & \text { Bit } \mathrm{N} \text { enable } \\ & \qquad \begin{array}{l} \mathrm{N}+1 \\ \mathrm{~N}+2 \\ \mathrm{~N}+3 \\ \mathrm{~N}+4 \\ \mathrm{~N}+5 \\ \mathrm{~N}+6 \\ \mathrm{~N}+7 \\ \mathrm{~N}+8 \\ \mathrm{~N}+9 \\ \mathrm{~N}+10 \\ \mathrm{~N}+11 \\ \mathrm{~N} \\ \text { Bit borrow } \\ \mathrm{N} \\ \text { N+1 } \\ \mathrm{N}+3 \\ \mathrm{~N}+4 \\ \mathrm{~N}+6 \\ \mathrm{~N}+7 \\ \mathrm{~N}+9 \\ \mathrm{~N}+10 \\ \text { Group } \mathrm{N} \\ \text { borrow } \\ \mathrm{N}+1 \\ \downarrow \end{array} \\ & \begin{array}{l} \text { Section enable } \\ 25 \end{array} \\ & \text { Go long add } \end{aligned}$ | . |  | - |  |  |  |  |  |

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The floating-multiply unit executes the following three instructions.
40 Floating product of ( Xj ) times ( Xk ) to Xi
41 Rounded floating product of ( Xj ) times ( Xk ) to Xi
42 Floating double-precision product of ( Xj ) times ( Xk ) to Xi

The Xi coefficient and Xi exponent are formed separately with consideration taken for single or double precision. If single precision is specified, the upper half of the coefficient result and an exponent (corrected for single precision) are sent to the X register. For double precision, the lower half of the coefficient result and the exponent result are sent to the $X$ register.

If both operands are normalized, the result is normalized. If rounding is specified, a round bit is generated and added to bit 46 of the coefficient result.

The multiply instructions require 5 clock periods for execution. Multiply instructions may enter the multiply unit every other clock period. Thus, during a 5-clock-period segment, more than one multiply instruction may be active in the unit.

## COEFFICIENT MULTIPLY

Coefficient multiply is basically the same as a multiply done with paper and pencil. Every bit of the multiplier is multiplied by every bit of the multiplicand. In binary arithmetic, this is the AND (logical product) function.

As in pencil and paper multiplication, the bit-by-bit products are arranged in a matrix as shown below with each row offset by one bit. The columns of the matrix are then added to obtain the final product, Xi .


The multiply is performed in two steps. First, the lower 24 Xj bits are multiplied by all 48 Xk bits, and the columns of the resulting matrix are partially added. The upper 24 Xj bits are then multiplied by all 48 Xk bits and the resulting matrix is added to the partial sums and carries from the first pass to form the 96 -bit double-precision product. Each step in the process involves forming 1152 binary products. These 1152 bits of data must then be added in the proper groupings to form a combined sum.

The multiply in the previous example appears as follows when executed in two steps.


In a 24 -by 48 -bit multiply, in order to handle 1152 bit products in the matrix, the summing is performed in several more stages; there are several more levels of sums and carries. Because of the size of the matrix, there are also several sums or carries per bit position at some levels of the multiply. Since a complex carry network to resolve all the carries requires excessive time and hardware, the multiply unit employs a carry save adder for efficiency.

Using these inputs, the first half of the adder forms pseudo sums and pseudo carries.

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## CARRY SAVE ADDER

The carry save adder permits the addition of columns of numbers without using the time-consuming carry and satisfy checks typical of full adders. Instead, each level of add has two outputs, pseudo sums and pseudo carries.

(Pseudo carries are displaced one bit to the left because they affect the next significant bit position.)

The sum of these two numbers is the actual answer; however, the sum is not taken until the final add. Instead, both quantities are put back into the next level of the adder and are summed with other pseudo sum and carry bits that represent the same bit position of the final product (that is, sums of bits in the same column of the matrix or carries to that column).

The inputs to the first level add are the logical products produced by the matrix. The inputs to the second level add are the pseudo sums and carries from the first level add. The inputs to the third level add are the pseudo sums and carries from the second level add. The output from the third level add is a pseudo sum and carry bit for each bit position of the product.

After the first pass, the output from the third level add is right-shifted 24 places, fed back into the second level add, and added to the results of the second pass through the matrix. The second and third level adds are then repeated for the entire product until a pseudo sum and carry bit again represent each bit position of the product. This time the output of the third level add is fed to the final add network and is reduced to one bit for each bit position of the final product (96. bits).

Two adders are used for the formation of the final exponent. The first adder is a half adder that forms the sum of the complemented exponents of Xj and Xk . When single precision is selected, this adder adds 60 (octal) to the exponents to compensate for truncating the coefficient of the product in single-precision mode. The second adder subtracts one from the result of the first adder if the coefficient is left-shifted one place to normalize it.

The exponent logic also examines the incoming exponents for special cases involving overflow, underflow, or indefinite operands before performing the additions. Upon receiving the output from the first adder, the exponent logic checks the result for overflow or underflow of the floating-point range, summarizes the special cases, and sets the appropriate special case flags.
$\mathrm{Xj}+\mathrm{Xk}+60$ (OCTAL) ADDER

The $\mathrm{Xj}+\mathrm{Xk}+60$ adder is a static network that sums the two exponents in ones complement mode during the second and third clock periods of instruction execution. For purposes of this discussion, the exponent portion of the multiply operand (bits 48 through 58) is referred to as bits 0 through 10.

The add network for bits 0 through 3 and 6 through 10 is a two-input adder that sums Xj and Xk . Bits 4 and 5 of the add network, however, have three inputs to allow the single-precision correction to be made. The following example illustrates the adder inputs.

| Bit | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: |
|  | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 1 | 1 | 0 | 1 |
| Xj | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 1 | 0 | 0 |
| Correction |  |  |  |  |  | 1 | 1 |  |  |  |  |

Using these inputs, the first half of the adder forms pseudo sums and carries. The second half of the adder uses the pseudo sums and carries to determine enables and carries for three-bit groups. The add network resolves the enables and carries to form the complemented result.

## MINUS ONE NETWORK

The minus one network subtracts one from the output of the first adder by adding one to the complement. This add network takes into account any end carry condition and also makes carry propagation checks. Bit 10 is complemented to add the exponent bias value. The complemented output of this add network (true output) is gated to the $\mathrm{X}^{\prime}$ register when the coefficient has been left-shifted one place to normalize it

An alternate network, with no minus one correction, is used when the coefficient is already normalized or cannot be normalized by a left-shift of one. This alternate network merely forms the exponent bias and complements the output so that the true exponent goes to the X register

## CONTROL

Multiply busy is the primary control signal in the multiply unit. It is set in the CPU at the end of the clock period in which a floating-multiply instruction issues from the CIW register. The flag is copied to other register ranks as data is processed by the unit. The flag blocks input to the multiply unit in the clock period immediately following instruction issue. This flag and copies of it control data movement through the multiply unit.



## INPUT REGISTERS

All instructions performed in the floating-multiply unit require 5 clock periods for execution. Data moves from the operating registers to the multiply coefficient input registers on the 4 GA 7 modules in the same clock period in which a floatingmultiply instruction issues from the CIW register. The input registers are cleared and new data entered whenever the multiply busy flag clears. When the multiply busy flag sets, the data in the input registers is held over into the following clock period. This data is held in the input register for a total of 2 clock periods.

The multiply busy flag serves the purpose of a go multiply flag as well as blocking further entry to the unit in the clock period following issue. The multiply busy flag serves as the basic timing control that gates data into the registers of the unit at the proper time.

As the Xj and Xk coefficients enter the input registers, they are complemented, if negative, so that the multiply coefficient hardware deals only with positive numbers. The 124 GA 7 modules hold 48 Xk bits and 48 Xj bits.

During the first clock period, bits 0 through 47 of the Xk coefficient and bits 0 through 23 of the Xj coefficient enter the input registers of the 4 GA 7 modules. The 4 GB 7 and 4 GC 7 modules use these bits during the second clock period for forming the first half of the product. During the first clock period, bits 24 through 47 of the Xj coefficient enter bit holding registers and are held there until gated into the input registers by the multiply busy signal in the second clock period. Xj bits 0 through 23 are discarded at this time. The 4GB7 and 4GC7 modules use Xk bits 0 through 47 and Xj bits 24 through 47 during the third clock period for forming the second half of the product.

MATRIX AND FIRST LEVEL ADD

The bit-by-bit product of Xk and Xj is formed on the 244 GB 7 and 4 GC 7 modules during the second and third clock periods. This product is formed in two passes through a 24 -bit-by-48-bit multiply matrix.

On the first pass (second clock period), the product of $X k$ and the lower half of the Xj coefficient (bits 0 through 23) is formed. During the second pass (third clock period), the product of Xk and the upper half of Xj (bits 24 through 47) is formed. Each pass through the matrix produces 1152 binary products, which must be added in the proper groupings to form a combined sum.

Several bits are produced that represent the same bit position of the product. (Refer to diagram for 4 GC 7 module located at B06.) For example, Xi product bit 9 is produced from the following combinations of operand bits.

| $\left.\begin{array}{l} X k \text { bit } \tilde{2} \cdot X j \text { bit } 7=X i \text { bit } 9 \\ X k \text { bit } 3 \cdot X j \text { bit } 6=X i \text { bit } 9 \end{array}\right\}$ | $\begin{aligned} \text { First level add }= & \text { pseudo sum bit } 9 \\ & \text { pseudo carry bit } 10 \end{aligned}$ |
| :---: | :---: |
| $\left.\begin{array}{l}\mathrm{Xk} \text { bit } 6 . \mathrm{Xj} \text { bit } 3=\mathrm{Xi} \text { bit } 9 \\ \mathrm{Xk} \text { bit } 7 . \mathrm{Xj} \text { bit } 2=\mathrm{Xi} \text { bit } 9\end{array}\right\}$ | $\begin{aligned} \text { First level add }= & \text { pseudo sum bit } 9 \\ & \text { pseudo carry bit } 10 \end{aligned}$ |

The first level add sums these product bits in groups of two or three bits to produce pseudo sum and carry bits. For example, the first level add of the top two Xi bits 9 produces a pseudo sum bit 9 and a pseudo carry bit 10. Similarly, the sum of the lower two Xi bits 9 produces a pseudo sum bit 9 and a carry bit 10 . The first level add of the six Xi bits 8 produces two pseudo carry bits 9 . Thus, four bits are output by the 4 GC 7 module for Xi bit 9 : two pseduo sum bits and two carry bits.

The results of the first level add ( 788 bits ) are sent to the $4 \mathrm{GD} 7,4 \mathrm{GE} 7,4 \mathrm{GF} 7$, 4GG7, 4GH7, 4GI7, 4GK7, 4GL7, 4GO7, and 4GP7 modules for a second level add. The result of the add of Xk bit 47 and Xj bit 47 (bit 94 ) is sent to the 3 BA 7 module where it is used in the integer multiply check. If bit 94 is a zero and both operands have underflow exponents, an integer multiply is performed.

MULT 3.0 TEST POINTS



The second level add modules receive 788 pseudo sum and pseudo carry bits from the first level adder and continue to sum groups of bits to form pseudo sum and pseudo carry bits for each bit position of the result. The resulting bits are held in registers for 1 clock period.

On the first pass through the second level add, the only bits summed are from the first pass through the 4 GB 7 and 4 GC 7 modules. However, on the second pass through the second level add, loop bits from the first pass through the 4GK7, 4GL7, 4GM7, and 4GN7 modules are merged with the pseudo. sum and pseudo carry bits from the second pass through the 4 GB 7 and 4 GC 7 modules. The hardware for both passes through the second level adder is the same, except that the 4 GJ 7 modules are not used during the first pass. The modules containing logic for the second level add are 4GJ7, 4GI7, 4GD7, 4GE7, 4GF7, 4GG7, 4GH7, 4GK7, 4GL7, 4GO7, and 4GP7. The 4GK7, 4GL7, 4GO7, and 4GP7 modules also contain logic for the third level add.

On the 4 GE 7 and 4 GF 7 modules, reference is made to duplicate bits. These bits are received from fanouts on the 4 GB 7 and 4 GC 7 modules. The duplicate bits provide a means of generating a pseudo carry to bit N from bit $\mathrm{N}-1$. The pseudo sum for bit $N-1$ is generated by the next lower module. Since separate modules generate the pseudo sum and pseudo carry bits, duplication of the input bit is necessary.

## FIRST PASS

During the first pass (second clock period), the 4GD7, 4GE7, 4GF7, 4GG7, 4GH7, and 4GI7 modules receive pseudo sums and carries and sum these bits in groups to form pseudo sums and carries for each bit position of the lower part of the product (bits 0 through 70). The resulting pseudo sums and carries are held in registers for use during the third clock period. In the beginning of the third clock period, add logic on the same modules further sums the pseudo sums and carries before sending them to the $4 \mathrm{GK} 7,4 \mathrm{GL} 7,4 \mathrm{GM} 7,4 \mathrm{GN} 7,4 \mathrm{GO} 7$, and 4 GP 7 modules for the third level add.

The 4GJ7 modules are not used during the first pass
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## SECOND PASS

During the second pass (third clock period), the 4GD7, 4GE7, 4GF7, 4GG7, 4GH7, 4GI7, and 4GJ7 modules merge 141 loop bits (bits 0 through 70) from the first pass with pseudo sums and carries of the upper half of the product (bits 24 through 94). The loop bits come from the 4GK7, 4GL7, 4GM7, 4GN7, 4GO7, and 4GP7 - modules and are right-shifted 24 places so that the lower 24 bits enter the 4GJ7 modules and are properly positioned for the merge. Except for the 4 GJ 7 modules, the pseudo sums and carries that result from the second level add are held in registers for use during the fourth clock period. In the beginning of the fourth clock period, add logic on all except the 4 GJ 7 modules further sums the pseudo sums and carries before sending them to the $4 \mathrm{GK} 7,4 \mathrm{GL} 7,4 \mathrm{GM} 7,4 \mathrm{GN} 7,4 \mathrm{GO}$, and 4GP7 modules for the second pass through the third level add.

The 4GJ7 modules have two loop bits entering for each bit position of the final product. The two loop bits enter registers on the 4 GJ 7 modules during the third clock period and are held there during the fourth clock period when the final add is performed. This part of the final add generates an enable for each of 12 bit positions, a carry for each of 9 bit positions, and a group carry and group enable for each 4 bits of the final product. The output of the 4 GJ 7 modules goes directly to the 4GS7 and 4GT7 modules, where the final add is completed.

## ROUNDING

If rounding is specified by the multiply instruction, two copies of the round bit signal enter at bit 46 of the 4 GE 7 module during the second pass. These two bits generate a carry into bit 47 of the product during the fourth clock period. The round bit signals come from the 3KV7 module of the multiply exponent logic during the third clock period of instruction execution. 5-7-6. 1

MULT 3.1 TEST POINTS



The third level add modules receive pseudo sum and pseudo carries from the second level adder and continue to sum groups of bits to form a pseudo sum and carry for each bit position of the result. The modules used for the third level add are 4GK7, 4GL7, 4GM7, 4GN7, 4GO7, and 4GP7. The 4GK7, 4GL7, 4GO7, and 4GP7 modules contain registers that hold some of the pseudo sum, pseudo carry, and loop bits. All the other pseudo sum and carries received by these modules were held in registers on the 4GD7, 4GE7, 4GF7, 4GG7, 4GH7, and 4GI7 modules.

On the 4GM7, 4GN7, 4GO7, and 4GP7 modules, reference is made to duplicate bits. These bits are received from fanouts on the second level add modules. The duplicate bits provide a means of generating a pseudo carry to bit N from bits $\mathrm{N}-1$ and $\mathrm{N}-2$. The pseudo sums for bit $\mathrm{N}-1$ are generated on the next lower module as bit $N+2$ pseudo sums. Since separate modules generate the pseudo sums and carries, duplication of the input bits is necessary.

FIRST PASS

The first pass through these modules takes place during the third clock period. The third level add further reduces the results of the first pass through the matrix to a pseudo sum and pseudo carry (loop bits) per bit position of the lower part of the product. These 141 bits loop back to the 4GJ7, 4GK7, 4GL7, GG17, 4GD7, 4GF7, 4GG7, and 4GH7 modules to be added with the results of the second pass through the matrix. An enable loop signal from the 3KV7 module enables the loop from the 4GM7, 4GN7, 4GO7, and 4GP7 modules during the third clock period and blocks the loop on all other clock periods.

## SECOND PASS

The second pass through the third level add takes place during the fourth clock period. At this time, the third level add further sums the entire product, the results of both passes through the matrix, until only two bits of data remain in each bit position of the upper part of the double-precision sum (bits 24 through 95). These 141 pseudo sum and carry bits are delivered to the 4GS7, 4GR7, and GQ7 modules for the final add.

MULT 3.2 TEST POINTS


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## FINAL ADD - LOWER HALF

The final add is performed in two parts. The first half of the final add sums the pseudo sums and pseudo carries and sends the resulting carries and enables to the second half of the final add. The second half of the final add uses a standard pass and carry check method of carry propagation. Both halves of the final add constitute a full adder. A full adder is not used in the earlier stages of the multiply unit because of the extra time involved in checking enables and carries.

The first half of the final add is located on the 4GJ7, 4GS7, 4GR7, and 4GQ7 modules. The output of the 4GJ7 modules goes to the two 4GT7 modules, where the second half of the final add for bits 0 through 23 takes place during the fourth clock period.

During the fourth clock period, the 4GS7 module generates enables for bits 24 through 31 by performing an exclusive $O R$ of the pseudo sum and carry bits input for each bit position. These enables are held in registers during the fifth clock period when they are used by the $4 G V 7$ modules for the second part of the final add.

The 4GS7 module generates carries for bits 25 through 31 during the fourth clock period by performing an AND of the pseudo sum and carry bits input for each bit position. The enable and carry bits for bits 24 through 31 of the final product go to a 4 GV 7 module.

During the fifth clock period, the 4GS7 module uses the bit enables and carries to form a section carry and a section enable. The eight bits of the final product that are handled by each module constitute a section. The section carry and section enable generated are fanned out to the $4 G U 7,4 G V 7,4 G W 7$, and $4 G X 7$ modules of the coefficient hardware and to the 4KU7 module of the exponent hardware for coefficient left-shift determination.

MULT 3. 3 TEST POINTS



## FIRST HALF

The first half of the final add is located on the 4GJ7, 4GS7, 4GR7, and 4GQ7 modules. During the fourth clock period, the $4 G R 7$ and $4 G Q 7$ modules generate enables for bits 32 through 94 by performing an exclusive $O R$ of the pseudo sum and pseudo carry bits input for each bit position. These enables are held in registers during the fifth clock period when they are used by the $4 G W 7$ and 4 GX 7 modules for the second part of the final add.

The 4GR7 and 4GQ7 modules generate carries for bits 33 through 95 during the fourth clock period by performing an AND of the pseudo sum and carry bits input for each bit position. The enable and carry bits for bits 32 through 95 of the final product go to the $4 \mathrm{GV} 7,4 \mathrm{GW} 7$, and $4 \mathrm{GX7}$ modules.

During the fifth clock period, the 4GR7 and 4GQ7 modules use the bit enables and carries to form a section carry and a section enable. The eight bits of the final product that are handled by each module constitute a section. The section carry and section enable generated are fanned out to the 4GU7, 4GV7, 4GW7, and 4GX7 modules of the coefficient hardware and to the 4 KU 7 module of the exponent hardware for coefficient left-shift determination.

The section carry generated by the 4GQ7 module includes special circuitry that prevents the final result from being left-shifted one place when neither of the source operands was normalized. One of the two bits entering the 4GQ7 module for bit 94 is the pseudo sum for bit 94 . This bit is the result of the logical product of bit 47 of the two source operands. Neither the 4 GC 7 module that formed the logical product nor the 4GP7 module has modified this product
other than to call it pseudo sum 94. Assuming that the two source operands were normalized, this bit should be a one, since it would then be the logical product of two one bits. If it is not a one, the two source operands were not normalized. The 4 GQ 7 module uses the complement of pseudo sum 94 to set the section carry bit and thereby prevents left-shifting the final result when the two source operands are not normalized.

## SECOND HALF

The second half of the final add is performed on the $4 G T 7,4 G U 7,4 G V 7,4 G W 7$, and 4GX7 modules. The 4 GW 7 and 4 GX 7 modules form the final product of bits 48 through 95. The input of these modules is an enable and a carry for each bit position and an enable and a carry for each section. The result is the complement of the final sum and is complemented again as it is gated to the X registers by the gate upper half signal. This recomplements the final product so that the X registers receive the positive value of the result. If the result is negative, it is complemented by the $X$ register input control (CPU) before it enters the destination X register.

## LEFT SHIFT NETWORK

Before gating the final result, the $4 G U 7$ and $4 G V 7$ modules determine whether to left-shift the result by one to normalize it. If bit 95 is a one, shifting is not performed. Bit 95 is a one when the multiply process has generated a carry to it or it is forced to a one when one or both of the multiply source operands are normalized. If bit 95 is a zero, the result is not normalized and the hardware left-shifts it one to normalize it. Bit 95 is a zero only when both source operands are not normalized and there was no carry into bit 95.

MULT 3.4 TEST POINTS

| Module | Location | Test Point | Description | Module | Location | Test Point | Description | Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 4GQ7* <br> 4GR7* | 7H13 $\square$ | 06 43 05 56 16 66 15,75 22 23 26 24 32 33 $36,35,76$ 21 25,55 31,65 54 54 06 42,43 05 56 16 66 15 15 71 22 23 26 24 32 33 36 34 |  |  |  | $\begin{aligned} & 21 \\ & 25,55 \\ & 31,65 \\ & 36,76 \\ & 54 \\ & 04,05 \\ & 04,05 \end{aligned}$ | Final add bit N and $\mathrm{N}+1$ enable <br> Final add bit $\mathrm{N}+2$ and $\mathrm{N}+3$ enable <br> Final add bit $\mathrm{N}+4$ and $\mathrm{N}+5$ enable <br> Final add bit N+6 and $\mathrm{N}+7$ enable 25 -nanosecond clock <br> Gate upper half <br> Gate upper half |  |  |  |  |

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## EXPONENT ARITHMETIC

## EXPONENT INPUT REGISTERS

The exponent input registers, located on the 4 KR 7 module, receive bits 48 through 59 of the multiply operands from the 4 RE 7 module during the first clock period. If the coefficient sign (bit 59) is negative, the associated exponent is complemented to obtain the true value. Bit 59 is then removed and is sent to the 3 BA 7 module. The two resulting 11 -bit exponents are sent to the 4 KS 7 module. From this point on, bits 48 through 58 are referred to as bits 0 through 10.

## ADD NETWORK

During the second clock period, the exponent logic on the 4 KS 7 module submits the complement of bits 0 through 9 of both exponents to the add network. Exponent bias is removed by not complementing bit 10 .

The add network sums the two exponents in a 13 -bit ones complement mode. Depending upon the instruction mode, this network also adds a third quantity. If the multiply double-precision flag is set, this third quantity is a zero. If the multiply double-precision flag is cleared, the third quantity is 60 (octal) (+48 decimal).

## First Half

The first half of the add network forms pseudo sums for bits 0 through 3 and 6 through 10 by performing an equivalence of the two exponents. For these bits, a pseudo sum is generated when the corresponding bits of the two exponents are equal.

Bits 4 and 5 are handled by a three-input adder to allow the single-precision correction to be made. The add network performs an exclusive OR of the exponents for bits 4 and 5. Then, depending upon whether double precison is selected or not, it periforms an exclusive OR of the result and the double-precision correction signal received from the 3 KV7 module during the second clock period.

The 4KS7 module also generates pseudo carries. The add network forms pseudo carries to bits 1 through 4 and 7 through 11 by performing an OR of the corresponding bits of the two exponents. Pseudo carries to bits 5 and 6 are handled by a three-input network to allow the single-precision correction to be made. When the pseudo sums and pseudo carries are half-added, the result is 60 (octal) added to the exponent if single precision is selected.

## Second Half

The 3BA7 module combines the pseudo sums and carries from the 4 KS 7 module to determine which stages are enables and which stages have carries coming into them.

A bit enable is produced by an exclusive OR of the pseudo sum and pseudo carry coming into that stage. Bit and group carries are produced by an AND of the pseudo sum and pseudo carry into a stage or an AND of an enable and carry generated by the next lower stage.

The add network uses the resulting bit enables, bit carries, and group carries to perform carry propagation checks until each bit of the sum is represented by an enable and a carry. An exclusive $O R$ of these two bits produces the sum in complemented form. This complemented result is fed to the $4 \mathrm{KU7}$ module.

## PRODUCT SIGN

The 3BA7 module generates the product sign by performing an exclusive $O R$ of the sign bits received from the $4 K R 7$ module.

## SPECIAL CASE CHECKS

The 4KS7 module receives the true values of the two operand exponents from the 4 KR 7 module during the second clock period and examines them for overflow, underflow, or indefinite conditions. Overflow of the floating-point range is indicated by an operand exponent value of 3777 in packed form, the largest exponent value that can be represented in floating-point format. Underflow of the floatingpoint range is indicated by 0000 in packed form, the smallest exponent value that can be represented in floating-point format. An indefinite condition is indicated by a minus zero exponent, 1777 in packed form. The exponent logic looks for each of these conditions and sends an overflow, underflow, or indefinite condition indicator for each operand exponent to the 3 BA 7 module.

The 3 BA 7 module summarizes the special cases and examines the final result for overflow or underflow of the floating-point range

## NDEFINITE

An indefinite signal is sent to the 3 KV 7 module when one of the following conditions exists.

One or both of the operands have an indefinite exponent.
One operand has an underflow exponent and the other operand has an overflow exponent

## OVERFLOW

An overflow signal is sent to the 3KV7 module when one of the following conditions exists.

One operand has an overflow exponent and the other operand has a normal exponent.

Both operands have overflow exponents.
The unpacked exponent of the result (sensed prior to reducing it by one when a left-shift is performed) is greater than +1777 (octal).

## UNDERFLOW

An underflow signal is sent to the 3KV7 module when one of the following conditions exists.

One operand has an underflow exponent and the other operand has a normal exponent.

Both operands have underflow exponents.
The unpacked exponent of the result (sensed prior to reducing it by one when a left-shift is performed) is less than -1776 (octal).

## PECIAL CASE

A special case signal is sent to the 3 KV 7 module if any of the previous special conditions exist.

## NDEFINITE-1

An indefinite-1 signal is sent to the 3KV7 module if one or both operands have an indefinite exponent. The indefinite bit is set in the exit condition register (CPU) for this case.

## INFINITE

An infinite signal is sent to the 3KV7 module if one or both operands have an overflow exponent. The infinite bit is set in the exit condition register (CPU) or this case.

## NTEGER MULTIPLY

If both operands have underflow exponents and the operand coefficients are not both normalized, an integer multiply is performed.

NOTE
If an integer multiply is performed with one coefficient normalized, an undetected overflow result may occur.

An integer multiply forces the second half and result bits 0,4 , and 5 to ones and forces bit 10 (bias bit) to a zero. This ensures that all zero bits are sent to the Xi register for the exponent portion of the result. The complement of the exponent portion is sent to the Xi register if the sign of the result is negative.

## RESULT REGISTER

The exponent result register on the 4 KU 7 module receives the complemented result of the exponent add from the 3BA7 module during the fourth clock period The result is used by two separate output networks.

One output network subtracts one from the output of the first adder by adding one to the complement. This network, called the minus one network, gates the exponent to the X register when the coefficient result has been left-shifted one place to normalize it. The other network, which does not have a minus one correction, gates the exponent when the coefficient result is not shifted. Both output networks complement bit 10 to add the exponent bias and then complement the entire output to form the true output.

The 4KU7 module determines whether or not the coefficient was shifted by checking for a carry into bit 95 using the section carries and enables from the 4GT7, 4GS7, 4GR7, and 4GQ7 modules. If no carry was generated into bit 95 (forced or otherwise), the coefficient was shifted and the network with the minus one correction is gated out.

Gating of the exponent from either network occurs upon receipt of the gate exponent signal from the 3KV7 module during the fifth clock period.

## MULTIPLY CONTROL LOGIC

The multiply control logic for both the exponent and coefficient arithmetic of the floating-multiply unit is located on the 3KV7 module. The multiply busy flag, which is received from the 4 LE 7 module, controls data movement through the multiply unit. The multiply busy flag on the 4 LE 7 module sets at the end of the first clock period.

Consequently, multiply busy on the 3KV7 module sets during the second clock period and is ciear on ali others. Similarly, the multiply busy signal on the 3KV7 module is a one during the first clock period and a zero during the second clock period. A timing chain and the multiply busy signal control the output of following signals.

## ROUND FLAG

Two copies of the round flag cause the addition of a round bit to bit 49 of the double-precision coefficient during the third clock period of instruction execution. The round flag sets during the first clock period of execution when rounding is specified by the multiply instruction. This flag is copied to another register rank for use during the third clock period when two copies of the flag are sent to coefficient bit 45 on a 4 GE 7 module. A carry bit, generated by these two bits, rounds bit 46 of the coefficient.


The DP correction flag sets during the first clock period when double precision is specified by the multiply instruction and sent immediately to the 4KS7 module to control the exponent arithmetic. This flag is also copied to other register ranks on the 3 KV 7 module until the fifth clock period when it gates the lower half of the double-precison coefficient. The complement of this signal gates the upper half of the double-precision coefficient.

## ENABLE LOOP

The enable loop signal, conditioned by multiply busy, gates the loop bits from the first pass through the coefficient logic during the third clock period. Looping is blocked at all other times to prevent the possibility of bits from the second pass through the matrix from interfering with the execution of a new multiply instruction that may have entered the multiply unit.

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MULT 3.5 TEST POINTS



The 3BA7 module forms the second half of the exponent add and examines the complement of the result for overflow or underflow of the floating-point range. The 3BA7 module also determines whether an integer multiply should be performed.

The overflow and underflow tests are made primarily by examining bits 10, 11, and 12 of the result. Since the result is in complement form at this point of the add operation, bit 12 (term H12) set indicates a positive exponent result, and term 7 H set indicates a negative exponent result. Bit 11 may be set for a positive result, but should never be set when the result is negative. The value of bit 10 depends upon whether the exponent result is negative or postive.

## OVERFLOW

When bits 10 and 11 of a positive result are set (terms H 10 and H11), the result is $\leq+1777$. Term 7 I indicates the complement of this condition or that the result is $>+1777$. Therefore, an overflow condition (term 7I set) occurs when the result is $>+1777$. The overflow signal is sent to the 3 KV 7 module if an indefinite condition does not already exist.

## UNDERFLOW

When the result is negative (term 7 H set), two possible overflow conditions are examined. Bit 10 (term H 10 ) set indicates that overflow has occurred because the exponent result is $\leq-2000$. Terms JA, JB, and JC set indicate that the result is equal to - 1777. If either of these conditions exists, an underflow condition (term UF) is detected. The underflow signal is sent to the 3 KV 7 module if an indefinite condition does not already exist.

## INTEGER MULTIPLY

When both operands have underflow exponents (terms $K U$ and $J U$ are ones) and the operand coefficients are not both normalized (term BI is one), an integer multiply is performed. However, if one of the operand coefficients is normalized, an undetected overflow result normally occurs.

When an integer multiply is performed, exponent adder output bits 0,4 , and 5 are forced to one by term IM, and the bias bit (bit 10) is forced to a zero. This ensures that all zero bits are sent to the Xi register for the exponent portion of the integer result.

During an integer multiply operation, an underflow condition exists, but the special case condition is blocked by term NIM. Blocking special case to multiply control (3KV7 module) enables the multiply unit to output to the Xi register even though the underflow condition exists.




## PART 8

BIVIDE. UNIT

The floating-divide unit executes the two CPU instructions, floating divide (44) and round floating divide (45). These instructions direct the computer to divide Xj by the divisor Xk and send the quotient to Xi . This unit involves a 17 -step iterative process to form the quotient from the two operands. Only one divide instruction may be executed in the iterative portion of the divide unit at a given time.

The divide instructions require 20 clock periods for execution. Data moves from the operating registers to the divide unit input registers each clock period in which the divide busy flag is clear. The data is used for. instruction execution only if a divide instruction issues from the CIW register and sets the divide busy flag. The data which arrives at the divide unit during the clock period of instruction issue is then used in the execution of the following divide sequence. The divide busy flag prevents the CIW from issuing another divide instruction for the 17 clock periods following instruction issue. However, in the 18 th clock period after instruction issue, a second divide instruction may issue.

Bit 0 of the $m$ designator is held in an input register along with the operand data in Xj and Xk . This bit is the divide round flag which distinguishes between the two instruction modes.

The divide unit operates on positive coefficient values only. Each coefficient for Xj and Xk is individually complemented in static networks if its sign is negative. The sign of the result is determined by the divide unit and is sent to X register input control in the CPU. This sign bit is the logical difference of the two operand sign bits.

The divide busy flag initiates a chain of divide sequence control flags which sequence the steps in the instruction execution. This chain controls the sequence of events for the 19 clock periods following the issue of the divide instruction. These static conditions then control the data movement within the divide unit and the data transmission of the X register input path at the end of the divide sequence.

The format of a floating-point number is $\mathrm{k} 2^{\mathrm{n}}$, where k is a 48 -bit integer coefficient and $n$ is a 10 -bit integer exponent.

Division of floating-point numbers requires the subtraction of the exponents and the division of two 48-bit coefficients. Double-precision division is not provided and a remainder cannot be retrieved. Since the divide hardware produces a quotient in the range, 1.7777777777777777 through 0.0000000000000000 , the ratio of Xj to Xk must always be less than 2 to 1 . If the divisor is normalized, this requirement is always met. If this requisite is not met, the resulting quotient is meaningless. If both the dividend Xj and the divisor Xk are normalized, the quotient Xi is also normalized.

The quotient coefficient is formed three bits per clock period in 17 iterative steps. The result of the first iteration is stored in the control module. The results of the 16 succeeding iterations (a total of 48 bits) are stored in the quotient shift register. If the quotient is of the form, $0 . \mathrm{X}------\mathrm{X}$ (the ratio of Xj to Xk is less than 1 to 1 ), the result of the first iteration is a zero, and the 48 bits in the quotient shift register are taken as the coefficient of the result.

$$
\text { Given: } X i=X j / X k
$$

$(X j)=20574400000000000015$
$(\mathrm{Xk})=20326000000000000000$
The divide coefficient logic forms:

## $\frac{4400000000000015}{6 \theta 0 \theta 0 \theta 0000 \theta 00000}=0.6000000000000021$

In floating-point format, the coefficient must be an integer. Therefore, the binary point must be moved 48 places to the right. Since the exponent must be decreased by one for every place, the binary point is right-shifted, 60 (octal) is subtracted from the difference of the exponents, and the final result is:
$\left(\mathrm{Xi}_{\mathrm{i}}\right)=17446000000000000021$
[Final exponent $=2057-2032-60=-33$ (unbiased) $=1744]$


If the quotient is of the form, $1 . \mathrm{X}-------\mathrm{X}$ (the ratio of Xj to Xk is 1 to 1 or greater, but less than 2 to 1 ), the result of the first iteration is a one, and the upper 47 bits in the shift register are interpreted as bits 0 through 46 of the quotient coefficient. The $X$ register access control interprets bit 47 as a one.

$$
\text { Given: } \mathrm{Xi}=\mathrm{Xj} / \mathrm{Xk}
$$

$(\mathrm{Xj})=20167000000000000000$
$(X k)=2025400000000000000$

The divide logic forms:

## $\frac{7000000000000000}{4000000000000000}=1.6000000000000000$

In order to make this quantity an integer, the binary point must be moved 47 places to the right. Therefore, 57 (octal) is subtracted from the difference of the exponents and the final result is:
$(\mathrm{Xi})=17117000000000000000$
[Final exponent $=2016-2025-57=-66$ (unbiased) $=1711$ ]

If the ratio of Xj to Xk is 2 to 1 or greater, the result of the first iteration is forced to a two by the release remainder control signal. The control network then sends a special case signal (indefinite) to the $X$ register input control network along with an operand consisting of all ones. The X register input control network then generates an indefinite result.

The quotient coefficient is formed three bits per clock period. The iteration network functions exactly like a pencil and paper octal divide. A multiplication network forms seven multiples of the divisor, Xk through 7 Xk . The dividend is entered into the remainder register, which holds 51 bits (initially bits 48, 49, and 50 are all 0 ). The trial subtraction network simultaneously compares each of the seven divisor multiples with the contents of the remainder register. The largest multiple that is smaller than the remainder is subtracted from the remainder. The resulting quantity is left-shifted three binary (one octal) positions and entered into the remainder register. The number of the multiple chosen (pick number) becomes the first quotient digit. The pencil and paper method is


Divisor (Xk) $=6000000000000000$ 2000000000000000

The multiplication network forms:
$\mathrm{Xk}=6000000000000000$
$2 \mathrm{Xk}=14000000000000000$
$3 \mathrm{Xk}=220000000$ 0000 0000
$4 \mathrm{Xk}=30000000000000000$
$5 \mathrm{Xk}=36000000000000000$
$6 \mathrm{Xk}=44000000000000000$
$7 \mathrm{Xk}=52000000000000000$
The dividend enters the remainder register
RMDR $=04400000000000015$
Initially, the quotient shift register has contents: XXXX XXXX XXXX XXXX The 17 iterative operations are performed as follows:

1. Compare RMDR versus Xk through 7Xk:

RMDR is smaller thian any of them.
Enter octal digit 0 in shift register:
Quotient $=\mathrm{XXXX} \mathrm{XXXX}$ XXXX XXXO (the quotient
is left-shifted one octal digit each clock period). Enter chosen digit (in this example, a 0 ) in control module this iteration only.
Enter RMDR-0 and left-shift three: RMDR $=44000000000000150 \sim$ becomes a 2525 pattern if instruction 45 enter 0 since left-shift hree

Compare RMDR versus Xk through 7 Xk : $6 \mathrm{Xk} \leq \mathrm{RMDR}<\mathrm{T} \mathrm{Xk}$ Enter second octal digit: QUOT = XXXX XXXX XXXX XX06

Enter RMDR-6Xk and left-shift three: RMDR = 00000000000001500
3 through 14 In the next 12 iterations, an octal digit 0 is picked, and the quotient and remainder are left-shifted three binary places each clock period.

RMDR $=01500000000000000$ QUOT $=$ XX06 000000000000
15. Compare RMDR versus Xk through 7Xk: RMDR < Xk

Enter octal digit 0: QUOT $=$ X060 000000000000 Enter RMDR-0: RMDR = 15000000000000000

Compare RMDR versus Xk through 7 Xk : $2 \mathrm{Xk} \leq \mathrm{RMDR}<3 \mathrm{Xk}$
Enter octal digit 2: QUOT $=0600000000000002$ Enter RMDR-2Xk: RMDR $=10000000000000000$ Compare RMDR versus Xk through 7 Xk : $\mathrm{Xk} \leq \operatorname{RMDR}<2 \mathrm{Xk}$ Enter octal digit 1: QUOT $=6000000000000021$ Discard remainder and transmit quotient to Xi




## EXPONENT MANIPULATION

## INPUT REGISTERS

The data input registers for both operand exponents, bits 48 through 59, are on the 4KR7 module. The data that is in the registers when the divide busy flag sets is held for the 17 clock periods during which the flag remains set. Xj bit 59 and Xk bit 59 are sensed and the exponents (bits 48 through 58) are individually complemented if they are negative. The output goes to the 4 DO 7 module (bits 48 through 58) and the 3DP7 module (bit 59).

## EXPONENT FORMATION

The exponent subtraction occurs while the coefficient is being formed. The quotient exponent is held until the quotient coefficient is completely assembled. The exponent subtraction network, comprised of the preliminary and final addition networks, is on the 4DO7 and 3DP7 modules. Instead of subtracting the divisor exponent and the correction factor from the dividend exponent, the dividend exponent is complemented and added to the divisor exponent and to the correction factor (CF). The CF is determined in CP04 by bit 0 (overflow bit) of the first quotient coefficient digit. The resulting sum (exponent) is complemented on output by the 4DS7 module to obtain the quotient exponent. In effect, the exponent network forms minus ( $-\mathrm{Xj}+\mathrm{Xk}+\mathrm{CF}$ ) instead of ( $\mathrm{Xj}-\mathrm{Xk}-\mathrm{CF}$ ).

The exponent addition is performed in two halves. The network that performs the preliminary exponent addition is on the 4 DO 7 module; the final exponent addition network is on the 3DP7 module. The bias is removed from both exponents on input to the 4 DO 7 module. The exponents and the correction factor are then added. If the overflow bit from the 4DQ7 module, bit 0 of the first iteration, is a zero, the quotient is of the form $0 . \mathrm{XXXX} \mathrm{XXXX} \mathrm{XXXX} \mathrm{XXXX}$, and the correction factor is 60 (octal). If bit 0 of the first iteration is a one, the quotient is of the form 1. XXXX XXXX XXXX XXXX , and the correction factor is 57 (octal).

The result of the first half addition is a bit carry and a bit enable for each position. These bits are transferred to the final exponent addition network on the 3DP7 module.

The final addition network combines the carries and enables in a 13-bit mode. The two extra bits are formed by sign-extending the exponents. The upper three bits of the 13 -bit sum are sensed to detect special case conditions. Exponent bias is added, and the lower 11 bits of the sum are then transferred to the output network on the $4 \mathrm{DS7}$ module.

The sign of the quotient is formed by a network on the 3DP7 module and sent to the X register input control. This sign bit is the logical difference of the two operand sign bits. A network on the 4 DO 7 module checks for operand exponent overflow, underflow, and indefinite exponents. The rest of the special case checks are performed on the $3 D P 7$ module. If any special case exists, the special case signal to the 4 DQ 7 module is a one, in which case, a quotient of all ones is sent to the $X$ register. The three signals to the $X$ register input control specify which, if any, of the special case conditions exist.

## SPECIAL CASE

A number of special cases are treated in the floating-divide unit. If any special case conditions exist, the special case signal from 3 DP 7 to the 4 DQ 7 module becomes a one. When the go divide special case signal from the 4 DQ 7 module to the X register input control becomes a one, the specific condition is indicated by the exponent indefinite, overflow, and underflow signal from the 3DP7 module. In any special case condition, the transmission of data from the divide unit output registers to the destination $X$ register is blocked. The gate quotient signal from 4DQ7 to 4DS7 remains a zero, forcing the exponent output registers to transmit all ones. The gate output signal (which is the gate quotient signal delayed 1 clock period) forces the coefficient output registers to transmit all ones. Thus, a quotient of all ones is delivered to the $X$ register and the input control forms the particular special case word. The special condition flags cause the appropriate bit to be set in the exit condition register.

If no special case conditions exist, the special case signal from 3DP7 to 4DQ7 remains a zero. The go divide special case signal to the X register input control is a zero and the gate divide and the gate quotient signals are ones. The gate quotient signal gates the quotient output.

## SPECIAL CASE OPERANDS

One category of special case conditions involves overflow, underflow, or indefinite operand values. These situations are sensed in the 4DO7 module. The combination of special operand values which cause specific results are sensed on the 3DP7 module. If either operand is indefinite, or if both operands are indefinite, the result is indefinite. The operand coefficients are ignored in this case, and the resulting word delivered to the Xi register is positive indefinite with a zero coefficient. The indefinite bit is set in the exit condition register (CPU) for this case.

If either operand has an overflow exponent, or if both operands have overflow exponents, the result is infinite. The infinite bit is set in the exit condition register (CPU) for this case.

If Xj has an overflow exponent and Xk is in floating-point range or has an underflow exponent, the result is a complete overflow word delivered to the Xi register. The coefficients of the operands are ignored in this case, and the result is a zero coefficient. The sign of the result is calculated in the same manner as for operands in range.

If Xj has an underflow exponent and Xk is in floating-point range or has an overflow exponent, the result is a complete underflow word delivered to the Xi register. The coefficients of the operands are ignored in this case, and the result is a zero word.

If Xk has an overflow exponent and Xj is in floating-point range, the result is a complete underflow word delivered to the Xi register. The coefficients of the operand are ignored in this case, and the result is a zero word.

If Xk has an underflow exponent and Xj is in floating-point range, the result is a complete overflow word delivered to the $X i$ register. The coefficients of the operands are ignored in this case. The sign of the result is calculated in the same manner as for operands in range.

The combination of operand exponents of overflow divided by overflow and underflow divided by underflow results in a positive indefinite word delivered to the Xi register.

## SPECIAL CASE QUOTIENT

A second category of special cases occurs if there is an underflow or an overflow of the floating-point exponent range during the exponent calculation. In these cases, the special case signal is sent to the $X$ register input control, and the output from the divide unit is blocked in the same manner as for the special case operands

A complete overflow occurs for this instruction whenever the exponent computation results in an exponent greater than plus 1777 (unbiased). If any combination of operand exponents causes an indefinite condition, the overflow situation is ignored. Otherwise, this situation is sensed as a special case, and a complete overflow word with proper sign is delivered to the Xi register. The coefficient calculation is ignored in this case.

A complete underflow occurs for this instruction whenever the exponent computation results in an exponent less than minus 1777 (unbiased). If any combination of operand exponents causes an indefinite condition, this underflow situation is ignored. Otherwise, this situation is sensed as a special case, and a complete zero word is delivered to the Xi register. The coefficient calculation is ignored in this case.

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## SPECIAL CASE: DIVIDE FAULT

A third special case category occurs if the dividend coefficient is larger than the divisor by a factor of two or more. This is a divide fault situation which can occur if the divisor is not normalized. The initial trial subtraction in the coefficient calculation results in an octal digit with a value of two. If an overflow or underflow condition does not exist, an indefinite condition result is indicated for this case, and it is treated in the same manner as the other special cases.

If an overflow or an underflow condition caused by an underflow or infinite operand does exist, the divide fault situation is ignored.

## PARTIAL OVERFLOW OR UNDERFLOW

A partial overflow occurs for this instruction whenever the exponent computation results in exactly plus 1777 (unbiased). The result is delivered to the Xi register in a normal manner. Subsequent use of this result as an operand in a floating-point unit may, however, result in overflow detection.

A partial underflow occurs for this instruction whenever the exponent computation results in exactly minus 1777 (unbiased). The result is delivered to the Xi register in a normal manner. Subsequent use of this result as an operand in a floating-point unit may, however, result in underflow detection.

## OUTPUT

The exponent output register is on the 4DS7 module. The complemented 11-bit biased quotient exponent is delivered from the 3DP7 final addition network. In CP18, a gate quotient signal from the 4 DQ 7 module enables a clock gate so that the exponent enters the output register. The exponent is complemented on output because the exponent network forms the complement of the quotient exponent.

Since a positive quotient is delivered to the X register, bit 59 is entered as a zero. The absence of a gate quotient signal in CP18 indicates that one of the special case conditions exists, in which case, an exponent of 12 ones is delivered to the X register.

The quotient coefficient is assembled three bits per clock period in the 48-bit shift register. In each of CP03 through CP19, an octal digit is delivered from 4DD7 to 4DS7. After 17 iterations, the 4DS7 module has bits 0 through 2 and 45 through 47; the two 4DR7 modules have bits 3 through 44. Every clock period, the quotient is left-shifted three places, three new quotient bits from 4DD7 enter from the iteration network, and the upper three bits are discarded. After 17 iterations, the shift register contains 48 bits of data. Bits 0 and 1 of the octal digit formed in the first iteration are stored in the $3 \mathrm{DQ7}$ module. If bit 1 is a one, a divide fault condition (the dividend exceeds the divisor by a factor of two or more) exists. If this condition or any other special case condition exists, the gate quotient signal remains a zero and the delivery of the quotient is blocked. As a result, a coefficient of all ones is in the path to the X register in CP19. If bit 1 is a zero and no other special case conditions exist, the gate quotient signal becomes a one in CP18 and the quotient coefficient is delivered to the $X$ register during CP19.

If bit 0 of the first iteration digit is a zero, the quotient is of the form $0 . \mathrm{XXXX}$ XXXX XXXX XXXX. In this case, a correction factor of 48 is subtracted from the exponent, the overflow bit is a zero, and the 48 bits in the shift register are delivered to the X register.

If bit 0 of the first iteration digit is a one, the quotient is of the form $1 . \mathrm{XXXX}$ XXXX XXXX XXXX. In this case, a correction factor of 47 is subtracted from the exponent. The overflow bit becomes a one, causing a one-bit right-shift of the coefficient output. The upper 47 bits in the shift register are delivered to the X register as bits 0 through 46 . In this case, the bit 47 output is blocked. Therefore, bit 47 is entered as a one. All outputs are timed to occur in CP19:

## CONTROL

The divide sequence control of the $4 D Q 7$ module times the entire divide instruction. The clock periods on the left of the divide sequence control indicate the set times. The times on the right are the clear times.

The three remainder bits are part of the coefficient iteration network. They are used as rank 0 bit enables in the 4DD7 and 4DE7 modules.

The special case signal indicates a special case exponent. The gate divide and go divide signals go to the $X$ register input control to gate in the appropriate data. The gate quotient signal causes the quotient to be delivered during CP19.

The begin sequence signal initiates the divide timing sequence. The divide fault signal is bit 1 of the first coefficient iteration digit. This bit is sent to the 3DP7 module where it forces a special case condition if it is a one.

The overflow bit is bit 0 of the first coefficient iteration digit. This bit goes to the 4DO7 module where it determines the correction factor. It also goes to the 4 DR 7 and 4 DS 7 modules where it determines which 48 bits of the coefficient are delivered to the $X$ register.

The divide round flag ( $m$ bit 0 ) is set when a round floating divide instruction issues from the CIW register. This flag modifies the dividend in the third clock period of instruction execution. Octal digits with a value of 25252 are entered in the lowest order bits of the remainder register if this flag is set. This modification of the dividend increases the dividend value by one-third of the least significant bit in the original operand.

The circulate remainder signal goes to the four 4DI7 modules where it gates the current, remainder to the final subtraction network. When this signal is a zero, the content of the dividend input register is gated to the final subtraction network.

The release remainder signal goes to the 4DD7 and 4DE7 modules. It allows the iteration network to generate a digit of three through seven. When this signal is a zero, no digit greater than a two can be generated by the iteration network. This signal is a one on the 4DQ7 module from CP02 through CP17.

The divide 15 (T15) condition originates in the divide sequence control and is used in instruction issue control and in the $X$ register access control. This condition exists during the 16 th clock period of execution for a divide instruction. It is used in the instruction issue control to block issue of an instruction which would conflict with the delivery of data from the divide unit to the X register data input path. It is used in the $X$ register access control to initiate the process of register access for the destination X register.

The clear divide busy condition originates in the divide sequence control and is used to clear the divide busy flag. This condition exists during the 18 th clock period of execution for a divide instruction.

DIV 3.0 TEST POINTS


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## DIVISOR MULTIPLICATION NETWORK

The divisor (Xk) coefficient input registers are on the four 4D.J7 modules. Each module handles 12 bits. Every clock period, a 48-bit quantity from the X register arrives at the modules. The sign bit (bit 59) is delivered along with the coefficient. The registers are cleared and new data is entered every clock period in which the enter Xk control signal is a one. When the divide instruction issues in CP00, enter Xk is a one. The signal becomes a zero in CP01 and remains a zero until CP18. The data that arrives at the input register in the clock period of instruction issue ( CPOO ) is entered into the register at the end of CP00 and held until the end of CP18 when a new quantity is gated into the register. This data is used in executing the divide instruction.

The divide unit uses the coefficient magnitude of the operands in executing the divide instructions. The coefficient is complemented (4DJ7 module) if it is negative. If bit 5 is a one, the operand is negative.

The Xk coefficient magnitude is sent to the multiplication network. The 4DK7, 4DL7, 4DM7, and 4DN7 modules form the seven multiples of the divisor coefficient from Xk to 7 Xk .

## 3Xk ADDER

A network on the 4 DK 7 and 4DN7 modules creates a 50 -bit quantity 3 Xk and shifts this quantity to form 6 Xk . Bits 0 through 47 of 3 Xk are created on the four 4 DK 7 modules, 12 bits to a module. The 12 bits on a 4 DK 7 module are called a section and are referred to as bits $N$ through $N+11$. Bits 48 and 49 are created in the 4 DN 7 module.

Each 4 DK 7 module receives 13 bits of Xk from the divisor input registers on the 4DJ7 modules. These bits are used as both Xk and 2 Xk . 2 Xk is formed by left-shifting Xk one bit position. The 13 Xk bits received are bits $\mathrm{N}-1$ through $\mathrm{N}+11$. Of these bits, $\mathrm{N}-1$ through $\mathrm{N}+10$ are used as 2 Xk bits N through $\mathrm{N}+11$. Note that $\mathrm{N}+11$ in one section is $\mathrm{N}-1$ in the next higher section. For example:

| N+11 |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Xk: | 6 |  |  | 5 |
| input: | 110 | 010 | 0 | $\underline{101}$ |
| 2Xk: | 4 | 4 | 1 | 3 |

The four 4 DK 7 modules form 3 Xk by adding bits 0 through 47 of 2 Xk to bits 0 through 47 of Xk . A preliminary adder on each module generates section carries and enables.

The 4DN7 module receives the four section carries and three section enables from the 4 DK 7 modules and performs carry propagation checks on each section. The carryout of section 4 (to bit 48) is added to Xk bit 47 , which left shifted one place is 2 Xk bit 48 . The sum of this bit and the section 4 carry bit produces bits 48 and 49 of 3 Xk .

The other three section carries are fed back into the final add network on the 4DK7 modules to produce 3 Xk bits 0 through 47.

## 6Xk NETWORK

6 Xk is formed from 3 Xk by left-shifting one bit position. The multiples are sent to partial subtraction and trial subtraction networks. Each module in the two subtraction networks handles a three-bit grop. Therefore, each module needs three bits of 3 Xk and three bits of 6 Xk . The multiplication network sends four 3 Xk bits to each module. These bits are labeled bits $\mathrm{N}-1$ through $\mathrm{N}+2$. Bits $\mathrm{N}-1$ through $\mathrm{N}+1$ of 3 Xk are used by the 4 DA 7 and 4 DH 7 modules as bits N through $\mathrm{N}+2$ of 6 Xk . This is equivalent to left-shifting one bit or multiplying by two.

## 5Xk ADDER

The network that forms 5 Xk is on the $4 \mathrm{DL7}$ and 4 DN 7 modules. By left-shifting Xk two places, 4 Xk is formed, 5 Xk is formed by adding Xk to this quantity. Thus, 5 Xk is a 51 -bit quantity. The 4DL7 modules generate bits 0 through 47; the 4 DN 7 module generates bits 48 through 50 . Each 4DL7 module receives 14 Xk bits and outputs 125 Xk bits. The 14 Xk bits are left shifted two places to form bits N to $\mathrm{N}+11$ or 4 Xk .


Used on next higher module as $\mathrm{N}-1$ and $\mathrm{N}-2$.

Bits $\mathrm{N}-2$ through $\mathrm{N}+9$ are used as bits N through $\mathrm{N}+11$ of 4 Xk . These bits are added to bits N through $\mathrm{N}+11$ to form 5 Xk . Thus, 5 Xk is formed $\mathrm{b}_{\mathrm{j}}$ adding 4 Xk to Xk . The section carry and enable bits are used in exactly the same manner as in forming 3 Xk . The 4 DN 7 module is also used in the same manner as in forming 3 Xk .

## 7Xk SUBTRACTER

The subtraction network that forms 7 Xk is on the 4 DM 7 and 4 DN 7 modules. First, 10 Xk is formed by left-shifting Xk three places. 7 Xk is then formed by subtracting Xk from this quantity. The 4 DM 7 modules generate bits 0 through 47; the 4DN7 module generates bits 48 through 50. Each 4DM7 module receives 15 bits of Xk and outputs 12 bits of 7 Xk . The 15 Xk bits are leftshifted three places to form bits N through $\mathrm{N}+11$ of 4 Xk .

| Xk: | ${ }_{2}^{\mathrm{N}+}$ | 3 |  | 5 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | ${ }^{2}$ |  |  |  |  |
| Xk input: | 010 | 011 | 100 | 101 | $\underline{001}$ |
| 8 Xk : |  | 3 | 4 | 5 | 1 |

Used on next higher module as $\mathrm{N}-1$ through $\mathrm{N}-3$.

Bits $\mathrm{N}-3$ through $\mathrm{N}+8$ of Xk are left-shifted to become bits N through $\mathrm{N}+11$ of 10 Xk . Bits N through $\mathrm{N}+11$ of Xk are then subtracted from these bits. Thus, 7 Xk is formed by subtracting Xk from 10 Xk . The section borrows and enables are used in the 4 DN 7 module in the same manner as the section carries and enables are used in forming 3 Xk and 5 Xk .

## $\mathrm{Xk}, 2 \mathrm{Xk}$, and 4 Xk NETWORKS

The networks that form $\mathrm{Xk}, 2 \mathrm{Xk}$, and 4 Xk are on the $4 \mathrm{DK} 7,4 \mathrm{DL} 7$, and 4 DM 7 modules.- These multiples are formed by left-shifting the 48 Xk bits. By leftshifting Xk one bit position, 2 Xk is formed and has 49 bits. By left-shifting Xk two bit positions, 4 Xk is formed and has 50 bits. These three multiples are supplied to the trial subtraction network and the partial subtraction network.

DIV 3.1 TEST POINTS



## ITERATION NETWORK

## TRLAL SUBTRACTION NETWORK

The trial subtraction network determines which of the multiples of the divisor are larger than the current remainder. The trial subtraction network also generates a borrow into every three-bit group. The trial subtraction occurs in two stages.

## FIRST-STAGE TRIAL SUBTRACTION NETWORK

The first-stage trial subtraction network generates a borrow and an enable for each three-bit group. The first-stage network is on the 4DA7, 4DB7, 4DC7, 4DD7, and 4DE7 modules. Each module in this network handles three bits (one octal digit). Each of the seven multiples of the divisor is subtracted from the remainder. The rank of a subtraction refers to the multiple being subtracted. Thus, the rank 5 subtraction network subtracts 5 Xk from the remainder. The 4DC7 module handles group 0 (bits 0 through 2) of all seven ranks. The 154 DA 7 modules handle groups 1 through 15 (bits 3 through 47) of ranks 3, 6, and 7. The 4DD7 module handles group 16 (bits 48 through 50) of ranks 3, 6, and 7. The 154 DB 7 modules handle groups 1 through 15 (bits 3 through 47) of ranks $1,2,4$, and 5 . The 4DE7 module handles group 16 (bits 48 through 50 ) of ranks $1,2,4$, and 5 .

The data for the first-stage subtraction comes from two sources. The remainder is held within each module from the final subtraction on the previous iteration.

Each module in the first-stage trial subtraction network handles a three-bit group (one octal digit). The remainder digit is compared with each of the multiples on a module. The 4 DC 7 module does not generate any enables since no borrow can be propagated through bits 0 through 2 .

The control signals into the 4DD7 and 4DE7 modules can force the rank 1 through 7 group 16 borrow-out bits to be ones. If these borrow bits are ones, the end-around barrow (EAB) bits for the corresponding ranks are ones, and rank 0 is selected for the final subtraction. The release remainder signal
from the 4DQ7 module into the 4DD7 and 4DE7 modules forces the rank 3 through 7 group 16 borrow-out bits to ones. The release rank 1 and 2 borrow signal into the 4 DE 7 module forces the rank 1 and 2 borrow-out bits to ones. These signals are used in CP01 and CP02.

In CP01, both signals are zeros; the rank 1 through 7 group 16 borrow-out bits are all ones; the rank 1 through 7 EAB bits are all ones; the final subtraction network selects rank 0 for the final subtraction; the dividend is selected by the 4DI7 modules as the rank 0 bit enables; and the dividend is gated into the remainder register.

In CP02, the 4DD7 and 4DE7 modules receive the release remainder signal from the 4DQ7 module. The 4DD7 and 4DE7 modules complement the release remainder signal and use it to force the rank 3 through 7 group 16 borrow-out bits to ones during CP01 and CP02. This causes the rank 3 through 7 EAB bits to be ones, and the final subtraction network selects one of ranks 0,1 , or 2 for the final subtraction. Thus, if the dividend exceeds the divisor by a factor of two or more, the 4 DD 7 module generates a two as the first quotient digit.

## SECOND-STAGE TRIAL SUBTRACTION NETWORK

The outputs of the first-stage trial subtraction network go to the second-stage trial subtraction network on the 4 DG 7 and 4 DF 7 modules. The second-stage trial subtraction network determines which multiples of divisor are larger than the remainder and generates borrows into all groups for every rank. Each of the seven 4DF7 modules handles a different one of the seven ranks. The same is true of the seven 4DG7 modules. One 4DF7 module and one 4 DG 7 module together handle one rank.

The 4DF7 and 4DG7 modules receive identical inputs. Each module in the secondstage network receives 17 borrows, one from each of 17 groups, and 16 enables, one from each of 16 groups. All the inputs to one module pertain to the same rank.


The second-stage network determines which multiples are larger than the remainder. If the remainder is smaller than the multiple, that rank subtraction generates an EAB. Each 4DF7 and 4DG7 module generates an EAB bit for its rank. If the $E A B$ bit for a particular rank is a one, the remainder is smaller than the multiple. If this EAB bit is a zero, the remainder is greater than or equal to the multiple. These bits are sensed in the final subtraction network to determine which multiple to use for the final subtraction. The 4DF7 EAB bit goes to every 4DA7 and 4DD7 module. The 4DG7 EAB bit goes to every 4DB7 and 4DE7 module. The EAB bit generated by the 4DF7 module is a duplicate of the EAB bit generated by the 4DG7 module. This duplication is simply an additional means of fanning out the EAB bit to all the destination modules.

The second-stage network ( 4 DF 7 and $4 \mathrm{DG7}$ modules) generates borrows into groups 3 through 16 for each rank. Each 4DF7 module generates borrows into groups 3, 5, 7, 9, 11, 13, and 15 for its rank. Each 4DG7 module generates borrows into groups $4,6,8,10,12,14$, and 16 for its rank. The $4 D F 7$ and 4DG7 modules generate the group borrow inputs for each group in the normal fashion. However, in transmitting the group borrow inputs to the destination 4DA7, 4DB7, 4DD7, and 4DE7 modules, the group borrow inputs are leftshifted one group to align them properly with the inclusive bit enables from the 4 DH 7 modules. Thus, instead of generating a group borrow into group 8, the following conditions generate a borrow into group 9: group 7 generates a borrow-out; group 6 generates a borrow-out; group 7 generates an enable; any of groups 0 through 5 generate a borrow; and all the higher groups (up to and including group 7) generate an enable.

## PARTIAL SUBTRACTION NETWORK

The partial subtraction network provides the inclusive bit enables used in the final subtraction. The partial subtraction network is on the 164 DH 7 modules. Bits 0 through 47 are divided into 16 three-bit groups (that is, into 16 octal digits) with one digit per module.

The partial subtraction network receives the complemented remainder from the remainder register and seven multiples of the remainder from the multiplication network. The subtraction network combines the remainder bits and the multiple bits to generate inclusive bit enables. Each 4 DH 7 module adds the remainder digit to each of the seven multiples.

The outputs of the partial subtraction network go to the final subtraction network. All the outputs from the partial subtraction network are left-shifted one octal digit (that is, three binary bits). The $\overline{\text { remainder (rank } 0 \text { inclusive bit enables) }}$ is recomplemented to generate the true remainder bits. Remainder bits 45 through 57 are sent to the $4 \mathrm{DQ7}$ module, and bits 0 through 44 are sent to four 4 DI 7 modules. Because of the three-bit left shift, these bits are labeled bits 48 through 50 at the 4DQ7 module and bits 3 through 47 at the 4 DI 7 modules. The results of the computation, seven octal digits on each module, are sent to the final subtraction network on the 4DA7 and 4DB7 modules. The seven octal digits from 4DH7O16 are sent to the 4DD7 and 4DE7 modules. Because of the left shift, these are bits 3 through 50 enables.

The remainder digits (rank 0 inclusive bit enables) go through an intermediate network on the 4DI7 and 4DQ7 modules before they are transmitted to the final subtraction network. The $4 \mathrm{DQ7}$ module gates bits 48 through 50 to the final subtraction network in the 4DD7 and 4DE7 modules in clock periods CP02 through CP17 only. The circulate remainder signal in the 4 DI 7 modules chooses either the remainder bits 3 through 47 or the quantity in the dividend register.

In CP01, the dividend from the divide sequence is in the dividend register. In CP01, therefore, the complemented dividend is gated to the final subtraction network. During CP02 through CP17, the complemented remainder is gated to the final subtraction network. In CP02 through CP17, bits 0 through 2 are always zeros with one exception; a 25252 pattern is entered in CP02 if the instruction is a round floating divide. All output bits (bits 0 through 47 from the 4DI7 modules and bits 48 through 50 from the $4 \mathrm{DQ7}$ module) are complemented to become rank 0 inclusive bit enables before transmission to the final subtraction network.

## NOTE

When a divide round instruction is executed, bits
0 through 2 are alternately entered with either a 2 or a 5 . This effectively shifts a $25252 . .5$ pattern into the bit positions below the least signif-
icant bit of the dividend.

During a round floating divide instruction, a round bit is added to the dividend which has the effect of increasing the dividend by one-third count. The effect this has on the quotient varies depending upon the value of the divisor and the truncation point in the quotient. If the dividend is smaller than the divisor, the quotient is truncated one bit position lower than if the dividend is equal to, or larger than, the divisor. These effects cause the rounding to vary in the quotient from a value of $1 / 6$ of the least significant bit in the result to almost one. The average rounding bias over the entire range of coefficient values is zero.

When a divide instruction is executed, the m0 (round) bit is held in the 4DQ7 (control) module. If the instruction is a round floating divide instruction, this bit is a one. If the instruction is a floating divide instruction, this bit is a zero. The 4DQ7 module holds this round bit until CP02, when it is transmitted to the 4DI7-P10 module. This bit then becomes bits 0 through 2 of the rank 0 inclusive bit enables. Since bits 0 through 2 of the rank 0 bit enables are normally zero, the round bit has an effect only on round instructions (when the round bit is a one). The 4 DC 7 module enters the rank 0 bit 0 through 2 enables into the remainder register every clock period.

## FINAL SUBTRACTION NETWORK

The final subtraction network picks one rank, completes that rank subtraction, and gates the result into the remainder register. An octal digit (pick number) corresponding to the chosen rank is gated into the quotient shift register. The remainder is then available to the first-stage trial subtraction network and the partial subtraction network.

The final subtraction network is on the $4 \mathrm{DA} 7,4 \mathrm{DB7}, 4 \mathrm{DD} 7$, and 4 DE 7 modules. The 4DA7 and 4DB7 modules have identical final subtraction networks, as do the $4 \mathrm{DD7}$ and $4 \mathrm{DE7}$ modules. The 4DA7 modules output remainder bits to the partial subtraction network. The 4DD7 module generates the pick number. Each module in this network handles a three-bit group. The 154 DA 7 and the 60420300 K

15 4DB7 modules handle groups 1 through 15, bits 3 through 47. The 4DD7 and 4DE7 modules handle group 16 , bits 48 through 50 . Group 0 , bits 0 through 2 are handled in the 4 DC 7 module and will be discussed later.

The inputs to the subtraction network come from the second-stage trial subtraction network and the partial subtraction network. Each module in the final subtraction network receives seven EAB bits, one for each rank. Each 4DA7 or 4DD7 module receives one EAB from each of the seven 4DF7 modules. Each 4DB7 or 4DE7 module receives one EAB from each of the seven $4 D G 7$ modules. Each module in the final network receives a group borrow input for each of ranks 1 through 7 from the second-stage trial subtraction network. 4DA7-M02 and 4DB7-N02 receive group 0 borrows (left-shifted one group) from the $4 \mathrm{DC7}$ module. All other modules in this network receive inputs from a 4 DF 7 or $4 \mathrm{DG7}$ module, depending upon the group handled. Each module in the network receives three enable bits for each of the eight ranks. The ranks 1 through 7 enables come from the 4DG7 modules, and the rank 0 enables come from the 4DI7 modules. The group 16 (bits 48 through 50) rank 0 enables come from the $4 \mathrm{DQ7}$ module.

The data is combined to generate inree remainder bits for each module. The EAB bits are sensed, and the largest rank that did not produce EAB is used for the final subtraction. The borrow bit for that rank is added to the octal digit and the sum is complemented. The result is entered into the remainder register.

A network on the 4DD7 module senses the EAB bits and generates a pick number. The digit generated is the same as the rank chosen for the final subtraction. This digit is entered into a register and sent from there to the coefficient shift register on the 4DS7 module.

The 4 DC 7 module holds group 0 , bits 0 through 2 of the remainder. In CP01, bits 0 through 2 of the dividend are entered into the remainder register. In CP02 through CP17, zero bits are entered into bits 0 through 2. During a round floating divide instruction, a 25252 pattern is entered in bits 0 through 2 in CP02 through CP17. These bits increase the value of the dividend by one-third count.

Bits 0 through 47 of the remainder are sent to the partial subtraction network. These bits are transmitted from the 4 DC 7 and 4 DA 7 modules to the 4 DH 7 modules.

DIV 3.2 TEST POINTS

| Module | Location | Test Point | Description | Module | Location | Test Point | Description | Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 7M01-15 | $\begin{aligned} & 33 \\ & 32 \\ & 34 \\ & 04 \\ & 03 \\ & 11 \\ & 35 \end{aligned}$ |  |  |  |  |  |  |  |  |  |
|  | 7N01-15 | $\begin{aligned} & 33 \\ & 32 . \\ & 34 \\ & 04 \\ & 03 \\ & 14 \\ & 13 \\ & 35 \end{aligned}$ | $\left\|\begin{array}{l} \text { Rank } 1 \\ 2 \\ 4 \\ 5 \end{array}\right\|$ |  |  |  |  |  |  |  |  |
|  | $\left.\right\|_{\downarrow} ^{7 L 01}$ | $\begin{aligned} & 63,74 \\ & 65,73 \\ & 62,72 \\ & 12 \\ & 71 \end{aligned}$ | $\begin{array}{r} \text { Remainder bit } 0 \\ \downarrow \end{array} \frac{1}{2}$ |  |  |  |  |  |  |  |  |
|  | 7M16 | $\begin{aligned} & 13 \\ & 12 \\ & 11 \\ & 04 \\ & 05 \\ & 14 \\ & 31 \end{aligned}$ |  |  |  |  |  |  |  | , |  |
|  | $\overbrace{i}^{2 N 16}$ | $\begin{aligned} & 05 \\ & 06 \\ & 16 \\ & 15 \\ & 01 \\ & 35 \end{aligned}$ | $\left.\begin{gathered} \text { Rank } 1 \\ \hline \end{gathered} \begin{gathered} \text { enable } \\ 4 \\ 5 \end{gathered} \right\rvert\,$ |  |  |  |  |  |  |  |  |
|  | $\left.\right\|_{\square} ^{7 \mathrm{~L} 06-12}$ | $\begin{aligned} & 02 \\ & 11 \\ & 21 \\ & 36 \\ & 56,46 \end{aligned}$ | $\begin{aligned} & \text { Group } 4 \text { enable } \\ & \left\|\begin{array}{r} 8 \\ \\ \downarrow \\ 12 \\ 16 \end{array}\right\| \\ & \text { Group } 16 \text { borrow } \end{aligned}$ |  |  |  |  |  |  |  |  |
|  | 7006-12 | $\begin{aligned} & 01 \\ & 11 \\ & 21 \\ & 31 \\ & 56,46 \\ & 75 \end{aligned}$ |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |




## PART 10A

## INCREMENT UNIT <br> BLOCK DIAGRAMS

(Models 175, 740, 750, 760)


## PART 9

POPULATION COUNT UNIT

The population count unit executes CPU instruction 47 by counting the number of one bits in the Xk register and storing the results in the lower order six bits of the Xi register. Bits 6 through 59 are filled with zeros.

The population count instruction requires 2 clock periods for execution. Data moves from the $X k$ register to the population count unit in the same clock period in which the instruction issues from the CIW register. Data moves from the population count unit to the Xi register during the following clock period. A new instruction may be issued for execution in the population count unit each clock period.

The content of the Xk register is transmitted to the population count unit each clock period. This data enters a static network that partially sums the one bits and enters the partially reduced data into a 27 -bit holding register. This register is cleared and reset with new data each clock period. The data is further summed in a second static network until six bits represent the complement of the result. When go pop count is received from CPU instruction control, the six-bit result is recomplemented and sent to the lower order six bits of the Xi register, and 54 zero bits are sent to the rest of the Xi register. If go pop count is not received, the data in the population count unit is discarded.



The population count unit counts the number of one bits in the Xk register and stores the results in the lower order six bits of the Xi register. The population count unit can be considered to add a column of 60 bits (bits 0 through 59 of the operand). Instead of attempting to sum all 60 bits in the column at one time, the population count unit divides the operand into three six-bit sections and six seven-bit sections. During the first clock period of operation, the first stage of the adder generates two partial sum bits and a partial carry bit for each of these nine sections. The resulting 27 bits are held in a holding register during the second clock period while the second stage of the adder combines the results of all nine sections into one six-bit result.

When go pop count is received from the CPU, this six-bit result is sent to the X register as bits 0 through 5. Go pop count is also complemented and fanned out to extend zero bits in Xi bits 6 through 59

## FIRST-STAGE ADD

The first-stage add is located on three 4 KA 7 modules. Each module handles 20 bits of the operand. These 20 bits are divided into three sections. Sections 1 and 2 are 7 -bit sections that handle the lower order 14 bits of the 20 bits on the module. Section 3 is a six-bit section that handles the higher order six bits on the module.

The first-stage add network for each section consists of four adders. Initially, six bits in each section enter two three-bit adders ( X adders) and are partially added to form two partial sum bits (sum 0 ) and two partial carry bits (carry 1). (The numbers represent the position each partial sum or partial carry bit has relative to the final result.)

In sections 1 and 2, the two sum 0 bits are added to the seventh bit in the section to produce another partial sum (sum 0 ) and a partial carry bit (carry 1). Section 3 merely totals the two sum 0 bits. The resulting sum 0 bit in each section enters the bit holding register.

There are now three carry 1 bits in each section. They enter the last adder in the stage ( B adder) and are summed to form partial sum bit (sum 1) and a partial carry bit (carry 2). The sum 1 and the carry 2 bits from each stage also enter the bit holding register. There are now three sum 0 bits, three sum 1 bits, and three carry 2 bits in the bit holding register on each module.

## SECOND-STAGE ADD

The second-stage add begins on the 4KA7 modules but is primarily located on the 4 KB 7 module. The add network on each 4 KA 7 module combines the results from the three sections in separate adders. The $A$ adder sums the three sum 0 bits, the $C$ adder sums the three sum 1 bits, and the $D$ adder sums the three carry 2 bits. The sums and carries produced are the complement of the true result and are recomplemented before being sent to the 4 KB 7 module. The data sent to the $4 \mathrm{KB7}$ module from each 4KA7 module, therefore, consists of a sum 0 , sum 1, sum 2, carry 1, carry 2, and carry 3 bit.

The 4KB7 module receives this data from the three 4 KA 7 modules and combines it in a complex add and carry propagation network. The resulting six-bit sum is the complement of the true result. The result is gated by go pop count and is recomplemented before being sent to the X register as bits 0 through 5 . Go pop count is also complemented and fanned out to extend zero bits in Xi bits 6 througl 59.

Go pop count is set in the population count unit when a 47 instruction issues from the CIW register.

POP CT 3.0 TEST POINTS



The three 4 KA 7 modules form the first stage of the add network in the population count unit and hold the results of this add in a 27 -bit holding register. These 27 bits are further reduced to 18 bits before being sent to the second stage of the add in the 4KB7 module.

Each 4KA7 module operates on 20 bits of the 60 -bit operand. Within a module, the 20 bits are divided into three sections. Sections 1 and 2 contain seven bits each, and section 3 contains six bits.

The six or seven operand bits for each section enter a group of three three-bit adders called X adders. The first two X adders in each section add six operand bits to produce two sum 0 bits and two carry 1 bits. In sections 1 and 2, the third $X$ adder sums two of the sum 0 bits just generated (for example, AO and A1) with the seventh perand bit (X06) to produce a sum 0 bit (A2) and another carry 1 bit (B2). In section , the third $X$ adder sums the two sum 0 bits ( $A 6$ and $A 7$ ) from the first two adders to generate a sum 0 bit (A8) and a carry 1 bit (B8).

The three sum 0 bits (A2, A5, and A8) resulting from the three sections are held in the bit holding register. At this point, the $X$ adders have also generated three carry 1 bits for each section. These three carry 1 bits enter a B adder and are reduced to a sum 1 and a carry 2 bit for each section, which are also held in the bit holding register. There are now three sum 0 bits, three sum 1 bits, and three carry 2 bits in the bit holding register for each 4KA7 module.

The sum 0 , sum 1 , and carry 2 bits held in the bit holding registers for each section form an octal number that represents the number of one bits summed in that section.

From the holding register, the three sum 0 bits enter an $A$ adder that sums them to produce a $\overline{\text { sum } 0}$ bit and a carry 1 bit. The three sum 1 bits enter a $C$ adder to become a sum 1 bit and a carry 2 bit, and the three carry 2 bits enter a $D$ adder to become a sum 2 and a carry 3 bit. These bits are all complemented and sent to the 4 KB 7 module.


## PART 10

INCREMENT UNIT

The increment unit executes CPU instructions 50 through 77. These instructions involve arithmetic operations on two selected 18 -bit operands from the $\mathrm{A}, \mathrm{B}, \mathrm{X}$, and CIW registers. During 50 through 57 instructions, the result is transmitted to an A register. The result plus RAC is also sent to the SAS when the A1 through A7 registers are used. These two arithmetic operations are performed independently and in parallel with each other. During 60 through 67 instructions, the result is transmitted to a B register. During 70 through 77 instructions, the result is transmitted to an $X$ register.

## INPUT OPERAND SELECTION

Data arrives at the increment unit from five different input paths. One group of data paths consists of inputs from the $\mathrm{Aj}, \mathrm{Bj}$, and Xj registers. One of these operands is selected as one of the two increment operands. The selection is based on the value of the $m$ designator from the CIW register. The second increment operand is selected from another group of input data paths. This group consists of the $K$ field in the CIW register, the $B k$ register, and the complement of the Bk register. This selection is also based on the m designator value. The two selected operands enter both parallel computation sections simultaneously.

## COMPUTATION

One computation is performed in a two-operand adder. The two selected operands are partially added in the first stage of the adder. The resultant bit/group borrows and enables are stored in the partial sum holding register. The computation is completed in the second stage of the adder.

A second computation is performed in a three-operand adder. This happens simultaneously with the operation described previously. The two selected operands and bits 0 through 17 from the RAC register are partially added in the first stage of the adder. The resultant bit/group borrows and enables are stored in the partial sum holding register. The computation is completed in the second stage of the adder.

## INCREMENT TEST HOLDING REGISTER

An 18-bit incremented operand is sent to the increment test holding register during every clock period. The content of this register is compared with FLC in the CPU. When the incremented operand is equal to or greater than FLC, CMC blocks the CM write reference and returns a zero word for a CM read reference. This occurs only when a 50 through 57 instruction causes a CM reference.

## DESTINATION REGISTER SELECTION

The 18 -bit incremented operand is gated to an $A, B$, or $X$ register whenever the CPU generates a go increment signal. The destination register selection is based on the value of the lower two bits of the $f$ designator from the CIW register.

The Ai register receives the incremented operand during 50 through 57 instructions (f designator bits equal X01). When these instructions occur, the CPU generates a go increment to storage signal. This signal gates the incremented operand (plus RAC) to the SAS. The Ai register and the SAS receive increment data simultaneously. If the $i$ designator in the CIW register has an octal value of zero, the go increment to storage signal is not generated.

The Bi register receives the incremented operand during 60 through 67 instructions (f designator bits equal X 10 ).

The Xi register receives the incremented operand during 70 through 77 instructions (f designator bits equal X11). When these instructions occur, the sign bit is extended in the Xi register.



The increment unit executes CPU instructions 50 through 77. These instructions involve arithmetic operations on two selected 18 -bit operands from the $\mathrm{A}, \mathrm{B}, \mathrm{X}$, and CIW registers. During 50 through 57 instructions, the result is transmitted to an A register. The result plus RAC is also sent to the SAS. These two arithmetic operations are performed independently and in parallel with each other. During 60 through 67 instructions, the result is transmitted to a B register. During 70 through 77 instructions, the result is transmitted to an $X$ register.

## INPUT OPERAND SELECTION

Data arrives at the increment unit ( $4 \mathrm{KE7}$ and $4 \mathrm{KF7}$ modules) from five different input paths. One group of data paths consists of inputs from the $\mathrm{Aj}, \mathrm{Bj}$, and Xj registers. These 18 -bit operands are complemented after entering the 4 KE 7 and 4KF7 modules. One of these operands is selected as one of the two increment operands. The selection is based on the value of the $m$ designator from the CIW register. The second increment operand is selected from another group of input data paths. This group consists of the complement of the K field in the CIW register, the Bk register data, and the complement of the Bk register data. This selection is also based on the $m$ designator value. The two selected operands enter both parallel computation sections simultaneously.

## COMPUTATION

One computation is performed on the $4 \mathrm{KE} 7,4 \mathrm{KF7}, 4 \mathrm{KG} 7$, and 4 KH 7 modules. The two selected operands are partially added in the first stage of the adder. The resultant bit/group borrows and enables are stored in the partial sum holding register ( $4 \mathrm{KE7}$ and 4 KF 7 modules). The computation is completed in the second stage of the adder ( 4 KG 7 and 4 KH 7 modules).

A second computation is performed simultaneously with the one described previously. This computation is performed on the $4 \mathrm{KI7}$ and $4 \mathrm{KJ7}$ modules. The two selected operands and bits 0 through 17 from the RAC register are partially added in the first stage of the adder. The resultant bit/group borrows and
enables are stored in the partial sum holding register (4KI7 modules). The computation is completed in the second stage of the adder (3KJ7 module).

## INCREMENT TEST HOLDING REGISTER

An 18-bit incremented operand is sent to the increment test holding register ( 4 KK 7 module) every clock period. The content of this register is compared with FLC in the 4 LH 7 module. When the incremented operand is equal to or greater than FLC, CMC blocks the CM write reference and returns a zero word for a CM read reference. This occurs only when a 50 through 57 instruction causes a CM reference.

## DESTINATION REGISTER SELECTION

The 18 -bit incremented operand is gated to an $\mathrm{A}, \mathrm{B}$, or X register whenever the 4LE7 module generates a go increment signal. The destination register selection is based on the value of the lower two bits of the $f$ designator from the CIW register. The f designator bits are fanned out in the 4 KF 7 module to a translator in the 4 KG 7 and 4 KH 7 modules.

The $\mathrm{Ai}_{\mathrm{i}}$ register receives the incremented operand during 50 through 57 instructions (f designator bits equal X01). When these instructions occur, the 4 HG 7 module generates a go increment to storage signal. This signal gates the incremented operand (plus RAC) to the SAS. The SAS can accept an address at a maximum rate of one every 50 nanoseconds. The Ai register and the SAS receive increment data simultaneously. If the $i$ designator in the CIW register has an octal value of zero, the go increment to storage is not generated.

The Bi register receives the incremented operand during 60 through 67 instructions (f designator bits equal X 10 ).

The Xi register receives the incremented operand during 70 through 77 instructions ( $f$ designator bits equal X11). When these instruction occur, the sign bit is extended in the Xi register. A fanout ( 4 KH 7 module) extends the sign of bits 17 to bit positions 18 through 59 .

INCR 3. OA TEST POINTS




INCREMENT UNIT
block diagrams
(Models 865, 875)

The increment unit executes CPU instructions 50 through 77. These instruction involve adds or subtracts of two selected 18-bit operands.from the A, B, X, and CIW registers. The increment unit sends the result to an A register ( 50 through 57 instructions), a B register ( 60 through 67 instructions), or an $X$ register ( 70 through 77 instructions). During 50 through 57 instructions which involve a memory reference, the increment unit also sends the result plus the 22-bit RAC to the SAS in CMC

## INPUT OPERAND SELECTIO

Operand 1 and 2 select networks determine two operands for computation based on the of the $m$ designator in the CIW register. The two selected operands enter the increment and increment plus RAC adders simultaneously.

## INCREMENT (18-BIT) ADDER

The 18-bit increment adder produces a one's complement sum or difference of the selected operands. The first stage generates bit/group borrows and enables stored by the partial sum holding register. The second stage completes the computation.

## INCREMENT PLUS RAC (22-BIT) ADDER

The 22-bit increment plus RAC adder produces a one's complement sum or difference of the 18-bit selected operands and a two's complement sum of the result plus the 22-bit RAC. The first stage generates group borrows and enables, partial sums and carries, and bit enables stored by the partial sum holding register. The second stage completes the computation.

The second stage monitors the 18 -bit adder for end-around carry. If the sum or difference does not produce an end-around carry, a correction factor is added to the 22-bit result. Correction is required because the 18 -bit adder can produce an end-around carry, which should be a plus one added to the least significant bit (bit 0 ) and not passed from bit 17 to 18 of the 22 -bit adder. (A carry that passes from bit 17 to 18 with no correction causes the resulting address to be high by 007777777 .) The second stage produces the correction factor by blocking the end-around carry from the 18 -bit adder and adding plus 1 to complemented bits 18
through 21.

## INCREMENT FIELD LENGTH TEST

The increment unit sends I bits 0 through 17 to a field length tester in the CPU every clock period. When this increment operand is equal to or greater than FLC, CMC blocks the CM write reference or returns a zero word for a $a M$ read reference. This occurs only when a 50 through 57 instruction causes a CM reference.

## DESTINATION REGISTER SELECTION

The CPU gates the increment operand to a destination register by generating a go increment signal. The lower 2 bits of the $f$ designator from the CIW register
determine whether an Ai ( 50 through 57 instructions), Bi ( 60 through 67 instructions), or Xi ( 70 through 77 instructions) register is selected. During 50 through 57 instructions with 1 not equal to zero, the CPU also gates the increment operand plus RAC to the SAS in CMC by generating a go increment to storage-1 signal.


| CONTROL DATA | SECONDARY BLOCK DIAGRAM INCREMENT UNIT | $\left\lvert\, \begin{gathered} \text { EOOT IOERTT } \\ 34010 \end{gathered}\right.$ | $\begin{aligned} & \text { put no } \\ & 60420300 \end{aligned}$ | $Y$ |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \text { DEVELOPMENT } \\ \text { DIVISION } \end{gathered}$ |  | 6 | $\begin{aligned} & \text { HitI } \\ & \text { INCR } 2.08 \frac{1}{5} \\ & 5 \end{aligned}$ |  |

The increment unit executes CPU instructions 50 through 77. These instructions involve add or subtracts of two selected 18-bit operands from the A, B, X, and CIW registers. The increment unit sends the result to an A register ( 50 through 57 instructions), a B register ( 60 through 67 instructions), or an $X$ register ( 70 through 77 instructions). During 50 through 57 instructions which involve a memory reference, the increment unit also sends the result plus the 22 -bit RAC to the SAS in CMC. -

## INPUT OPERAND SELECTION

The 4KE7 and 4KF7 modules determine two operands for computation based on the value of the designator in the CIW register. The two selected operands enter the increment and increment plus RAC adders simultaneously.

## INCREMENT (18-BIT) ADDER

The 18 -bit increment adder produces a one's complement sum or difference of the 18 -bit selected operands and a two's complement sum of the result plus the $22-b i t$ RAC. The firs stage (4iding and ing broup borrows an enables the partia

## INCREMENT PLUS RAC (22-BIT) ADDER

The 22-bit increment plus RAC adder produces a one s complement sum or difference of the 18 -bit selected operands and a two's complement sum of the result plus the 22-bit RAC. The first stage (1KIH modules) generates group borrows and enables, partial sums and carries, and bit enables stored by the partial sum holding register. The second stage (1KJH module) completes the computation

The second stage monitors the 18 -bit adder for end-around carry. If the sum or difference does not produce an end-around carry, a correction factor is added to the 22 -bit result Correction is required because the 18-bit adder can produce an end-around carry, which should be a plus 1 added to the least significant bit (bit 0 ) and not passed from bit 17 to the resulting adit adder. (A carry that passes from bit 17 to 18 with no correction cause factor by blocking the end-around carry from the 18 -bit adder and adding plus 1 to complemented bits 18 through 21 .

## INCREMENT FIELD LENGTH TEST

The 1 KGH and 4 KH 7 modules send I bits 0 through 17 to a field length tester in the CPU every clock period. When this increment operand is equal to or greater than FLC, CMC blocks the $C M$ write reference or returns a zero word for a $C M$ read reference. This occurs only when a 50 through 57 instruction causes a CM reference.

## DESTINATION REGISTER SELECTION

The CPU gates the increment operand to a destination register by sending a go increment signal to the 1KGH and 4 KH 7 modules. The lower 2 bits of the $f$ designator from the CIV register determine whether an ai ( 70 through 57 instructions), Bi (60 through 6 through 57 instructions with i not equal to zero, the CPU also gates the increment
operand plus RAC to the SAS in CMC by sending a go increment to storage -l signal to the 1 KJH module.

| Module | Location | Test Point | Description |
| :---: | :---: | :---: | :---: |
| 4KE7 | 6K09-12 | 53 | Bit $N$ enable |
|  |  | 52 | Bit $\mathrm{N}+1$ enable |
|  |  | 54 | $\mathrm{N}+2$ |
|  |  | 55 | $\mathrm{N}^{+3}$ |
|  |  | 64 | Bit N borrow |
|  |  | 62 | Bit $\mathrm{N}+1$ borrow |
|  |  | 65 41 |  |
|  |  | 44 | Group borrow |
|  |  | 76 | Select $\mathrm{Xj}_{j}$ |
|  |  | 75 | Select Aj |
|  |  | 71 | Select K |
|  |  | 72 | Select Bk |
|  |  | 61 51 | Select Select Bj |
|  |  | 66 | 25-nanosecond clock |
| 4KF7* | 6K13 | 36 | Bit 16 enable |
|  |  | 34 | Bit 17 enable |
|  |  | 32 | Bit 16 borrow |
|  |  | 26 | Group 4 enable |
|  |  | 21 | Group 4 borrow |
|  |  | 56 | f0 flag |
|  |  | 51 | fl flag |
|  |  | 63 | 25-nanosecond clock |
| 4KG7* | 6K14-15 | 63 | Transmit to X |
|  |  | 67 | Transmit to B |
|  |  | 65 | Transmit to A |
| 4KH7* | 6K16 |  | Transmit to X |
|  |  | 66 | Transmit to B |
|  |  | 64 | Transmit to A |
| 1KIH | 5K01-02 | 05 | Partial Carry Bit 0/11 |
|  |  | 02 | 1/12 |
|  |  | 13 | 2/13 |
|  |  | 55 | 10/21 |
|  |  | 03 | Partial Sum Bit 0/11 |
|  |  | 04 | 1/12 |
|  |  | 16 | 2/13 |
| 1K.J | 5K03 | 15 | Go Incr to Stor-1 |



INCREMENT UNIT

## LOGIC AND TIMING DIAGRAMS







Two 1 KIH modules and a 1 KJH module add the 22-bit RAC to the sum or difference of two 18-bit operands. The 1 KIH modules generate bit enables, group enables, group borrows, partial sums, and partial carries. These signals proceed to the lKJH module where the computation is completed.

The following example shows a pencil-and-paper addition for values which require correction.

$$
\left[\begin{array}{llllll}
1 & 2 & 3 & 4 & 5 & 6 \\
6 & 5 & 4 & 3 & 2 & 2 \\
-0 & 0 & 0 & 0 & 0 & 0
\end{array}\right]
$$

$$
\begin{aligned}
& \quad 1-1 \\
& \hline 00001
\end{aligned}
$$

$$
\begin{array}{llllllll}
0 & 3 & 0 & 0 & 0 & 0 & 0 & 0 \\
\hline 0 & 3 & 0 & 0 & 0 & 0 & 0 & 1
\end{array}
$$

18-bit sum with end around carry correction
22-bit result

The following table shows the conditions of the terms and output signals when the three operands are assigned values of the above example. Refer to the 1KJH module for the final result of this example.

| Input Signals | Groups | 5 |  | 4 |  |  |  | 3 |  |  |  |  | 2 |  |  |  | 1 |  |  |  | 0 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Bits | 21 | $20 \quad 19$ | 18 | 17 | 16 | 15 | $\begin{array}{lllll}14 & 13 & 12 & 11\end{array}$ |  |  |  |  | 10 | 9 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | $\overline{\mathrm{Aj}}=654321$ | $\begin{array}{lll} \text { F1 } & \text { F1 } & \text { F1 } \\ \text { F1 } & \text { F1 } & \text { F1 } \end{array}$ |  | $\begin{aligned} & \text { F1 } \\ & \text { F1 } \end{aligned}$ | 1 | 10 |  |  |  | 0 | 1 | 1 | 000 |  |  |  | 1 | 11 | 0 |  | 0 | 0 | 1 |  |
|  | $\overline{\mathrm{K}}=123455$ |  |  | 0 |  | 1 |  |  | 1 | 0 | 0 | 1 |  | 1 | 1 | 0 |  |  |  | 1 |  |  |  |
|  | $\overline{\mathrm{RAC}}=4717777$ | F0 10 |  |  | 0 | 1 | 1 |  |  | 1 |  | 11 |  | 11 |  |  | 1 | 1 | 1111 |  |  | 1 | 1 | 1 |  |
|  | PS10-PSO terms |  | 10 | 0 | 0 | 0 | 0 | 0 |  | 0 | 0 | 0 | 0 |  | 0 | 0 | 0 | 000 |  |  | 0 | 0 | 1 |  |
|  | PS3-PS0 terms |  | x x | x | x | x | x |  |  | 0 | 0 | 0 | x | X | x | x | x | x | x | x | 0 | 0 | 1 |  |
|  | PC10-PCO terms |  | 11 | 1 | 1 | 1 | 1 |  |  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |  |
|  | PC10, PC2-PCO terms |  | x x | x | x | x | x |  |  | 1 | 1 | 1 | 1 | X | x | x | x | x | X | x | x | 1 | 0 |  |
| Out put | E10-E4 terms |  | 01 | 1 | 1 | 1 | 1 |  |  | x | x | x | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | x | x | x |  |
|  | Group enables |  | Termn |  | 1 |  |  |  | x |  |  |  |  | 1 |  |  |  | 1 |  |  |  | x |  |  |
|  | Group borrows |  | 1 |  | 0 |  |  |  | x |  |  |  | 0 |  |  |  | 0 |  |  |  | x |  |  |  |

F=Forced

Partial sums PSO through PS10 are determined by the following truth table

| $\overline{\text { RAC }}$ | $\bar{j}$ | $\overline{\mathrm{~K}}$ | Condition |
| :---: | :---: | :---: | :---: |
| 1 | 1 | 1 | 1 |
| 1 | 1 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 1 |
| 0 | 0 | 1 | 1 |
| 0 | 0 | 0 | 0 |

Partial carries PCO through PC10 are determined by the following truth table.
$\overline{\mathrm{RAC}} \quad \bar{j} \quad \overline{\mathrm{~K}} \quad$ Condition

| 1 | 1 | 1 | 1 |
| :--- | :--- | :--- | :--- |
| 1 | 1 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 0 |
| 0 | 0 | 0 | 0 |



4KI7 MODULE

Two 4 KI 7 modules and a 3 KJ 7 module add RAC to the sum or difference of two 18 -bit operands. Each 4KI7 module performs a partial computation on nine bits of the three operands. The upper input pin pairs receive complemented data from the $A j, B j$, or $X j$ register. The middle input pin pairs receive complemented data from the CIW register $K$ designator or complementary (true or false) data from the Bk register. The lower input pin pairs receive data from the RAC register. The KI modules generate bit enables, bit borrows, group enables, and group borrows. These signals are sent to the 3 KJ 7 module where the computation is completed.

The following table shows the condition of the terms and output signals when the three operands are assigned arbitrary values. The output signals cause the 3KJ7 module to provide a result of 00236. (Note that Aj and K enter the 4 KI 7 modules in a complemented form.)

| Groups | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bits | $\begin{array}{lll}17 & 16 & 15\end{array}$ | $14 \quad 13 \quad 12$ | $\begin{array}{llll}11 & 10 & 9\end{array}$ | $8 \quad 76$ | $\begin{array}{llll}5 & 4 & 3\end{array}$ | 210 |
| $\overline{\mathrm{RAC}}=777600$ | $\begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | 110 | 000 | $0 \quad 00$ |
| $\overline{A_{j}}=777752$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{llll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 0 & 1\end{array}$ | 0 1 10 |
| $\overline{\mathrm{K}}=777765$ | $\begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{llll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 0\end{array}$ | $10_{1} 0$ |
| A8-A1 terms | $1 \begin{array}{lll}1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 0\end{array}$ | $\begin{array}{llll}0 & 1 & 1\end{array}$ | 11 X |
| B8-B1 terms | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{llll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{llll}1 & 1 & 1\end{array}$ | 100 | 0 O |
| C8-C3'terms | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{lll}1 & 1 & 1\end{array}$ | $1 \begin{array}{lll}1 & 1 & 1\end{array}$ | $\begin{array}{llll}1 & 1 & 1\end{array}$ | 0000 | $\mathrm{X} \times \mathrm{X}$ |
| Bit enables | $1 \begin{array}{lll}1 & 1 & \end{array}$ | $1 \begin{array}{lll}1 & 1 & 0\end{array}$ | $1 \begin{array}{lll}1 & 1\end{array}$ | $1 \begin{array}{llll}1 & 0 & 1\end{array}$ | $\begin{array}{llll}0 & 1 & 1\end{array}$ | $\begin{array}{lll}1 & 1 & 1\end{array}$ |
| Bit borrows | 1 X X | X X $\quad \mathrm{X}$ | $\mathrm{X} \quad 111$ | 1 XX | X X X | X 00 |
| Group enables | 0 | 0 | X | 0 | 0 | X |
| Group borrows | 1 | 1 | X | 1 | 0 | X |

The A1 through A8, B1 through B8, and C3 through C8 terms are the key to determining the enables and borrows for three given input operands. Bit $\mathrm{N}, \mathrm{N}+1, \mathrm{~N}+2$ enables, and A1 through A8 terms are determined by the following truth table.

| $\overline{\mathrm{RA}}$ | $\overline{\mathrm{j}}$ | $\overline{\mathrm{K}}$ | Condition |
| :---: | :---: | :---: | :---: |
| 1 | 1 | 1 | 1 |
| 1 | 1 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 1 |
| 0 | 0 | 1 | 1 |
| 0 | 0 | 0 | 0 |

Bit $\mathrm{N}, \mathrm{N}+1, \mathrm{~N}+8$ borrows, and B 1 through B 8 terms are determined by the following truth table.

| $\overline{R A}$ | $\overline{\mathrm{j}}$ | $\overline{\mathrm{K}}$ | Condition |
| :---: | :---: | :---: | :---: |
| 1 | 1 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 0 |
| 0 | 0 | 0 | 1 |

The C3 through C8 terms are the B2 through B7 terms left-shifted one bit position.


Two 1KIH modules and a 1 KJH module add the 22 -bit RAC to the sum or difference of two 18 -bit operands. The lKJH module completes the computation and sends the result to SAS when go increment to storage-1 is present. The result is based on input signals received from the 1KIH modules.
The following table shows the condition of various terms when the input signals are assigned an arbitrary value. (Note that the input signals are the same as the output signals shown
in the IKIH module table.)

(

## 3KJ7 MODULE

Two $4 \mathrm{KI7}$ modules and a $3 \mathrm{KJ7}$ module add RAC to the sum or difference of two 18 -bit operands. The' 3 KJ 7 module completes the computation and sends the result to the SAS when the gq increment to storage signal is present. This result is based on the condition of bits 0 through 15 enables; bits $0,1,8,9,10,17$ borrows; groups 1,2 , 4, 5 enables; and groups 1, 2, 4, 5 borrows. These input signals are generated by the 4KI7 modules. The following table shows the condition of various cage wire terms when the input signals are assigned an arbitrary value. (Note that the input signals are the same as the output signals shown on the 4 KI 7 module table.)

| Groups |  | 5 |  |  | 4 |  |  | 3 |  |  | 2 |  |  | 1 |  |  | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bits | 17 | 16 | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 |






# MANUAL TITLE: CDC CYBER 170 Models $175,740,750,760,865,875$ Functional Units Hardware Maintenance Manual PUBLICATION NO.: 60420300 REVISION: AA 

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