

# TECHNICAL MANUAL F口R 

## AM－500

HARD－DISK
CONTROLLER


Manufactured By
ALPHA MICROSYSTEMS
17881 SKY PARK NORTH
IRVINE, CALIFORNIA 92714

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SECTION 1
GENERAL DESCRIPTION

### 1.0 INTRODUCTION.

This Manual provides operating and maintenance instructions for the AM-500 Hard Disk Controller Circuit Board manufactured by Alpha Microsystems Inc., located in Irvine, California. Circuit board description, operating and usage instructions, programming, theory of operation, and maintenance instructions are included to provide the user with the information necessary to utilize this circuit board to its full capability.

### 1.1 CIRCUIT BOARD DESCRIPTION.

The AM-500 Hard Disk Controller Circuit Board provides data processing, control and interface capability between a standard S-100 bus system and one to four CDC 9427 H (Hawk) hard disk drives. A CPU module, a DMA module, a 16 x 8 First-In-First-Out (FIFO) memory and 1 K of Random Access Memory (RAM) provide the sophisticated data processing necessary for control of up to four ten-megabyte disk drives. A 2 K Read Only Memory (ROM) is contained on the board to provide a bootstrap load program and also to contain the microcode necessary for the CPU module operation.

A simplified block diagram of the circuit board is shown in Figure 1-1. For a complete detailed description of circuit board operation, see Section 4 of this manual. For programming requirements, see Section 3 of this manual.

### 1.2 APPLICATION.

This circuit board provides the data processing and interface capability necessary for operation of up to four CDC 9427H (Hawk) hard disk drives for a maximum of ten megabytes of memory capacity per drive. This circuit board is not recommended for any other disk drives at this time. See Section 2
of this manual for wiring instructions and system interface information. For complete information on the disk drive, see the Control Data Corporation 9427 H Hardware Maintenance Manual.


Figure 1-1. AM-500 Simplified Block Diagram

## SECTION 2 <br> OPERATING DATA

2.0 INTRODUCTION.

This Section contains information on the use of the AM-500 Hard Disk Controller circuit board. Capabilities, specifications, interface wiring, and user option descriptions are provided for the successful integration of the AM-500 into the user's system.

### 2.1 CAPABILITIES AND SPECIFICATIONS.

This circuit board operates from the standard S-100 bus structure to interface with the CDC 9427 H (Hawk) hard disk drive with 10 megabyte capacity (five fixed/five removable). A sing1e AM-500 board can support up to four drives for a total capacity of 40 megabytes of hard disk memory. Detailed specifications are listed in Table 2-1.

Table 2-1. AM-500 Specifications

| PARAMETER | SPECIFICATION |
| :--- | :--- |
| CPU Interface | Standard S-100 Bus - Interrupt Driven |
| Storage Capacity, <br> Each Drive | 10 Megabyte Total; 5 Megabytes on <br> fixed surface, 5 Megabytes on removable <br> cartridge. |
| Maximum Storage <br> capacity from each <br> AM-500 Controller <br> Board | 40 Megabytes with 4 Drives connected. |
| AM-500 to disk Bit <br> Transfer Rate. | 2.5 MHz. <br> Sector Size |
| 512 Bytes/Sector plus CRC Error Code <br> and Sentinel Check byte. |  |
| CPU Interface <br> Description | Multilevel interrupt driven; full <br> sector block transfers from or to <br> internal sector buffer. |

### 2.2 INTERFACE DESCRIPTION AND WIRING.

The AM-500 Hard Disk Controller Circuit Board provides interface capability between the Standard S-100 bus and CDC 9427 H (Hawk) hard disk drives.
2.2.1 S-100 BUS INTERFACE.

The AM-500 circuit board is fully $\mathrm{S}-100$ bus compatible. The board and its associated disk drives are addressed through the address lines and data is transferred through the standard data in and data out ports. The $S-100$ bus connections are made via the bottom edge connector and are described in Table 2-2.
2.2.2 DISK DRIVE INTERFACE.

The AM-500 circuit board accommodates a maximum of four hard disk drives. Unit select outputs from the circuit board select any one of the four drives individually and in arbitrary sequence. All disk drive interface signals are described in Table 2-3

Table 2-2. S-100 Bus Interface Signals List

| MNEMONIC | NAME | PIN | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| A0 | Address 0 | 79 | 16 bits of S-100 addressing. |
| A1 | Address 1 | 80 |  |
| A2 | Address 2 | 81 |  |
| A 3 | Address 3 | 31 |  |
| A4 | Address 4 | 30 |  |
| A 5 | Address 5 | 29 |  |
| A6 | Address 6 | 82 |  |
| A 7 | Address 7 | 83 |  |
| A 8 | Address 8 | 84 |  |
| A9 | Address 9 | 34 |  |
| A10 | Address 10 | 37 |  |
| A11 | Address 11 | 87 |  |
| A12 | Address 12 | 33 |  |
| A13 | Address 13 | 85 |  |
| A14 | Address 14 | 86 |  |
| A15 | Address 15 | 32 |  |
| DATAIN 0 | Input Data Bit 0 | 95 | Data input port. Bus |
| DATAIN 1 | Input Data Bit 1 | 94 | master input from slaves. |
| DATAIN 2 | Input Data Bit 2 | 41 |  |
| DATAIN 3 | Input Data Bit 3 | 42 |  |
| DATAIN 4 | Input Data Bit 4 | 91 |  |
| DATAIN 5 | Input Data Bit 5 | 92 |  |
| DATAIN 6 | Input Data Bit 6 | 93 |  |
| DATAIN 7 | Input Data Bit 7 | 43 |  |
| DATAOUT 0 | Output Data Bit 0 | 36 | Data output port. Bus |
| DATAOUT 1 | Output Data Bit 1 | 35 | master output to slaves. |
| DATAOUT 2 | Output Data Bit 2 | 88 |  |
| DATAOUT 3 | Output Data Bit 3 | 89 |  |
| DATAOUT 4 | Output Data Bit 4 | 38 |  |
| DATAOUT 5 | Output Data Bit 5 | 39 |  |
| DATAOUT 6 | Output Data Bit 6 | 40 |  |
| DATAOUT 7 | Output Data Bit 7 | 90 |  |

Table 2-2 (Cont.). S-100 Bus Interface Signals List

| MNEMONIC | NAME | PIN | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| PDBIN | Data Bus In | 78 | Read enable. Used by bus master to request addressed slave to place data on input port. |
| $\overline{\text { PHANTOM }}$ | Phantom | 67 | Disables phantom controlled memories when asserted. |
| $\overline{\text { PWR }}$ | Write Strobe | 77 | Write strobe. Generated by bus masters as write command to slaves. |
| $\overline{\text { RESET }}$ | Reset | 75 | AM-500 reset input. |
| SINP | I/O Input Cycle Status Line | 46 | Current bus cycle is a bus master input from an I/O address. |
| SMEMR | Memory Cycle <br> Input Status Line | 47 | Current bus cycle is a bus master input from a memory address. |
| SOUT | I/O Output Cycle Status Line | 45 | Current bus cycle is a bus master output to an I/O address. |
| $\begin{array}{ll} \hline \overline{\mathrm{VI}} & 0 \\ \hline \overline{\mathrm{VI}} & 1 \\ \hline \mathrm{VI} & 2 \\ \hline \mathrm{VI} & 3 \\ \hline \mathrm{VI} & 4 \\ \hline \mathrm{VI} & 5 \\ \hline \overline{\mathrm{VI}} & 6 \\ \hline \overline{\mathrm{VI}} & 7 \end{array}$ | Vectored Interrupt 0 <br> Vectored Interrupt 1 <br> Vectored Interrupt 2 <br> Vectored Interrupt 3 <br> Vectored Interrupt 4 <br> Vectored Interrupt 5 <br> Vectored Interrupt 6 <br> Vectored Interrupt 7 | 4 5 6 7 8 9 10 11 | ```Jumper selected interrupts. Standard interrupt leve1 = 4.``` |

Table 2-3. Disk Drive Interface Signal List

| MNEMONIC | NAME | PIN | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline \overline{\mathrm{CA}} 0 \\ & \hline \mathrm{CA} \\ & \hline \mathrm{CA} \\ & \hline \mathrm{CA} \\ & \hline \mathrm{CA} \\ & \hline \end{aligned}$ | Cylinder Address Bit 1 <br> Cylinder Address Bit 2 <br> Cy1inder Address Bit 4 <br> Cylinder Address Bit 8 <br> Cylinder Address Bit 16 <br> Cylinder Address Bit 32 <br> Cylinder Address Bit 64 <br> Cylinder Address Bit 128 <br> Cylinder Address Bit 256 | $\begin{aligned} & 39 \\ & 17 \\ & 40 \\ & 33 \\ & 28 \\ & 35 \\ & 31 \\ & 11 \\ & 24 \end{aligned}$ | The absolute address location of the cylinder to which the heads are to be positioned. |
| $\overline{\text { CYLSTR }}$ | Cylinder Address Strobe | 18 | Gates cylinder address information into the cylinder address register and initiates servo controlled head positioning. |
| DRVRDY | Drive Ready | 30 | Unit is ready to se $\kappa$, read, or write. |
| $\overline{\text { EGATE }}$ | Erase Gate | 37 | Enables erase current to flow through selected head. |
| FAULT | Fault | 22 | Attempted to write or erase when not on cylinder. <br> Attempted to read when write or erase is enabled. Current fault during read, write, or erase. |
| $\overline{\text { HDSEL }}$ | Head Select 1 | 26 | $1=$ Select upper head. <br> $0=$ Select lower head. |

Table 2-3 (Cont.). Disk Drive Interface Signal List

| MNEMONIC | NAME | PIN | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| INDEX | Index Mark | 27 | Index mark once per disk revolution (25 msec). |
| ONCYL | On Cylinder | 45 | Seek operation is complete and unit is ready to read or write. |
| $\overline{\text { PLSEL }}$ | Platter Select | 13 | 1=Select fixed disk; 0= Select removable cartridge. |
| READ CLOCK | Read Clock | 44 | Separated clocks concurrent with read data. |
| READ DATA | Read Data | 42 | Separated data input. |
| $\overline{\text { RESTOR }}$ | Return to Zero Seek | 14 | Positions the heads to cylinder zero (clears seek error). |
| $\overline{\text { RGATE }}$ | Read Gate | 41 | Enables separated data and clock to the interface lines. |
| SECTOR | Sector Mark | 29 | Sector mark pulse (2 msec). |
| SEKERR | Seek Error | 5 | Seek operation did not complete within 500 ms . |
| $\overline{S E L} 1$ <br> $\overline{S E L ~} 2$ <br> $\overline{S E L} 3$ <br> SEL 4 | Unit Select 1 <br> Unit Select 2 <br> Unit Select 3 <br> Unit Select 4 | $\begin{aligned} & 46 \\ & 47 \\ & 49 \\ & 48 \end{aligned}$ | Selects any one of the 4 disk units individually and in arbitrary sequence. |

Table 2-3 (Cont.). Disk Drive Interface Signal List

| MNEMONIC | NAME | PIN | DESCRIPTION |
| :--- | :--- | :---: | :--- |
| WGATE | Write Gate | 23 | Enables write current to <br> flow through selected head. |
| WPROT | Write Protected | 34 | Inhibits erase and write <br> currents. 1=Inhibit; <br> $0=$ Enable. |
| WRITE DATA | Write Data | 50 | Double frequency encoded <br> clock and data. |
| +5VDC <br> GND <br> GND <br> GND | G5VDC (not used) <br> GND <br> GND | 43 <br> 1 <br> 15 <br> 32 | Power and ground. |

### 2.2.3 WIRING AND CONNECTIONS.

When the AM-500 circuit board is received, it is ready for use. No adjustment or calibration is required for operation. The hardware requirements for use are described in this Section and the software requirements are described in Section 3.

First ensure that the proper power wiring is available and the correct voltages are connected to the various pins as shown on sheet 1 of the schematic. All power connections are made through the bottom edge connector.

All connections to the $\mathrm{S}-100$ bus are made through the bottom edge connector (Table 2-2) and connections to the disk drive are made through the top connector (Table 2-3). Ensure that these connections are correct before plugging the AM-500 circuit board into the system.
2.3 USER OPTIONS.

Several features of the AM- 500 circuit board can be jumper selected at the user's option. These jumpers must be installed before the AM- 500 circuit board is installed in the system. The standard jumper connections are shown in Figure 2-1.


Figure 2-1. Standard Jumper Connections

### 2.3.1 ADDRESS CODE.

Circuit board addressing can be selected at the user's option for any address block (in increments of four) on the address lines A0-A7. Connect jumper wires to +5 V or ground for the desired address. Jumper pads are located on the circuit board between U36 and U37 (schematic sheet 2). The standard I/O address is C0-C3.

### 2.3.2 INTERRUPT LINES.

Interrupt compatibility for any $\mathrm{S}-100$ bus system is provided with jumpers to any of the vectored interrupt lines VIO-VI7. Attach the jumper wire from the pad connected to pin 3 of U35 to the desired interrupt line (schematic sheet 11). Jumper pads are located on the circuit board directly below U35. The standard interrupt level is VI4.

### 2.3.3 PHANTOM FEATURE.

A phantom feature can be used with the AM-500 Hard Disk Controller. This feature is supported in operating system software release 3.0 and later and will not operate if a video display module (VDM) is being used for software program DYSTAT because the system memory overlaps the VDM memory (see Alpha Micro Operating System AMOS manual). A1'so, the memory board must support the phantom feature.

The AM-500 circuit board comes from the factory configured to enable the phantom feature. If it is desired to disable the phantom feature in order to use the disk drive as a peripheral device, it will be necessary to cut the etch that runs between J1-67 and U35-5. Cut the etch between the two feed thru holes on the solder side of the board. The S-100 pins on the back of the card start at 51 on the right and count up to 100 on the left.

The AM-500 already has the phantom signal connected to pin 67 of the S-100 bus. During bootstrap, the line goes low disabling the upper memory board or any board that monitors the phantom line. The CPU starts executing the bootstrap PROM at location F 400 and moves it into location 76000 octal or 7C00 hex. The bootstrap then starts executing from memory. Any memory board that overlaps the bootstrap PROM must have phantom enabled and all the others must have it disabled.

### 2.3.4 BOOT OPTIONS.

Since the AM-200, AM-210 and AM-400 also contain internal bootstrap loader PROMS, the user may desire to disable the bootstrap from the AM-500 and boot from one of these other circuit boards. The following paragraphs describe the requirements for these options.
2.3.4.1 THE AM-500 AS A PERIPHERAL TO THE AM-200 OR AM-210. Remove the 74 LS 30 IC (U40) from the AM-500 (bottom row, 6th chip from the right). If the AM- 200 or AM- 210 are using phantom (phantom must be used if you have more than 56 K of memory), the phantom etch must be cut on the AM-500. Turn the AM-500 board over (component side down) and cut the etch going to pin 67. Cut the etch between the two eyelets so a wire may be inserted between the two eyelets at a later date to reenable phantom. The $S-100$ pins on the back side start at 51 on the right and go to 100 on the left.

Boot the system with both the AM-500 and AM-200 or AM-210 in the system.

Log into area 1,4 on the floppy disk and modify your SYSTEM.INI to add the following:

> BITMAP HWK,606,0,1

Be sure that the DEVTBL command in the SYSTEM.INI contains HWK0 and HWK1.

Log into area 1,6 and do the following:

COPY HWK.DVR=HWK500.DVR

Now reboot the system again and by mounting HWKO and HWK1, you may access them as peripherals to the AM- 200 or AM- 210. Remember when in this mode the fixed platter is always HWKO.
2.3.4.2 OPERATING THE AM-500 WITH THE AM-400/TRIDENT. When running these two in combination, phantom must be disabled on the controller that is not the system device. If it is desired to remove phantom on the AM-400, remove $Z 32$
pin 5 from the socket so it sticks out. If you are booting from the AM-500, disable the PROM on the AM-400 by removing Z29, which is the IC to the left of the header in the bottom row.

The proper bitmap sizes and device names must be in the SYSTEM.INI file to accommodate the proper configuration. Be sure the proper driver in area 1,6 is renamed to the same name as the device you defined in the DEVTBL. The bitmap sizes are 562 for the $T-25,1022$ for the $T-50,1534$ for the T-80, and 1630 for the $\mathrm{T}-300$.

If booting is off the AM-500, initialization must be from the AM-400. This is done by running a program. For the $T-25$, T-50, and T-200 use TRIINI.PRG, and for the T-80 and T-300 use T80INI.PRG. To utilize it, just type TxxINI followed by a carriage return. The program should return a status less than 3 octal. If you do not run this initialization program, the drives will show as NOT READY on a SYSTAT.

## SECTION 3

PROGRAMMING

### 3.0 INTRODUCTION.

This section describes the programming requirements for the AM-500 circuit board. Circuit board addressing, bootstrap loader, and AM-500 internal programming are described for complete system compatibility.

When the AM-500 circuit board is received, it is ready for use with the CDC 9427 H (Hawk) disk drive as described in Section 2 of this manual. The bootstrap loader and the microcode contained in internal firmware is designed only for this disk drive.

### 3.1 ADDRESSING.

The AM-500 and associated disk drives are addressed through the $\mathrm{S}-100$ bus address lines. The circuit board address is jumper selectable by the jumpers connected to U37. Jumpers to either +5 V or ground select the board address on $\mathrm{S}-100$ bus address lines AD2-AD7. The board always occupies four I/O ports even though only two are actually used.

### 3.2 INITIALIZATION.

On initial power-up or reset, the AM-500 goes into a dormant state waiting for initialization. With the phantom option enabled, the phantom signal is activated, disabling any memory. The $S-100$ bus is enabled to access the bootstrap loader contained in the AM-500 PROM.

The 16 K PROM (U21) contains both the bootstrap routine and the controller microcode. The microcode resides in the first 1 K bytes, and the bootstrap routine resides in the second 1 K bytes of this PROM. The current bootstrap routine is written in

AM-100 code, and other user codes may be used if the microcode in the first 1 K bytes is dup1icated exactly.

NOTE
If the user code contained in the AM-500 PROM is changed from the current boot routine, the microcode in the first 1 K of memory must be duplicated exactly.

Initialization of the AM-500 is accomplished by writing a zero byte into the $\mathrm{AM}-500$ command register during system initialization. The CPU finishes initialization of the disk controller and releases the Phantom signal. See paragraph 2.3.4 for boot options.

### 3.3 COMMAND REGISTER.

The command register receives AM-500 commands from the $\mathrm{S}-100$ bus. It is accessed by writing to the selected address (see paragraph 2.3.1). The legal base addresses are in multiples of four from the first address 177400 (octal), 177404 (oct 1), 177408 (octal), through 177774 (octal). A software hands nake sequence must be used for the command transfer (see paragraph 3.6).

There are two basic types of commands used for operation of the AM-500 system; non-disk commands and disk commands. Both types are described in Figure 3-1.

### 3.4 STATUS REGISTER.

The status register is used to determine the current status of the AM-500 controller. It is accessed from the $S=100$ bus by reading from the user selected base address of the circuit board. The format of the status register is described in Table 3-1.

NON-DISK COMMANDS:

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IE | CLR | M | M | F | $\mathrm{R} / \mathrm{W}$ | 1 | 0 |

DISK COMMANDS:


Figure 3-1. AM-500 Commands

Table 3-1. Status Word Formats

| BIT |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $7 *$ | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
| CR | 0 | 0 | 0 | 0 | 1 | 0 | 1 | Disk Not Ready |
| CR | 0 | 0 | 0 | 1 | 0 | 1 | 1 | Write Protected |
| CR | 1 | 0 | 0 | 0 | 0 | 0 | 1 | Seek Error |
| CR | 1 | 0 | 0 | 0 | 0 | 1 | 1 | Sector Not Found |
| CR | 1 | 0 | 0 | 0 | 1 | 0 | 0 | CRC Error |
| CR | 1 | 0 | 0 | 0 | 1 | 0 | 1 | Sentine1 Field Error |
| $* \mathrm{CR}=$ COMMAND RECEIVED, $0=$ CLEARED, $1=$ RECEIVED. |  |  |  |  |  |  |  |  |

The sentinel field error indicates that data was not correctly transferred within the AM-500. The sentinel feature checks byte 513 for the correct code 5 A (hex) after a 512 byte transfer.

### 3.5 INTERRUPT MODE.

The interrupt mode in the AM-500 can be selected by both disk and non-disk commands. If the interrupt mode is selected, the interrupt occurs at the completion of the current AM-500 command. The interrupt must be cleared by the user by setting bit $6=1$ of any non-disk command.
3.6 HANDSHAKE SEQUENCE.

A handshaking type interaction is required between the AM-500 circuit board and the CPU any time a command is sent to the AM-500. This handshaking is a software operation that uses bit 7 of the Status Register. Handshake operations proceed in the following sequence:

1. A command is sent from the $\mathrm{S}-100$ bus to the Command Register.
2. The system software waits for bit 7 of the Status Register to become a 1 to guarantee that the

AM-500 has received the command.
3. The system software writes a zero into the Command Register to clear the previous command.
4. The system software waits until bit 7 of the Status Register becomes a 0 indicating that the Command Register has been cleared.

This completes the handshake operation and the program continues.

### 3.7 RELKEY DESCRIPTION.

The term RELKEY refers to a double byte number which points to one of 9696 possible sectors on the disk. The user loads a number from 0 to 9695 into the Relkey bytes of the DMA buffer to indicate which sector is to be operated on. If bit 5 of the disk command is a 1 , the microcode converts this number along with the drive select byte into an absolute disk address and stores it back into the DMA buffer. The microcode replaces the drive number and Relkey bytes of the DMA buffer with the converted absolute disk address. Since all DMA buffer locations are accessable, the absolute disk address can be retrieved if necessary.

If the user has the desired absolute disk address, it can be placed in the DMA buffer, and the Relkey conversion can be skipped. This is accomplished as follows:

1. Place the actual drive number (1-3) into the second byte of the buffer.
2. Place the low byte of the track address in the third byte of the buffer.
3. P1ace the high byte of the track address (which specifies the surface, the platter select, and the sector address) into the fourth byte of the buffer as shown below.

3.8 DMA BUFFER.

The DMA Buffer is contained in the first 1 K RAM in the AM-500. A11 data transfers to or from the disk memory must pass through this buffer.

The DMA buffer is accessed by addressing the AM-500 base address plus one and all data between the $\mathrm{AM}-500$ and the $\mathrm{S}-100$ bus must pass through this single port. No DMA type operations occur between the AM-500 and system memory because they are handled on a programmed basis only. The microcode initializes the buffer and sets the DMA channels for proper operation as shown in Figure 3-2. The Handshake sequence is shown in Figure 3-3, and the data transfer sequence is shown in Figure 3-4.
3.8.1 DMA BUFFER OPERATION FOR DISK WRITE.

Transferring data to the DMA buffer and then performing ink write operations takes place as follows:

1. Send DMA write command X06 (octa1) to AM-500 command register and perform a handshake. ( $X$ depends on selection of interrupt mode.)
2. Send retry count to AM- 500 DMA buffer.
3. Send disk number ( $0-7$ ) to DMA buffer.
4. Send low byte of Relkey to DMA buffer.
5. Send high byte of Relkey to DMA buffer.
6. Send data (0-512 bytes) to DMA buffer.
7. Send disk write command X 71 (octa1) to $\mathrm{AM}-500$ command register and perform a handshake.
8. Read the $\mathrm{AM}-500$ status register and check for zero


Figure 3-2. DMA Buffer Operation


Figure 3-3. Handshake Sequence


Figure 3-4. DMA Transfer Sequence
condition indicating no error.
9. Continue with next command if no errors were detected.
3.8.2 DMA BUFFER OPERATION FOR DISK READ.

Transferring data from the disk into the DMA buffer and then reading the DMA buffer is performed as follows:

1. Send DMA write command X06 (octa1) to AM-500 command register and perform a handshake.
2. Write retry count into DMA buffer.
3. Write a disk number (0-7) into DMA buffer.
4. Write low byte of Relkey into DMA buffer.
5. Write high byte of Re1key into DMA buffer.
6. Send disk READ command X35 (octal) to AM-500 command register and perform handshake.
7. Read AM-500 status register and check for zero condition indicating no errors occurred.
8. Assuming no errors occurred during the disk read, send the DMA read command X12 (octal) to the AM-500 command register and perform the normal handshake.
9. Read one byte of data from the DMA buffer and discard it (the first byte is always invalid data).
10. Read valid data, 1 to 512 bytes, from the DMA buffer.

### 3.9 RESTORE COMMAND.

The restore command is used to reset the disk logic and force the disk head to return to track zero. The user may force a restore command with a disk command 3 (octal), or the microcode will exercise the restore command on its own following certain errors.

If a disk fault occurs, the microcode attempts one restore and then checks the fault status again. If the fault still exists, the microcode aborts the command operation and reports an error to the CPU.
3.10 DISK FORMATTING COMMANDS.

There are two disk formatting commands, Format Record and Format Track, that are identical in function. Both commands cause the currently addressed track to be formatted.

## CAUTION

The Format Record and Format Track commands both destroy previously stored data on the addressed track by writing all new records with data fields of zeros.

### 3.11 STATUS CHECK.

The Read and Write Status check commands are used to access specific status information from the disk drive.

The Write Status check command 001 (octal) is used to determine if the currently addressed disk is write protected. It does not attempt to actually write any data. If the addressed disk is write protected, error code 13 (octal) is placed in the status register.

The Read Status check command (if SEEK bit 4 is on) seeks for the currently addressed track and then returns. Data is not read from the disk. If a seek error occurred, an error code 14 (octal) is placed in the status register.

### 3.12 ERROR CODE REPORTING.

The operating system does not normally report soft disk errors. A soft disk error is defined as a non-recoverable error after a set number of retries, usually 4 on the AM-500. If after 4 retries the error is non-recoverable, the system aborts the command and reports an error.

To enable reporting of the error type the operator should type:

SET DSKERR

The AM-500 error codes are as follows:

## 1 - SEEK ERROR

2 - NOT USED
3 - SECTOR NOT FOUND
4 - CRC CHECKSUM
5 - SOFTWARE CHECKSUM

## SECTION 4 <br> FUNCTIONAL THEORY OF OPERATION

### 4.0 INTRODUCTION.

The AM-500 Hard Disk Controller circuit board contains integrated circuit elements for the necessary data processing for the performance of the functions as described in Sections 1, 2 and 3 of this manual. This Section describes the functional theory of operation of the circuit board and also provides information for each of the integrated circuit elements.
4.1 CIRCUIT BOARD OPERATION.

This circuit board provides control and interface capability between the S-100 bus and the CDC 9427 H (Hawk) hard disk drive. The functional block diagram of the circuit board is shown in Figure 4-1, and the circuit board schematic is contained in Section 6 of this manual. Table 4-1 contains a list of the signals used in this circuit board with definitions of their functions. For S-100 bus interface signals, see Table 2-2. For Hard Disk drive signals, see Table 2-3.
4.1.1 POWER-UP RESET.

On initial power-up, the circuit board is in a dormant state waiting for initialization. System clocks are generated from the 10 MHz oscillator on U 1 and divided down by $\mathrm{U} 2, \mathrm{U} 22$ and U 43 , to generate CLOCK, $\overline{0 N E}$ and $\overline{\text { CELL }}$ clock signals. The power-on reset circuit activates $\overline{R E S E T}$ to reset the INIT and RUN flipflops. The RUN signal activates the $\overline{\text { PHANTOM }}$ signal to enable any memory under phantom control, and the $\overline{\text { INIT signal enables }}$ the boot logic so the $S-100$ bus can access the bootstrap program contained in the PROM U21.



Table 4-1. AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF <br> SOURCE | FUNCTION |
| :---: | :---: | :---: | :---: |
| ADSTB | ADDRESS STROBE | 7 | Strobes upper address byte into external latch. |
| AEN | ADDRESS ENABLE | 7 | Active high that enables external latch which holds upper byte of address, and disables system during DMA cyc1e. |
| A0-A15 | ADDRESS LINES | 2, 6, 12 | S-100 Bus Address Lines. |
| B00T | BOOT LOAD | 5 | Enables a boot load from the PROM. |
| BUSAK | BUS ACKNOWLEDGE | 6 | Active low CPU output to indicate to the requesting device that the CPU address bus, data bus and tri-state control bus signals have been set to their high impedance state. |
| BUSRQ | BUS REQUEST | 7 | Request to the CPU for control of the system bus. |
| CELL | CELL CLOCK | 3 | Clock generated for clock and data output to disk drive. |
| CLOCK | 2.5 MHz CLOCK | 3 | Clock generator output. |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF SOURCE | FUNCTION |
| :---: | :---: | :---: | :---: |
| CMDWRT | COMMAND WRITE | 2 | Sets initialize flip-f1op. |
| CNTRL | CONTROL | 9 | Clocks data from Data Bus (D0-07) into control register. |
| COMAND | COMMAND | 9 | Enables the output of the command register to place its contents on the Data Bus (D0-07). |
| CPS 1 | SERIAL INPUT CLOCK | 10 | Clock input for serial data input to FIFO memory stack. |
| CPSO | SERIAL OUTPUT CLOCK | 9 | (Input to FIFO.) |
| CRC | CYCLIC REDUNDANCY CHECK | 9 | Data output of CRC checker. |
| CRC CLK | CRC CLOCK | 9 | Clock input to CRC flip-flop generated from either $\overline{\text { CELL }}$ or $\overline{\text { RCLOCK. WRITE }=1 ~ s e l e c t s ~ \overline{C E L L}}$ WRITE=0 selects $\overline{\text { RCLOCK. }}$ |
| CRC DAT | CRC DATA | 9 | Data input to CRC checker. |
| CRC ERR | CRC ERROR | 9 | Error output of CRC checker. |
| CYL ADR | CYLINDER ADDRESS | 9 | Clocks Data from Data Bus into Drive Register 1. |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF SOURCE | FUNCTION |
| :---: | :---: | :---: | :---: |
| DACK 0 | DMA ACKNOWLEDGE 1 | 7 | DMA controller output. |
| DACK 3 | DMA ACKNOWLEDGE 3 | 7 | DMA controller output. |
| DATA | DATA RESET | 11 | Master reset input to FIFO Memory and CRC checker. |
| DATRD | DATA READ | 2 | Detects Read Cycle. |
| DATWRT | DATA WRITE | 2 | Detects Write Cycle. |
| DIOR | DATA I/O READ | 5 | Enables data on S-100 bus D00-D07 to be placed on internal Data Bus and resets DREQ 0 to DMA controller. |
| DIOW | DATA I/O WRITE | 5 | Resets DREQ1 to DMA controller. |
| DMASEL | DMA SELECT | 9 | Chip Select input to enable DMA Controller. |
| D0-D7 | DATA LINES | 5 | Tri-State bus internal to AM-500. |
| DREQ 0 | DMA REQUEST 0 | 7 | DMA Request input. |
| DREQ 1 | DMA REQUEST 1 | 7 | DMA Request input 1. |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF <br> SOURCE | FUNCTION |
| :---: | :---: | :---: | :---: |
| DREQ 3 | DMA REQUEST 3 | 9 | DMA Request input 3. |
| DRLCLR | DMA REQUEST <br> LATCH CLEAR | 9 | Clears DMA request latches and presets the RUN flip-f1op. |
| DRVSEL | DRIVE SELECT | 9 | Clock for Drive Register 1. |
| DRVSTS | DRIVE STATUS | 9 | Gates status data from the disk drive onto the internal data bus. |
| FEMPTY | FIFO MEMORY EMPTY | 9 | Enables the CRC checker and write logic when the FIFO memory is not empty. |
| FSYNC | FIFO MEMORY SYNC | 9 | Enables clock input to FIFO memory. |
| INIT | INITIALIZE | 3 | Reset signal to CPU. |
| IOR | I/O READ | 6 | Bidirectional, active low, three state line. |
| IOSEL | I/O SELECT | 2 | Enables the circuit board when address lines AD2-AD7 match the address code on the address jumpers. |
| IOW | I/O WRITE | 6 | Bidirectional, active low, three state line. |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF SOURCE | FUNCTION |
| :---: | :---: | :---: | :---: |
| IRF | INPUT REGISTER FULL | 10 | Active low output of the upper FIFO memory when its input. register is full. |
| LIRF | - LOWER INPUT REGISTER FULL | 10 | Active low output of the lower FIFO memory when its input register is full. |
| LORE | LOWER OUTPUT REGISTER EMPTY | 10 | Active low output of the lower FIFO memory when its output register is empty (FIFO ORE). |
| MSL | MEMORY SELECT | 9 | Enables the memory select decoder. |
| ONE | ONE CLOCK | 3 | Clock generated for clock and data output to disk drive. |
| ORE | OUTPUT REGISTER EMPTY | 10 | Active low output of the upper FIFO memory when its output register is empty. |
| PL | PARALLEL LOAD | 10 | Paralle1 load signal to the FIFO memory. |
| RAMSEL | RAM MEMORY SELECT | 9 | Chip Select signal to enable the two Random Access Memory modules. |
| RCLOCK | READ CLOCK | 3 | Clock input from disk. |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF <br> SOURCE | FUNCTION |
| :--- | :--- | :---: | :--- |
| RDATA | READ DATA | 3 | Latched data input from disk. |$|$| READ |
| :--- |
| READ |
| REGSEL |
| REGISTER |
| SELECT |

Table 4-1 (Cont.). AM-500 Signal List

| SIGNAL | NAME | SCHEM <br> PAGE OF <br> SOURCE | FUNCTION |
| :---: | :--- | :--- | :--- |
| TOP | TRANSFER OUT | 10 | FIFO memory input to transfer <br> parallel data into its output <br> register. |
| WDATA | WRITE DATA | 10 | Serial output of FIFO memory <br> to be written onto disk. |
| WGATE | WRITE GATE | 11 | Write enable output to disk. |

### 4.1.2 ADDRESSING.

Address data is received from the $\mathrm{S}-100$ bus on address lines A0-A15. This address data is transferred to a tri-state address bus for use by the controllers and memories within the circuit board and is also used for direct addressing of the circuit board.

Tri-State Buffers on U38 and U39 gate the address data to the internal bus. Initialize signal INIT controls these gates when set by the Command Write signal $\overline{\text { CMDWRT. }}$

Address lines A2-A7 are wired directly to comparator U37 for circuit board addressing. The other inputs come from jumper wires to either +5 V or ground to produce the desired circuit board address. Signal $\overline{\text { IOSEL }}$ is asserted when the data from the adedress lines compare with the address of the card. This generates $I / O$ read and write commands when combined with the S-100 bus control signals in decoder U17. The board address is programmable from 00 to $F C$ (hex) in increments of 4.
4.1.3 CPU OUTPUT.

CPU Output data is transmitted to the AM-500 by the $\mathrm{S}-100$ data and control lines. It is received in the AM-500 by the Command Register and Data Out Register.

### 4.1.3.1 COMMAND REGISTER.

The command register is used by the AM-500 to receive commands from the $\mathrm{S}-100$ bus and consists of the register on U30. The input to the command register comes from the $\mathrm{S}-100$ bus data lines D00-D07; and the output, when enabled, goes to the internal AM-500 data bus D0-D7.

Data from the $\mathrm{S}-100$ bus data lines is clocked into the command register by the Command Write signal $\overline{\text { CMDWRT. This comes from }}$ the I/O command register when the AM-500 is addressed and A0 and A1 are zero. When I/O Output signal (SOUT) from the $\mathrm{S}-100$ bus is asserted, $\overline{\text { CMDWRT }}$ clocks the command word from the $\mathrm{S}-100$ bus data lines into the command register and also sets the initialize flip-flop U36 to begin an AM-500 cycle. Flip-flop U36 remains set until the RESET pushbutton is pressed or power turned off. When the CPU issues a Memory Read Command (RD low and WR high) and address lines $A 0$ and $A 1$ are zero, the Register/Memory Select decoders issue the COMAND signal to transfer the command register contents to the AM-500 data bus.

### 4.1.3.2 COMMAND TYPES.

The non-disk commands issued to the AM-500 are used for control and operation of the logic internal to the circuit card. The bit functions are listed in Table 4-2 and the commands are listed in Table 4-3. Disk Commands are used to control the disk drive and access status information from the drive. Disk command bit functions are listed in Table 4-4, and disk command codes are listed in Table 4-5.

Table 4-2. Non-Disk Command Bits


Table 4-3. Non-Disk Command Formats

|  | FUNCTION |
| :---: | :---: |
| $\begin{array}{llllllll} 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ \mathrm{X} & \mathrm{X} & 0 & 0 & 0 & 1 & 1 & 0 \\ \mathrm{X} & \mathrm{X} & 0 & 1 & 0 & 1 & 1 & 0 \end{array}$ | ```Initialization/Boot Load Write data to DMA buffer (RAM Base) Write data to DMA buffer (Data Base)``` |
| $\begin{array}{llllllll} \mathrm{X} & \mathrm{X} & 0 & 0 & 0 & 0 & 1 & 0 \\ \mathrm{X} & \mathrm{X} & 0 & 1 & 0 & 0 & 1 & 0 \end{array}$ | Read data from DMA buffer (RAM Base) <br> Read data from DMA buffer (Data Base) |
| $\begin{array}{llllllll} 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ \mathrm{X} & 1 & \mathrm{X} & \mathrm{X} & \mathrm{X} & \mathrm{X} & 1 & 0 \end{array}$ | ```Clear Interrupt (Clear Only) Clear Interrupt (P1us some other non- disk command)``` |
| $\begin{array}{lllllllll}\mathrm{X} & \mathrm{X} & 1 & 0 & 0 & 1 & 1 & 0\end{array}$ | Lead user microcode pointer |
| $\begin{array}{llllllll}\mathrm{X} & \mathrm{X} & 1 & 1 & 0 & 1 & 1 & 0\end{array}$ | Indirect load via user pointer |
| $\begin{array}{llllllll}\mathrm{X} & \mathrm{X} & 1 & 1 & 1 & 0 & 1 & 0\end{array}$ | Execute from users microcode pointer |
| $\begin{array}{llllllllll}X & X & 1 & 1 & 0 & 0 & 1 & 0\end{array}$ | Indirect read via user pointer |
| $\begin{array}{llllllll}\mathrm{X} & \mathrm{X} & 1 & 0 & 0 & 0 & 1 & 0\end{array}$ | Read user microcode pointer |

Table 4-4. Disk Command Bits

| BIT | NAME | FUNCTION |
| :---: | :--- | :---: |
| 0 | DSK | Always 1 for disk type commands |
| 1 | TRK* | Part of command function |
| 2 | FN* | Command function comprised of bits 1, <br> 2 and 3 taken as a whole |
| 3 | R/W* | $0=$ Read $1=$ Write |
| 4 | SK | Seek |
| 5 | CVT | Convert Relkey $1=$ Convert 0=No Conversion |
| 6 | - | Not used |
| 7 | IE | Interrupt enable 0=Disable $1=$ Enable |

Table 4-5. Disk Command Codes

| BITS |  |  | FUNCTION |
| :--- | :--- | :--- | :--- |
| 3 <br> R/W | 2 <br> FN | 1 <br> TRK |  |
| 0 | 0 | 0 | Read Status Check |
| 0 | 0 | 0 | Restore to Cylinder Zero |
| 0 | 1 | 0 | Read Record |
| 0 | 1 | 1 | Read Track |
| 1 | 0 | 0 | Write Record |
| 1 | 0 | 1 | Write Status Check |
| 1 | 1 | 0 | Format Record (same result as 010) |
| 1 | 1 | 1 | Format Track (Same result as 011) |

4.1.3.3 DATA OUT REGISTER.

The Data Out register controls the transfer of data from the S-100 bus data lines to the data bus internal to the AM- 500 circuit board and consists of the register on U41. The input to the Data Out register comes from the $\mathrm{S}-100$ bus data lines DOO-DO7; and the output, when enabled, goes to the internal AM-500 data bus D0-D.7.

Data from the $\mathrm{S}-100$ bus data lines is clocked into the Data Out register by the Data Write signal DATWRT. This comes from the $I / O$ command register when the $A M-500$ is addressed and $A 0=1$ and $A 1=0$. When $I / O$ Output signal (SOUT) from the $S-100$
bus is asserted, $\overline{\text { DATWRT }}$ clocks the data word from the $\mathrm{S}-100$ bus into the Data Out Register and also sets the DMA request latch for a DMA service request. The DMA controller issues an acknowledge ( $\overline{\text { DACKO }}$ ) that is combined with address bit $\mathrm{A} 15=1$ and a Read Output from the CPU to generate Data I/O Read signal DIOR. Signal $\overline{\mathrm{DIOR}}$ transfers the contents of the Data Out Register onto the internal data bus and also resets the DMA request latch.
4.1.4 CPU INPUT.

CPU input data is transmitted from the AM-500 by the $\mathrm{S}-100$ data and control lines. It is transmitted from the AM-500 by the Status Register and Data In Register.
4.1.4.1. STATUS REGISTER.

The Status Register is used to determine the current status of AM-500 Controller operation and consists of the register on U31. The input to the Status Register comes from the AM-500 internal data bus D0-D7; and the output, when enab1ed, goes to the S-100 bus data lines DIO-DI7.

Data from the internal data bus is clocked into the Status Register by the $\overline{\text { STATUS }}$ signal from the Register/Memory Select logic. This occurs when the board is addressed and lines A0, A1 and $\overline{W R}$ are all low. The Register/Memory Select decoders issue the STATUS signal to clock the data into the Status Register from the internal data bus. The contents of the Status Register are applied to the S-100 bus data lines by the Status Read signal ( $\overline{\text { STSRD }}$ ) from the $I / O$ Command Register. This signal is issued when address line $\mathrm{A} 0=0$, Write Strobe $\overline{P W R}$ is false (high), Read enable is True (high), and Activate Input (SINP) is high.
4.1.4.2 STATUS WORD FORMATS.

Status words contain information regarding the operation of
the AM-500 controller. Information and formats of the status words are contained in Table 4-6. Bit 7 is the Command Received (CR) bit and $1=$ Received, $0=$ Cleared.

Table 4-6. Status Word Formats

| BIT |  |  |  |  |  | FUNCTION |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $7 *$ | 6 | 5 | 4 | 3 | 2 |  | 0 |  |
| CR | 0 | 0 | 0 | 0 | 1 | 0 | 1 | Disk Not Ready |
| CR | 0 | 0 | 0 | 1 | 0 | 1 | 1 | Write Protected |
| CR | 1 | 0 | 0 | 0 | 0 | 0 | 1 | Seek Error |
| CR | 1 | 0 | 0 | 0 | 0 | 1 | 1 | Sector Not Found |
| CR | 1 | 0 | 0 | 0 | 1 | 0 | 0 | CRC Error |
| CR | 1 | 0 | 0 | 0 | 1 | 0 | 1 | Sentine1 Field Error |

* CR = Command Received $0=$ Cleared, $1=$ Received.
4.1.4.3 DATA IN REGISTER.

The Data In Register controls the transfer of data from the data bus internal to the $\mathrm{AM}-500$ to the $\mathrm{S}-100$ bus data lines and consists of the register on U42. The input to this lagister comes from the internal data lines $\mathrm{D} 0-\mathrm{D} 7$ and the output, when enabled, goes to the $\mathrm{S}-100$ bus data lines DIO-DI7.

Data from the internal bus is clocked into the register when either Data I/O Write ( $\overline{\mathrm{DIOW}})$ is true or RUN is false. Signal $\overline{\text { DIOW }}$ comes from Bus Acknowledge ( $\overline{\text { BUSAK }}$ ) and Write ( $\overline{\mathrm{WR})}$ from the CPU when DMA is selected by the Memory/Register Select logic. Data is placed on the $\mathrm{S}-100$ bus data lines when in the Bootstrap Mode or when the I/O Command Register issues a Data Read $\overline{D A T R D}$ command. The $\overline{D A T R D}$ is issued when the board is addressed and $\mathrm{A} 0=1, \mathrm{~A} 1=0$, Read Enable PDBIN is true (high) and Write Strobe ( $\overline{\mathrm{PWR})}$ is false (high).
4.1.5 BOOT LOAD PROCEDURE.

The AM-500 circuit board contains a bootstrap load routine stored in the ROM on U21. The microcode for internal AM-500 operation is also contained in the first 1 K of ROM memory. The bootstrap program is contained in the second 1 K of memory.

NOTE
If the user code is changed from the current boot routine, the microcode in the first 1 K of memory must be duplicated exactly.

The CPU reads the bootstrap load program by addressing the circuit card and reading from 172000 (octal) up. This address, combined with SMEMR from the $\mathrm{S}-100$ bus and a high $\overline{\mathrm{INIT}}$ signal which takes place on initial start-up, generates the $\overline{\text { BOOT }}$ signal. The $\overline{\text { BOOT }}$ signal enables the PROM (U21) and enables the output of the Data In Register. The clock input of the Data In Register is supplied by the RUN signal through gate U32. The RUN signal was set low by initial start-up and remains so until DRLCLR presets the RUN flip-flop. The Data In Register consists of transparent latches so that under these conditions, the boot data from the PROM enters the internal data bus and transfers directly to the $\mathrm{S}-100$ bus data lines.

When the block data transfer of the boot loader is completed, the phantom is disabled and the CPU and DMA modules initialize the circuit board so that it can accept commands.

### 4.1.6 INTERRUPTS.

The interrupt mode can be selected for AM-500 operation when the software program sets bit 7 in either disk or non-disk commands. When the interrupt mode is selected, the interrupt occurs at the completion of the current AM-500 command. This
is taken from the internal data bus by the control register (U11) and connected to the $S-100$ bus vectored interrupt lines. Any one of the seven interrupt lines (VIO-VI7) may be selected by a jumper from pin 3 of U35 to the selected 1ine.

The interrupt mode may be cleared by the software program so the interrupts do not occur. The program clears the interrupt mode by setting bit 6 in any non-disk command to a one (see Table 4-2).

### 4.1.7 OUTPUT TO DISK DRIVE.

The AM-500 circuit board transfers disk write data and controls to the disk drive through the write data logic and drive control registers. See Table 2-3 for a complete list of the signals between the AM-500 circuit board and the disk drive.

### 4.1.7.1 DATA WRITE LOGIC.

The data write logic provides the double frequency encoded clock and data output necessary for the disk drive write circuitry. When the AM-500 is in the write mode, the WRITE signal is high, which sets the multiplexer in the CRC logic (U9) to select the write data (WDATA) from the FIFO memory stack to the CRC checker module U23. The serial data is then sent to the FIFO sync flip-flop U25 and also to the data write logic. Should a CRC error occur, CRCERR signal goes high and sets bit 7 of the drive status buffer.

When the FIFO memory is not empty, $\overline{\text { FEMPTY }}$ is high enabling the CRC checker and gating the serial data output of the FIFO through the write logic where it is combined with the $\overline{\text { ONE }}$ clock and applied to the $\overline{\text { WRITEDATA }}$ output to the disk drive. When the write current is flowing through the disk write head from the WGATE command, the CELL clock is also combined into the $\overline{\text { WRITEDATA }}$ command through OR gate U43. When the FIFO memory is empty, $\overline{\text { FEMPTY }}$ is low, gating CRC data to the $\overline{\text { WRITEDATA }}$ line.

### 4.1.7.2 DRIVE REGISTERS.

The disk drive registers control the selection and operation of the disk mechanism. Three flip-f1op registers (U11, U12, U13) perform these functions when enabled by the $\overline{R U N}$ signal.

Drive Register No. 1 controls the address data to the disk drive. The address data is contained on the internal data bus and is transferred into Drive Register No. 1 by $\overline{\text { CYLADR }}$ from the register select logic. When the software selects the cylinder address, $\overline{\text { CYLADR }}$ clocks the data from the internal data bus into Drive Register No. 1.

Drive Register No. 2 controls the disk and head select data to the disk drive. This data is contained on the internal data bus when the software selects the drive commands and is clocked into Drive Register No. 2 by the $\overline{\text { DRVSEL }}$ signal from the register select logic.

The Drive Control Register handles the gate signals for the disk read and write operations. This includes Write Gate (WGATE), Erase Gate (EGATE), Read Gate (RGATE), Cylinder Address Strobe (CYLSTR) and return to Zero Seek (RESTOR). In addition, this register generates the READ, WRITE, and DATA signals for internal control and the $\mathrm{S}-100$ bus interrupt signal. This data is clocked into the Drive Control Register by the $\overline{\text { CNTL }}$ signal from the register select logic.

### 4.1.8 INPUT FROM DISK DRIVE.

The AM-500 circuit board receives memory data and status from the disk drive through the data read logic and drive status register. See Table $2-3$ for a complete list of the signals between the circuit board and the disk drive.

### 4.1.8.1 DATA READ LOGIC.

The data read logic receives the separate clock and data outputs of the disk drive and applies them to the clock and preset inputs of a flip-flop. When the $\mathrm{AM}-500$ is not in the read mode, the READ signal holds this flip-flop in the clear state. The flip-flop output (RDATA) and the input clock ( $\overline{\text { RCLOCK }}$ ) are sent to the CRC logic and also to the FIFO memory. When the AM-500 is in the Read mode, the $\overline{\text { RCLOCK }}$ and RDATA are selected by the multiplexer in the CRC logic (U9) for checking by the CRC module U23 and also entered into the serial input of the FIFO memory gated by the FSYNC output of the CRC logic.
4.1.8.2 DRIVE STATUS BUFFER.

The Drive Status Buffer receives status information from the disk drive and also the error signal from the CRC checker. When the CPU interrogates the disk status from the AM-500 (see Tables 4-4 and 4-5), the $\overline{\text { DRVSTS }}$ output of the register select logic is low which transfers the status data from the disk drive to the internal data bus for transfer to the CPU. The data format is as follows:

| $\frac{1}{\text { BIT }}$ | STATUS |
| :--- | :--- |
| 0 | $\overline{\text { INDEX }}$ |
| 1 | $\overline{\text { SECTOR }}$ |
| 2 | $\overline{\text { ONCYLINDER }}$ |
| 3 | $\overline{\text { SEEKERROR }}$ |
| 4 | $\overline{\text { FAULT }}$ |
| 5 | $\overline{\text { DRIVEREADY }}$ |
| 6 | $\overline{\text { WRITEPROTECTED }}$ |
| 7 | $\overline{\text { CRCERR }}$ |

See Table 2-3 for a description of the disk interface signals.

### 4.1.9 FIFO MEMORY AND CONTROL.

The FIFO Memory and Control circuitry consists of two 4-bit First-In-First-Out memory modules connected for a stack of 16 8-bit bytes. See paragraph 4.2.3 for a detailed description of the module. The FIFO memory is used for both disk read and write operations. It converts the serial data from the disk to parallel data for internal processing and transfer, and converts parallel data to serial for disk write operations. The data is transferred between the FIFO and $1 K$ of the Random Access Memory (RAM) by the DMA controller. For memory full or memory empty conditions, the FIFO synchronizing logic makes the DMA service requests to the DMA controller to ensure that the DMA controller runs at the disk speed.
4.1.10 DMA BUFFER.

The DMA buffer is contained in the Random Access Memory (RAM) modules U5 and U6. A11 data transfers to or from the disk memory must pass through this buffer.

DMA requests (DREQ0 and DREQ1) are generated by the DMA request latches on U16. For DREQ0, the Data Write (DATWRT) command is received from the $I / O$ command register, setting the DREQ0 flip-flop. The acknowledge from the DMA controller ( $\overline{\mathrm{DACKO}}$ ) and the $I / O$ Read ( $\overline{\mathrm{IOR}}$ ) are combined to reset the DREQ0 flip-f1op by $\overline{\mathrm{DIOR}}$. DREQ1 flip-flop operates in a similar way; set by Data Read ( $\overline{\text { DATRD }}$ ) and clocked by $\overline{\text { DIOW. Both request }} \mathrm{flip-flops}$ are cleared by DMA Request Clear $\overline{\text { DRLCLR }}$.

### 4.2 CIRCUIT MODULE DESCRIPTION.

This section describes the operation of the individual circuit packages (DIPS) contained on the AM- 500 circuit board. Most of the processing is handled by the CPU and DMA controller modules, so these are described in detail. The control logic and interface modules are also described with logic and connection diagrams for each one.

### 4.2.1 CPU MICROPROCESSOR (U19).

The CPU microprocessor is a single DIP module that handles the data processing of the AM-500 circuit board.

Figure 4-2 is a block diagram of the CPU, and Figure 4-3 details the internal register configuration which contains 208 bits of Read/ Write memory that are accessible to the programmer. The registers include two sets of six general purpose registers that may be used individually as 8 -bit registers or as 16 -bit register pairs. There are also two sets of accumulator and flag registers. The programmer has access to either set of main or alternate registers through a group of exchange instructions. This alternate set allows foreground/background mode of operation or may be reserved for very fast Interrupt response. Each CPU also contains a 16 -bit stack pointer which permits simple implementation of multiple level interrupts, unlimited subroutine nesting and simplification of many types of data hand1ing.

The two 16 -bit index registers allow tabular data manipulation and easy implementation of relocatable code. The Refresh register provides for automatic, totally transparent refresh of external dynamic memories. The $I$ register is used in a powerful interrupt response mode to form the upper 8 bits of a pointer to an interrupt service address table, while the interrupting device supplies the lower 8 bits of the pointer. An indirect call is then made to this service address.


Figure 4-2. CPU Block Diagram


Figure 4-3. CPU Registers

Figure 4-4 shows the CPU pin in configuration and Table 4-7 contains a list of the CPU signals.


Figure 4-4. CPU Pin Configuration

Table 4-7. CPU Signal List

| SI GNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| $\mathrm{A} 0-\mathrm{A} 15$ <br> ADDRESS BUS |  | Tri-state output, active high. A0-A15 constitute a 16 -bit address bus. The address bus provides the address for memory (up to 64 K bytes) data exchanges and for $1 / 0$ device data exchanges. |
| D0-D7 <br> DATA BUS |  | Tri-state input/output, active high. D0-D7 constitute an 8-bit bidirectional data bus. The data bus is used for data exchanges with memory and I/O devices. |
| M1 MACHINE CYCLE ONE | 27 | Output, active low. $\overline{M 1}$ indicates that the current machine cycle is the OP code fetch cycle of an instruction execution. |
| $\overline{\text { MREQ }}$ <br> MEMORY REQUEST | 19 | Tri-state output, active low. The memory request signal indicates that the address bus holds a valid address for a memory read or memory write operation. |
| $\overline{\text { IORQ }}$ <br> INPUT/OUTPUT REQUEST | 20 | Tri-state output, active low. The $\overline{\text { IORQ }}$ signal indicates that the lower half of the address bus holds a valid I/O address for a I/O read or write operation. An IORQ signal is also generated when an interrupt is being acknowledged to indicate that an |

Table 4-7 (Cont.). CPU Signal List

| SIGNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
|  |  | interrupt response vector can be placed on the data bus. |
| $\overline{\mathrm{RD}}$ <br> MEMORY READ | 21 | Tri-state output, active low. $\overline{\mathrm{RD}}$ indicates that the CPU wants to read data from memory or an $1 / O$ device. The addressed $I / O$ device or memory should use this signal to gate data onto the CPU data bus. |
| $\overline{\mathrm{WR}}$ <br> MEMORY WRITE | 22 | Tri-state output, active low. $\overline{W R}$ indicates that the CPU data bus holds valid data to be stored in the addressed memory or $I / O$ device. |
| $\overline{\text { RFSH }}$ <br> REFRESH | 28 | Output, active low. $\overline{\text { RFSH }}$ indicates that the lower 7 bits of the address bus contain a refresh address for dynamic memories and the current $\overline{\text { MREQ }}$ signal should be used to do a refresh read to all dynamic memories. |
| $\overline{\text { HALT }}$ <br> HALT STATE | 18 | Output, active low. $\overline{\text { HALT }}$ indicates that the CPU has executed a $\overline{\text { HALT }}$ software instruction and is awaiting either a nonmaskable or a maskable interrupt (with the mask enabled) before operation can resume. While halted, the CPU executes NOP's to maintain memory refresh activity. |


| SI GNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| $\overline{\text { WAIT }}$ <br> WAIT | 24 | Input, active low. $\overline{\text { WAIT }}$ indicates to the CPU that the addressed memory or $I / O$ devices are not ready for a data transfer. The CPU continues to enter wait states for as long as this signal is active. |
| $\overline{\mathrm{INT}}$ <br> INTERRUPT REQUEST | 16 | Input, active low. The Interrupt Request signal is generated by I/O devices. A request will be honored at the end of the current instruction if the internal software controlled interrupt enable flip-flop (IFF) is enabled. |
| $\overline{\text { NMI }}$ <br> NON MASKABLE INTERRUPT | 17 | Input, active low. The non-maskable interrupt request line has a higher priority than $\overline{\mathrm{INT}}$ and is always recognized at the end of the current instruction, independent of the status of the interrupt enable flipflop. $\overline{N M I}$ automatically forces the CPU to restart to location 0066 H . |
| $\overline{\text { RESET }}$ | 26 | Input, active low. $\overline{\text { RESET }}$ initializes the CPU as follows: reset interrupt enable flip-flop, clear $P C$ and registers $I$ and $R$ and set interrupt to 8080 A mode. During reset time, the address and data bus go to a high impedance state |


| S I GNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
|  |  | and all control output signals go to the inactive state. |
| $\begin{aligned} & \overline{\text { BUSRQ }} \\ & \text { BUS REQUEST } \end{aligned}$ | 25 | Input, active low. The bus request signal has a higher priority than $\overline{\text { NMI }}$ and is always recognized at the end of the current machine cycle and is used to request the CPU address bus, data bus and tri-state output control signals to go to a high impedance state so that other devices can control these busses. |
| $\begin{aligned} & \overline{\text { BUSAK }} \\ & \text { BUS ACKNOWLEDGE } \end{aligned}$ | 23 | Output, active low. Bus acknowledges is used to indicate to the requesting device that the CPU address bus, data bus and tri-state control bus signals have been set to their high impedance state and the external device can now control these signals. |

### 4.2.1.1 INSTRUCTION OP-CODE FETCH.

The program counter content (PC) is placed on the address bus immediately at the start of the cycle. One half clock time later $\overline{M R E Q}$ goes active. The falling edge of $\overline{\text { MREQ }}$ can be used directly as a chip enable to dynamic memories. $\overline{\mathrm{RD}}$ when active indicates that the memory data should be enabled onto the CPU data bus. The CPU samples data with the rising edge of the clock state T3. Clock states T3 and $T 4$ of a fetch cycle are used to refresh dynamic memories while the CPU is internally decoding and executing the instruction. The
refresh control signal $\overline{\mathrm{RFSH}}$ indicates that a refresh read of all dynamic memories should be accomplished.

4.2.1.2 MEMORY READ OR WRITE CYCLES.

Illustrated here is the timing of memory read or write cycles other than an OP code fetch (M1 cycle). The $\overline{M R E Q}$ and $\overline{R D}$ signals are used exactly as in the fetch cycle. In the case of a memory write cycle, the $\overline{M R E Q}$ also becomes active when the address bus is stable so that it can be used directly as a chip enable for dynamic memories. The $\overline{W R}$ line is active when data on the data bus is stable so that it can be used directly as a R/W pulse to virtually any type of semiconductor memory.


### 4.2.1.3 INPUT OR OUTPUT CYCLES.

Illustrated here is the timing for an I/O read or I/O write operation. Notice that during $I / O$ operations a single wait state is automatically inserted (Tw*). The reason for this is that during I/O operations this extra state allows sufficient time for an $1 / 0$ port to decode its address and activate the $\overline{\text { WAIT }}$ line if a wait is required.


### 4.2.1.4 INTERRUPT REQUEST/ACKNOWLEDGE CYCLE.

The interrupt signal is sampled by the CPU with the rising edge of the last clock at the end of any instruction. When an interrupt is accepted, a special M1 cycle is generated. During this M1 cycle, the $\overline{I O R Q}$ signal becomes active (instead of $\overline{M R E Q}$ ) to indicate that the interrupting device can place an 8-bit vector on the data bus. Two wait states (Tw*) are automatically added to this cycle so that a ripple priority interrupt scheme, such as the one used in the peripheral controllers, can be easily implemented.

4.2.1.5 CPU INSTRUCTION SET.
The following is a summary of the CPU instruction set showing theassembly language mnemonic and the symbolic operation performedby the instruction. The instructions are divided into the fol-lowing categories:

8-bit loads
16-bit loads
Exchanges
Memory Block Moves
Memory Block Searches
8-bit arithmetic and logic 16-bit arithmetic

General purpose Accumulator

Miscellaneous Group
Rotates and Shifts
Bit Set, Reset and Test
Input and Output
Jumps
Ca11s
Restarts
Returns
\& Flag Operations
In the table the following terminology is used.
b $\quad \equiv$ a bit number in any 8 -bit register or memory location
cc $\equiv$ flag condition code
$\mathrm{NZ} \equiv$ non zero
Z $\equiv$ zero
NC $\equiv$ non carry
C $\equiv$ carry
PO $\equiv$ Parity odd or no over flow
PE $\equiv$ Parity even or over flow
P $\equiv$ Positive
M $\equiv$ Negative (minus)
d $\equiv$ any 8 -bit destination register or memory location
dd $\equiv$ any 16 -bit destination register or memory location
e $\quad \equiv 8$-bit signed 2 's complement displacement used in relative jumps and indexed addressing
L $\equiv 8$ special call locations in page zero. In decimal notation these are $0,8,16,24,32,40,48$ and 56
$\mathrm{n} \quad \equiv$ any 8 -bit binary number
$\mathrm{nn} \equiv$ any 16 -bit binary number
r $\equiv$ any 8 -bit general purpose register (A, B, C, D, E H. or L)
s $\equiv$ any 8-bit source register or memory location $\mathrm{s}_{\mathrm{b}} \quad \equiv$ a bit in a specific 8 -bit register or memory location
ss $\equiv$ any 16 -bit source register or memory location subscript "L" $\equiv$ the low order 8 bits of a 16 -bit register subscript "H" $\equiv$ the high order 8 bits of a 16 -bit register
() $\equiv$ the contents within the () are to be used as a pointer to a memory location or $\mathrm{I} / \mathrm{O}$ port number 8 -bit registers are A, B, C, D, E, H, L, I and R 16-bit register pairs are AF, BC, DE and HL 16-bit registers are SP, PC, IX and IY

Addressing Modes implemented include combinations of the following: Immediate Indexed Immediate extended Register Modified Page Zero Implied Relative Register Indirect Extended Bit

Tab1e 4-8. CPU Instruction Set


Table 4-8 (Cont.). CPU Instruction Set

| MEMORY BLOCK SEARCHES | Mnemonic | Symbolic Operation | Comments |
| :---: | :---: | :---: | :---: |
|  | CPI <br> CPIR <br> CPD <br> CPDR | $\begin{aligned} & \mathrm{A}-(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}+1 \\ & \mathrm{BC} \leftarrow \mathrm{BC}-1 \\ & \mathrm{~A}-(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}+1 \\ & \mathrm{BC} \leftarrow \mathrm{BC}-1, \text { Repeat } \\ & \text { until } \mathrm{BC}=0 \text { or } \mathrm{A}=(\mathrm{HL}) \\ & \mathrm{A}-(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}-1 \\ & \mathrm{BC} \leftarrow \mathrm{BC}-1 \\ & \dot{A}-(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}-1 \\ & \mathrm{BC} \leftarrow \mathrm{BC}-1, \text { Repeat } \\ & \text { until } \mathrm{BC}=0 \text { or } \mathrm{A}=(\mathrm{HL}) \end{aligned}$ | A-(HL) sets the flags only. $A$ is not affected |
| 8-BIT ALU | ADD s <br> ADC s <br> SUB s <br> SBC s <br> AND s <br> OR s <br> XOR s <br> CPs <br> INC d <br> DEC d | $\begin{aligned} & A \leftarrow A+s \\ & A \leftarrow A+s+C Y \\ & A \leftarrow A-s \\ & A \leftarrow A-s-C Y \\ & A \leftarrow A \wedge s \\ & A \leftarrow A \vee s \\ & A \leftarrow A \oplus s \\ & A-s \\ & d \leftarrow d+1 \\ & d \leftarrow d-1 \end{aligned}$ | CY is the carry flag $\begin{aligned} & s \equiv r, n,(H L) \\ & (I X+e),(I Y+e) \end{aligned}$ $\begin{aligned} & s=r, n(H L) \\ & (I X+e),(I Y+e) \\ & d=r,(H L) \\ & (I X+e),(I Y+e) \end{aligned}$ |
| 16-BIT ARITHMATIC | ADD HL, ss ADC HL, ss SBC HL, ss ADD IX. ss <br> ADD IY. ss <br> INC dd <br> DEC dd | $\begin{aligned} & \mathrm{HL} \leftarrow \mathrm{HL}+\mathrm{ss} \\ & \mathrm{HL} \leftarrow \mathrm{HL}+\mathrm{ss}+\mathrm{CY} \\ & \mathrm{HL} \leftarrow \mathrm{HL}-\mathrm{ss}-\mathrm{CY} \\ & \mathrm{IX} \leftarrow \mathrm{IX}+\mathrm{ss} \\ & \mathrm{IY} \leftarrow \mathrm{IY}+\mathrm{ss} \\ & \mathrm{dd} \leftarrow \mathrm{dd}+\mathrm{I} \\ & \mathrm{dd} \leftarrow \mathrm{dd}-\mathrm{I} \end{aligned}$ | $\begin{aligned} & \left\{\begin{array}{l} s s \equiv B C, D E \\ H L, S P \end{array}\right. \\ & s s \equiv B C, D E \\ & I X, S P \\ & s s \equiv B C, D E, \\ & I Y, S P \\ & d d \equiv B C, D E \\ & H L, S P, I X, I Y \\ & d d \equiv B C, D E \\ & H L, S P, I X, I Y \end{aligned}$ |
| GP ACC. AND FLAG | $\begin{aligned} & \text { DAA } \\ & \text { CPL } \\ & \text { NEG } \\ & \text { CCF } \\ & \text { SCF } \end{aligned}$ | Converts A contents into packed BCD following add or subtract. $\begin{aligned} & A \leftarrow \bar{A} \\ & A \leftarrow 00-A \\ & C Y \leftarrow \overline{C Y} \\ & C Y \leftarrow 1 \end{aligned}$ | Operands must be in packed BCD format |

Table 4-8 (Cont.). CPU Instruction Set

|  | Mnemonic | Symbolic Operation | Comments |
| :---: | :---: | :---: | :---: |
| MISCELLANEOUS | NOP <br> HALT <br> DI <br> EI <br> IM 0 <br> IM 1 <br> IM 2 | No operation <br> Halt CPU <br> Disable Interrupts <br> Enable Interrupts <br> Set interrupt mode 0 <br> Set interrupt mode 1 <br> Set interrupt mode 2 | 8080A mode Call to 0038 H Indirect Call |
| ROTATES AND SHIFTS | RLC s <br> RL s <br> RRC s <br> RR s <br> SLA s <br> SRA s <br> SRL s <br> RLD <br> RRD |  | $\begin{aligned} & \mathrm{s} \equiv \mathrm{r},(\mathrm{HL}) \\ & (\mathrm{IX}+\mathrm{e}),(\mathrm{IY}+\mathrm{e}) \end{aligned}$ |
| BIT S, R, AND T | BIT b, s SET b, s RES b, s | $\begin{aligned} & \mathrm{z} \leftarrow \overline{s_{b}} \\ & \mathrm{~s}_{\mathrm{b}} \leftarrow 1 \\ & \mathrm{~s}_{\mathrm{b}} \leftarrow 0 \end{aligned}$ | $\begin{aligned} & \mathrm{Z} \text { is zero flag } \\ & \mathrm{s} \equiv \mathrm{r},(\mathrm{HL}) \\ & (\mathrm{IX}+\mathrm{e}),(\mathrm{IY}+\mathrm{e}) \end{aligned}$ |
| INPUT AND OUTPUT | $\begin{aligned} & \text { IN A, (n) } \\ & \text { IN r, (C) } \\ & \text { INI } \\ & \text { INIR } \\ & \text { IND } \\ & \text { INDR } \end{aligned}$ | $\begin{aligned} & A \leftarrow(n) \\ & r \leftarrow(C) \\ & (H L) \leftarrow(C), H L \leftarrow H L+1 \\ & B \leftarrow B-1 \\ & (H L) \leftarrow(C), H L \leftarrow H L+1 \\ & B \leftarrow B-1 \\ & R e p e a t ~ u n t i l ~ B=0 \\ & (H L) \leftarrow(C), H L \leftarrow H L-1 \\ & B \leftarrow B-1 \\ & (H L) \leftarrow(C), H L \leftarrow H L-1 \\ & B \leftarrow B-1 \\ & \text { Repeat until } B=0 \end{aligned}$ | Set flags |


| INPUT AND OUTPUT | Mnemonic | Symbolic Operation | Comments |
| :---: | :---: | :---: | :---: |
|  | OUT(n), A <br> OUT(C), r <br> OUTI <br> OTIR <br> OUTD <br> OTDR | (n) $\leftarrow \mathrm{A}$ <br> (C) $\leftarrow \mathrm{r}$ <br> $(\mathrm{C}) \leftarrow(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}+1$ <br> $\mathrm{B} \leftarrow \mathrm{B}-1$ <br> $(\mathrm{C}) \leftarrow(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}+1$ <br> $\mathrm{B} \leftarrow \mathrm{B}-1$ <br> Repeat until $B=0$ <br> $(\mathrm{C}) \leftarrow(\mathrm{HL}), \mathrm{HL} \leftarrow \mathrm{HL}-1$ <br> B $\leftarrow \mathbf{B}-1$ <br> $(\mathrm{C}) \leftarrow(\mathrm{HL}) . \mathrm{HL} \leftarrow \mathrm{HL}-\mathrm{I}$ <br> $\mathrm{B} \leftarrow \mathrm{B}-1$ <br> Repeat until $\mathrm{B}=0$ |  |
| JUMPS | JP nn <br> JP cc, nn <br> JRe <br> JR kk.e <br> JP (ss) <br> DJNZ e | PC $\leftarrow \mathrm{nn}$ <br> If condition ic is true $\mathrm{PC} \leftarrow \mathrm{nn}$, else continue $\mathrm{PC} \leftarrow \mathrm{PC}+\mathrm{e}$ <br> If condition kk is true $\mathrm{PC} \leftarrow \mathrm{PC}+\mathrm{e}$. else continue $\mathrm{PC} \leftarrow \mathrm{ss}$ $B \leftarrow B-1$, if $B=0$ continue, else $\mathbf{P C} \leftarrow \mathrm{PC}+\mathrm{e}$ | $\begin{aligned} & \mathrm{cc} \begin{cases}\mathrm{NZ} & \mathrm{PO} \\ \mathrm{Z} & \mathrm{PE} \\ \mathrm{NC} & \mathrm{P} \\ \mathrm{C} & \mathrm{M}\end{cases} \\ & \mathrm{kk} \begin{cases}\mathrm{NZ} & \mathrm{NC} \\ \mathrm{Z} & \mathrm{C}\end{cases} \\ & \mathrm{ss}=\mathrm{HL}, \mathrm{IX}, \mathrm{IY} \end{aligned}$ |
| CALLS | CALL nn <br> CALL cc, nn | $\begin{aligned} & (\mathrm{SP}-1) \leftarrow \mathrm{PC}_{\mathrm{H}} \\ & (\mathrm{SP}-2) \leftarrow \mathrm{PC}_{\mathrm{L}} \cdot \mathrm{PC} \leftarrow \mathrm{nn} \end{aligned}$ <br> If condition cc is false continue. else same as CALL nn | cc $\begin{cases}\mathrm{NZ} & \mathrm{PO} \\ \mathrm{Z} & \mathrm{PE} \\ \mathrm{NC} & \mathrm{P} \\ \mathrm{C} & \mathrm{M}\end{cases}$ |
| RESTARTS | RST L | $\begin{aligned} & (\mathrm{SP}-1) \leftarrow \mathrm{PC}_{\mathrm{H}} \\ & \left(\mathrm{SP}_{-2}\right) \leftarrow \mathrm{PC}_{\mathrm{L}} \cdot \mathrm{PC}_{\mathrm{H}} \leftarrow 0 \\ & \mathrm{PC}_{\mathrm{L}} \leftarrow \mathrm{~L} \end{aligned}$ |  |
| RETURNS | RET <br> RET cc <br> RETI <br> RETN | $\begin{aligned} & \mathrm{PC}_{\mathrm{L}} \leftarrow(\mathrm{SP}), \\ & \mathrm{PC}_{\mathbf{H}} \leftarrow(\mathrm{SP}+1) \end{aligned}$ <br> If condition ic is false continue, else same as RET <br> Return from interrupt. same as RET <br> Return from nonmaskable interrupt | $\cdots \begin{cases}\mathrm{NZ} & \mathrm{PO} \\ Z & \mathrm{PE} \\ \mathrm{NC} & \mathrm{P} \\ \mathrm{C} & \mathrm{M}\end{cases}$ |

4.2.2 DMA CONTROLLER (U20).

This device is a multimode Direct Memory Access (DMA) controller for microprocessor systems. It enhances system performance by allowing other devices to directly transfer information to or from memory or to transfer data from one memory to another. Figure 4-5 shows the DMA controller connections and Table 4-9 lists these signals with their functions.

The three basic transfer modes allow programmability of the types of DMA seryice by the user. Each channel can be individually programmed to Autoinitialize to its original condition following an End of Process ( $\overline{\mathrm{EOP}}$ ).

Each channel has a full 64 K address and word count capability. An external $\overline{E O P}$ signal can terminate a DMA or memory-to-memory transfer. This is useful for block search or compare operations using external comparators or for intelligent peripherals to abort erroneous services.

## CONNECTION DIAGRAM Top View



Figure 4-5. DMA Controller Connections

Table 4-9. DMA Controller Signal List

| SI GNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| VCC | 31 | +5 Volt Supply |
| VSS | 20 | Ground |
| CLK <br> CLOCK, INPUT | 12 | This input controls the internal operations of the DMA Controller and its rate of data transfers. The input may be driven at up to 3 MHz for the standard DMA Controller and up to 4 MHz for the DMA Controller. |
| $\overline{\mathrm{CS}}$ CHIP SELECT, INPUT | 11 | Chip Select is an active low input used to select the DMA Controller as an I/O device during the Idle cycle. This allows CPU communication on the data bus. |
| RESET RESET, INPUT | 13 | Reset is an asynchronous active high input which clears the Command, Status, Request and Temporary registers. It also clears the first/ last flip/flop and sets the Mask register. Following a Reset the device is in the Idle cycle. |
| READY READY, INPUT | 6 | Ready is an asynchronous input used to extend the memory read and write pulses from the DMA Controller to accommodate slow memories or $1 / 0$ peripheral devices. |

Table 4-9 (Cont.). DMA Controller Signal List

| SIGNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| $\begin{aligned} & \text { HACK } \\ & \text { HOLD ACKNOWLEDGE, } \\ & \text { INPUT } \end{aligned}$ | 7 | The active high Hold Acknowledge from the CPU indicates that control, of the system buses has been relinquished. |
| DREQ 0 -DREQ 3 DMA REQUEST, INPUT | $\begin{aligned} & 19, \\ & 18, \\ & 17, \\ & 16 \end{aligned}$ | The DMA Request lines are individual asynchronous channe1 request inputs used by peripheral circuits to obtain DMA service. In Fixed Priority, DREQ0 has the highest priority and DREQ3 has the lowest priority. A request is generated by activating the DREQ line of a channe1. DACK will acknowledge the recognition of the DREQ signal. Polarity of DREQ is programmable. Reset initializes these lines to active high. |
| $\begin{aligned} & \text { DB0-DB } 7 \\ & \text { DATA BUS } \\ & \text { INPUT / OUTPUT } \end{aligned}$ | $\begin{aligned} & 30, \\ & 29, \\ & 28, \\ & 27, \\ & 26, \\ & 23, \\ & 22, \\ & 21 \end{aligned}$ | The Data Bus lines are bidirectional three-state signals connected to the system data bus. The outputs are enabled in the Program Condition during the $I / O$ Read to output the contents of an Address register, a Status register, the Temporary register or a Word Count register to the CPU. The outputs are disabled and the inputs are read during an I/O Write cycle when the CPU is programming the DMA controller control registers. During DMA cycles, the most significant eight bits of |

Table 4-9 (Cont.). DMA Controller Signal List

| SIGNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
|  |  | the address are output onto the data bus to be strobed into an external latch by ADSTB. In memory-to-memory operations, data from the memory comes into the DMA Controller on the data bus during the read-frommemory transfer. In the write-tomemory transfer, the data bus outputs place the data into the new memory location. |
| $\overline{\mathrm{IOR}}$ <br> I/O READ, INPUT/OUTPUT | 1. | I/O Read is a bidirectiona1, active low, three-state line. In the Idle cycle, it is an input control signal used by the CPU to read the control registers. In the Active cycle, it is an output control signal used by the DMA Controller to access data from a peripheral during a DMA Write transfer. |
| $\overline{\overline{I O W}}$ <br> I/O WRITE, INPUT/OUTPUT | 2 | I/O Write is a bidirectional, active low, three-state line. In the idle cycle, it is an input control signal used by the CPU to load information into the DMA Controller. In the Active cycle, it is an output control signal used by the DMA Controller to load data to the peripheral during a DMA Read transfer. |

Table 4-9 (Cont.). DMA Controller Signal List

| SIGNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| $\overline{\text { EOP }}$ <br> END OF PROCESS, INPUT/OUTPUT | 36 | $\overline{\mathrm{EOP}}$ is an active low, bidirectional signal. Information concerning the completion of DMA services is available at the bi-directional End of Process ( $\overline{E O P}$ ) pin. The DMA Controller allows an external signal to terminate an active DMA service. <br> This is accomplished by pulling the $\overline{\mathrm{EOP}}$ input low with an external $\overline{\mathrm{EOP}}$ signal. The DMA Controller also generates a pulse when the terminal count (TC) for any channel is reached. This generates an $\overline{E O P}$ signal which is output through the $\overline{\mathrm{EOP}}$ line. The reception of $\overline{E O P}$, either internal or external, will cause the DMA Controller to terminate the service, reset the request; and if Autoinitialize is enabled, to write the base registers to the current registers of that channel. The mask bit and TC bit in the status word will be set for the currently active channel by $\overline{\mathrm{EOP}}$ unless the channel is programmed for Autoinitialize. In that case, the mask bit remains clear. During memory-to-memory transfers, $\overline{\mathrm{EOP}}$ will be output when the TC for channel 1 occurs. |

Table 4-9 (Con't.). DMA Controller Signal List

| SIGNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| $\mathrm{A} 0-\mathrm{A} 3$ <br> ADDRESS, INPUT/OUTPUT | $\begin{aligned} & 32, \\ & 33, \\ & 34, \\ & 35 \end{aligned}$ | The four least significant address lines are bidirectional, three-state signals. In the Idle cycle, they are inputs and are used by the DMA Controller to address the control register to be loaded or read. In the Active cycle, they are outputs and provide the lower 4 bits of the output address. |
| $\mathrm{A} 4-\mathrm{A} 7$ <br> ADDRESS, OUTPUT | $\begin{aligned} & 37, \\ & 38, \\ & 39, \\ & 40 \end{aligned}$ | The four most significant address lines are three-state outputs and provide four bits of address. These lines are enabled only during the DMA service. |
| HREQ <br> HOLD REQUEST, OUTPUT | 10 | This is the Hold Request to "e CPU and is used to request control of the system bus. If the corresponding mask bit is clear, the presence of any valid DREQ causes the DMA controller to issue the $H R E Q$. |
| DACK0-DACK3 <br> DMA ACKNOWLEDGE, OUTPUT | $\begin{aligned} & 25, \\ & 24, \\ & 17, \\ & 16 \end{aligned}$ | DMA Acknowledge is used to notify the individual peripherals when one has been granted a DMA cycle. The sense of these lines is programmable. <br> Reset initializes them to active low. |

Table 4-9 (Cont.). DMA Controller Signal List

| SI GNAL | PIN | FUNCTION |
| :---: | :---: | :---: |
| AEN <br> ADDRESS ENABLE, OUTPUT | 9 | The Address Enable is an active high level used to enable the output of the external latch which holds the upper byte of address, and to disable the system bus during the DMA cycle. Note that during DMA transfers, HACK and AEN should be used to deselect all other I/O peripherals which may erroneously be accessed as programmed I/O during the DMA operation. The DMA Controller automatically deselects itself during DMA transfer. |
| ADSTB <br> ADDRESS STROBE, OUTPUT | 8 | The active high Address Strobe is used to strobe the upper address byte into an external latch. |
| $\overline{\text { MEMR }}$ <br> MEMORY READ, OUTPUT | 3 | The Memory Read signal is an active low, three-state output used to access data from the selected memory location during a DMA Read or a memory-to-memory transfer. |
| $\overline{\text { MEMW }}$ <br> MEMORY WRITE, OUTPUT | 4 | The Memory Write signal is an active low, three-state output used to write data to the selected memory location during a DMA Write or a memory-tomemory transfer. |

4.2.2.1 DMA CONTROLLER FUNCTIONAL DESCRIPTION.

The DMA Controller block diagram shown in Figure 4-6 includes the major logic blocks and all of the internal registers. The data interconnection paths are also shown. Not shown are the various control signals between the blocks. The DMA Controller contains 344 bits of internal memory in the form of registers. Figure 4-10 1ists these registers by name and shows the size of each. A detailed description of the registers and their functions can be found under Register Description.

Table 4-10. DMA Controller Internal Registers

| Name | Size | Number |
| :--- | :---: | :---: |
| Base Address Registers | 16 bits | 4 |
| Base Word Count Registers | 16 bits | 4 |
| Current Address Registers | 16 bits | 4 |
| Current Word Count Registers | 16 bits | 4 |
| Temporary Address Register | 16 bits | 1 |
| Temporary Word Count Register | 16 bits | 1 |
| Status Register | 8 bits | 1 |
| Command Register | 8 bits | 1 |
| Temporary Register | 8 bits | 1 |
| Mode Registers | 6 bits | 4 |
| Mask Register | 4 bits | 1 |
| Request Register | 4 bits | 1 |



Figure 4-6. DMA Controller Block Diagram

The DMA Controller contains three basic blocks of control logic. The Timing Control block generates internal timing and external control signals for the DMA Controller. The Program Command Control block decodes the various commands given to the DMA Controller by the microprocessor prior to servicing a DMA Request. It also decodes the Mode Control word used to select the type of DMA during the servicing. The Priority Encoder block resolves priority contention between DMA channels requesting service simultaneously.

The Timing Control block derives internal timing from the clock input.

### 4.2.2.2 DMA OPERATION

The DMA Controller operates in two major cycles. These are called Idle and Active cycles. Each device cycle is made up of a number of states. The DMA Controller can assume seven separate states, each composed of one full clock period. State 1 (S1) is the inactive state. It is entered when the DMA Controller has no valid DMA requests pending. While in S1, the DMA controller is inactive but may be in the Program Condition, being programmed by the processor. State $0(\mathrm{~S} 0)$ is the first state of a DMA service. The DMA Controller has requested a hold, but the processor has not yet returned an acknowledge. An acknowledge from the CPU signals that transfers may begin. S1, S2, S3 and S4 are working states of the DMA service. If more time is needed to complete a transfer than is available with normal timing, wait states (SW) can be inserted between S2 or S3 and S4 by the use of the Ready line on the DMA Controller.

Memory-to-memory transfers require a read-from and a write-tomemory to complete each transfer. The states, which resemble the normal working states, use two digit numbers for identification. Eight states are required for a single transfer. The first four
states (S11, S12, S13, S14) are used for the read-from-memory half and the last four states (S21, S22, S23 and S24) for the write-to-memory half of the transfer.

### 4.2.2.3 IDLE CYCLE.

When no channel is requesting service, the DMA Controller will enter the idle cycle and perform "SI" states. In this cycle, the DMA Controller samples the DREQ lines every clock cycle to determine if any channel is requesting a DMA service. The device also samples $\overline{C S}$, looking for an attempt by the microprocessor to write or read the internal registers of the DMA Controller. When $\overline{C S}$ is low and HREQ is low, the DMA Controller enters the Program Condition. The CPU can now establish, change or inspect the internal definition of the part by reading from or writing to the internal registers. Address lines A0-A3 are inputs to the device and select which registers are to be read or written. The $\overline{I O R}$ and $\overline{I O W}$ lines are used to select and time reads or writes. Due to the number and size of the internal registers, an internal flip/flop is used to generate an additional bit of address. This bit is used to determine the upper or lower byte of the 16 -bit Address and Word Count registers. The flip/flop is reset by Master Clear or Reset. A separate software command can a.so reset this flip/flop.

Special software commands can be executed by the DMA Controller in the Program Condition. These commands are decoded as sets of addresses with $\overline{\mathrm{CS}}$ and $\overline{\mathrm{IOW}}$. The commands do not make use of the data bus. Instructions include Clear First/Last Flip/Flop and Master Clear.

### 4.2.2.4 ACTIVE CYCLE

When the DMA Controller is in the Idle cycle and a channel requests a DMA service, the device outputs a HREQ to the microprocessor and enters the Active cycle. It is in this cycle that the DMA service takes place in one of three modes:

Single Transfer Mode. In Single Transfer mode, the device is programmed to make one transfer only. The word count is decremented and the address decremented or incremented following each transfer. When the word count goes to zero, a Terminal Count (TC) causes an Autoinitialize if the channel has been programmed to do so.

DREQ must be held active until DACK becomes active in order to be recognized. If $D R E Q$ is held active throughout the single transfer, HREQ goes inactive and releases the bus to the system. It will again go active and upon receipt of a new HACK, another single transfer is performed. In $8080 \mathrm{~A} / 9080 \mathrm{~A}$ systems, this will ensure one full machine cycle execution between DMA transfers.

Block Transfer Mode. In Block Transfer mode, the device is activated by DREQ to continue making transfers during the service until a TC, caused by the word count going to zero, or an external End of Process (EOP) is encountered. DREQ need only be held active until DACK becomes active. Again, an Autoinitialize occurs at the end of the service if the channel has been programmed for it.

Demand Transfer Mode. In Demand Transfer mode, the device is programmed to continue making transfers until a TC or external $\overline{E O P}$ is encountered or until DREQ goes inactive. Thus transfers may continue until the $I / O$ device has exhausted its data capacity. After the $I / O$ device has had a chance to catch up, the DMA service is re-established by means of a DREQ. During the time between services when the microprocessor is allowed to operate, the intermediate values of address and word count are stored in the DMA Controller Current Address and Current Word Count registers. Only an $\overline{E O P}$ can cause an Autoinitialize at the end of the service. $\overline{E O P}$ is generated either by $T C$ or by an external signal.

### 4.2.2.5 TRANSFER TYPES.

Each of the three active transfer modes can perform three different types of transfers. These are Read, Write and Verify. Write transfers move data from an I/O device to the memory by activating $\overline{M E M W}$ and $\overline{I O R}$. Read transfers move data from memory to an $\mathrm{I} / 0$ device by activating MEMR and IOW. Verify transfers are pseudo transfers. The DMA Controller operates as in Read or Write transfers generating addresses, and responding to $\overline{E O P}$, etc. However, the memory and $I / O$ control lines all remain inactive.

Memory-to-Memory. To perform block moves of data from one memory address space to another with a minimum of program effort and time, the DMA Controller includes memory-to-memory transfer feature. Programming a bit in the Command register selects channels 0 and 1 to operate as memory-to-memory transfer channels. The transfer is initiated by setting the software DREQ for channe1 0 . The DMA Controller requests a DMA service in the normal manner. After HACK is true, the device, using eight-state transfers in Block Transfer mode, reads data from the memory. The channel 0 Current Address register is the source for the address used and is decremented or incremented in the normal manner. The data byte read from the memory is stored in the DMA Controller internal Temporary register. Channel 1 then writes the data from the Temporary register to memory using the address in its Current Address register and incrementing or decrementing it in the normal manner. The channe1 1 Current Word Count is decremented. When the word count of channel 1 goes to zero, a TC is generated causing an $\overline{\mathrm{EOP}}$ output terminating the service.

Channel 0 may be programmed to retain the same address for all transfers. This allows a single word to be written to a block of memory.

The DMA Controller responds to external $\overline{\mathrm{EOP}}$ signals during memory-to-memory transfers. Data comparators in block search schemes
may use this input to terminate the service when a match is found.

Autoinitialize. By programming a bit in the Mode register, a channel may be set up as an Autoinitialize channel. During Autoinitialization, the original values of the Current Address and Current Word Count registers are automatically restored from the Base Address and Base Word Count registers of that channel following $\overline{\text { EOP. }}$. The base registers are loaded simultaneously with the current registers by the microprocessor and remain unchanged throughout the DMA service. The mask bit is not set when the channel is in Autoinitialize. Following Autoinitialize, the channel is ready to perform another service without CPU intervention.

Priority. The DMA Controller has two types of priority encoding available as software selectable options. The first is Fixed Priority which fixes the channels in priority order based upon the descending value of their number. The channel with the lowest priority is 3 followed by 2, 1 and the highest priority channe1, 0 . After the recognition of any one channel for service, the other channels are prevented from interfering with that service until it is complete.

The second scheme is Rotating Priority. The last channel to get service becomes the lowest priority channel with the others rotating accordingly.


With Rotating Priority in a single chip DMA system, any device requesting service is guaranteed to be recognized after no more than three higher priority services havé occurred. This prevents any one channel from monopolizing the system.

Compressed Timing. In order to achieve even greater throughput where system characteristics permit, the DMA Controller can compress the transfer time to two clock cycles. State S 3 is used to extend the access time of the read pulse. By removing state $S 3$, the read pulse width is made equal to the write pulse width; and a transfer consists only of state $S 2$ to change the address and state $S 4$ to perform the read/write. S1 states still occur when A8-A15 need updating (see Address Generation).

Address Generation. In order to reduce pin count, the DMA Controller multiplexes the eight higher order address bits on the data lines. State S 1 is used to output the higher order address bits to an external latch from which they may be placed on the address bus. The falling edge of Address Strobe (ADSTB) is used to load these bits from the data lines to the latch. Address Enable (AEN) is used to enable the bits onto the address bus through a three-state enable. The lower order address bits are output direct1y. Lines A0-A7 should be connected to the address bus.

During Block and Demand Transfer mode services which include multiple transfers, the addresses generated are sequentia . For many transfers, the data held in the external address latch remains the same. This data need only change when a carry or borrow from A7 to A8 takes place in the normal sequence of addresses. To save time and speed transfers, the DMA Controller executes S 1 states on 1 y when updating of A8-A15 in the latch is necessary. This means for long services, S 1 states may occur only once every 256 transfers, a savings of 255 clock cycles for each 256 transfers.

### 4.2.2.6 REGISTER DESCRIPTION.

Current Address Register. Each channel has a 16-bit Current Address register. This register holds the value of the address used during DMA transfers. The address is automatically incre-
mented or decremented after each transfer, and the intermediate values of the address are stored in the Current Address register during the transfer. This register is written or read by the microprocessor in successive 8-bit bytes. It may also be reinitialized by an Autoinitialize back to its original value. Autoinitialization takes place only after an $\overline{\mathrm{EOP}}$.

Current Word Count Register. Each channe1 has a 16-bit Current Word Count register. This register holds the number of transfers to be performed. The word count is decremented after each transfer. The intermediate value of the word count is stored in the register during the transfer. When the value in the register goes to zero, a TC is generated. This register is loaded or read in successive 8 -bit bytes by the microprocessor in the Program Condition. Following the end of a DMA service,it may also be reinitialized by an Autoinitialize back to its original value. Autoinitialize can occur only when an $\overline{\mathrm{EOP}}$ occurs.

Base Address and Base Word Count Registers. Each channel has a pair of Base Address and Base Word Count registers. These 16-bit registers store the original values of their associated current registers. During Autoinitialize, these values are used to restore the current registers to their original values. The base registers are written simultaneously with their corresponding current register in 8-bit bytes in the Program Condition by the microprocessor. These registers cannot be read by the microprocessor.

Command Register. This 8-bit register controls the operation of the DMA Controller. It is programmed by the microprocessor in the Program Condition and is cleared by Reset. See Figure 4-7 for the function of the command bits and Table 4-11 for address coding.


Figure 4-7. DMA Controller Command Bits

Table 4-11. DMA Controller Word Count and Address Register Command Codes

| Channel | Register | Operation | Signals |  |  |  |  |  |  | Internal Flip/Flop | $\begin{aligned} & \text { Data Bus } \\ & \text { DB0-DB7 } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\overline{\text { CS }}$ | $\overline{10 R}$ | IOW | A3 | A2 | A1 | AO |  |  |
| 0 | Base \& Current | Write | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | A0.A7 |
|  | Address |  | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | A8-A15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | A0-A7 |
|  | Address |  | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | A8-A15 |
|  | Base \& Current | Write | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | W0.W7 |
|  | Word Count |  | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | W8-W15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 1 | W8-W15 |
| 1 | Base \& Current | Write | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 | A0.A7 |
|  | Address |  | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 1 | A8-A15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | A0-A7 |
|  | Address |  | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 | A8-A15 |
|  | Base \& Current | Write | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 1 | W8-W15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | W8-W15 |
| 2 | Base \& Current | Write | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 | A0-A7 |
|  | Address |  | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1 | A8-A15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 | A0-A7 |
|  | Address |  | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 1 | A8-A15 |
|  | Base \& Current | Write | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 1 | W8-W15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 1 | W8-W15 |
| 3 | Base \& Current | Write | 0 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | A0-A7 |
|  | Address |  | 0 | 1 | 0 | 0 | 1 | 1 | 0 | 1 | A8-A15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 0 | AO-A7 |
|  | Address |  | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 | A8-A15 |
|  | Base \& Current | Write | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1 | W8-W15 |
|  | Current | Read | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 0 | W0-W7 |
|  | Word Count |  | 0 | 0 | 1 | 0 | . 1 | 1 | 1 | 1 | W8-W15 |

Mode Register. Each channel has a 6-bit Mode register associated with it. When the register is being written to by the microprocessor in the Program Condition, bits 0 and 1 determine which channel Mode register is to be written. See Figure 4-8 for the mode register bit configuration.


Figure 4-8. DMA Controller Mode Register Bits

Request Register. The DMA Controller can respond to requests for DMA service which are initiated by software as well as by a DREQ. Each channel has a request bit associated with it in the four bit Request register. These are nonmaskable and subject to prioritization by the Priority Encoder network. Each register bit is set or reset separately under software control or is cleared upon generation of a TC or external EOP. The entire register is cleared by a Reset. To set or reset a bit, the software loads the proper form of the data word. See Figure 4-9 for the request register bit configuration and Table 4-11 for address coding.


Figure 4-9. DMA Controller Request Register Bits

Software requests will be serviced only if the channel is in Block mode. When initiating a memory-to-memory transfer, the software request for channel 0 should be set.

Mask Register. Each channel has associated with it a mask bit which can be set to disable the incoming DREQ. Each mask bit is set when its associated channel produces an $\overline{E O P}$ if the channel is not programmed for Autoinitialize. Each bit of the four bit Mask register may also be set or cleared separately under software control. The entire register is also set by a Reset. This disables all DMA requests until a clear Mask register instruction allows them to occur. The instruction to separately set or clear the mask bits is similar in form to that used with the Request register. See Figure 4-10 for the mask register bit configuration and Table 4-12 for Register instruction codes.


A11 four bits of the Mask register may also be written with a single command.


Figure 4-10. DMA Controller Mask Register Bits

Table 4-12. DMA Controller Register Codes

| Register | Operation | Signals |  |  |  |  |  |  |  |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\overline{\text { CS }}$ | $\overline{\text { IOR }}$ | $\overline{\text { OWW }}$ | A3 | A2 | A1 | A0 |  |
| Command | Write | 0 | 1 | 0 | 1 | 0 | 0 | 0 |  |
| Mode | Write | 0 | 1 | 0 | 1 | 0 | 1 | 1 |  |
| Request | Write | 0 | 1 | 0 | 1 | 0 | 0 | 1 |  |
| Mask | Set/Reset | 0 | 1 | 0 | 1 | 0 | 1 | 0 |  |
| Mask | Write | 0 | 1 | 0 | 1 | 1 | 1 | 1 |  |
| Temporary | Read | 0 | 0 | 1 | 1 | 1 | 0 | 1 |  |
| Status | Read | 0 | 0 | 1 | 1 | 0 | 0 | 0 |  |

Status Register. The Status register is available to be read out of the DMA Controller by the microprocessor. It contains information about the status of the device at that point. This information includes which channels have reached a terminal count and which channels have pending DMA requests. Bits 0-3 are set every time a TC is reached by that channel. These bits are cleared upon Reset and on each Status Read. Bits 4-7 are set whenever their corresponding channel is requesting service. See Figure. 4-11 for status register bit configuration.


Figure 4-11. DMA Controller Status Register Bits

Temporary Register. The Temporary register is used to hold data during memory-to-memory transfers. Following the completion of the transfers, the last word moved can be read by the microprocessor in the Program Condition. The Temporary register always contains the last byte transferred in the previous memory-to-memory operation, unless cleared by a Reset.

Software Commands. These are additional special software commands which can be executed in the Program Condition. They do not depend on any specific bit pattern on the data bus. The two software commands are:

Clear First/Last F1ip/F1op: This command is executed prior to writing or reading new address or word count information to the DMA Controller. This initializes the flip/flop to a known state so that subsequent accesses to register contents by the microprocessor will address upper and lower bytes in the correct sequence.

Master Clear: This software instruction has the same effect as the hardware Reset. The Command, Status, Request, Temporary, and Internal First/Last Flip/F1op registers are cleared and the Mask register is set. The DMA Controller enters the Idle cycle. Table 4-13 lists the address codes for the software commands.

Table 4-13. DMA Controller Software Command Codes

| Operation | Registers Affected | Signals |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\overline{\text { CS }}$ | $\overline{\text { IOR }}$ | IOW | A3 | A2 | A1 | AO |
| Clear FF | Internal <br> First/Last <br> Flip/Flop | 0 | 1 | 0 | 1 | 1 | 0 | 0 |
| Master Clear | Clear: <br> Command <br> Status <br> Request <br> Temporary <br> Internal <br> First/Last <br> Flip/Flop <br> Set: Mask | 0 | 1 | 0 | 1 | 1 | 0 | 1 |

```
4.2.3 FIRST-IN FIRST-OUT (FIFO) BUFFER MEMORY (U7, U8).
```

This device is an expandable fall-through type, high-speed First-In First-Out (FIFO) buffer memory optimized for high speed disc or tape controllers and communications buffers. It is organized as 16 words by four bits and may be expanded to any number, of words or any number of bits (in multiples of 4). Data may be entered or extracted asynchronously in serial or parallel. The device has tri-state outputs. See Figure 4-12 for the FIFO Buffer connections.


CONNECTION DIAGRAM DIP (TOP VIEW)

| IR5 |  |
| :---: | :---: |
| $=\square^{1}$ | On: |
| $\square D_{0}$ | $u_{5}$ |
| - $\square^{\circ}$ | $\mathrm{c}_{0}$ |
| $\mathrm{O}_{2}$ | c. |
| ${ }_{6} \mathrm{D}_{3}$ | $\mathrm{a}_{2}$ |
| $\mathrm{D}_{5}$ | $\mathrm{O}_{3}$ |
| CPSI | to |
| 位 | $\overline{\text { CPSO }}$ |
| 10 TTS | Ofs |
| $\overline{M B}$ | Tos |
| 12 Gno | ror |

Figure 4-12. FIFO Buffer Connections

### 4.2.3.1 FUNCTIONAL DESCRIPTION.

As shown in the Block Diagram in Figure 4-13, the device consists of three parts:

1. An input Register with Parallel and Serial Data Inputs as well as control inputs and outputs for input handshaking and expansion.
2. A 4-bit wide, 14-word deep Fal1-Through Stack with selfcontained control logic.
3. An Output Register with Parallel and Serial Data Outputs as well as control inputs and outputs for output handshaking and expansion.

Since these three sections operate asynchronously and almost independently, they will be described separately in the following paragraphs.


Figure 4-13. FIFO Block Diagram

### 4.2.3.2 INPUT REGISTER (DATA ENTRY).

The Input Register can receive data in either bit-serial or in 4-bit parallel form, store it until it is sent to the Fall-Through Stack and generate and accept the necessary status and control signals.

Figure 4-14 is a conceptual logic diagram of the input section. As described later, this 5 -bit register is initialized by setting the F3 F1ip-F1op and resetting the other flip-flops. The $\bar{Q}$ Output of the last F1ip-F1op (FC) is brought out as the "Input Register Full" output ( $\overline{\text { IRF }}$ ). After initialization this output is HIGH.


Figure 4-14. FIFO Conceptual Input Section

Paralle1 Entry. A HIGH level on the PL Input loads the D0-D3 Data Inputs into the F0-F3 Flip-F1ops and sets the FC Flip-F1op, which forces $\overline{\mathrm{IRF}}$ LOW, indicating "Input Register Full." The D inputs must be stable while PL is HIGH. During parallel entry, the $\overline{\text { IES }}$ Input should be LOW; the $\overline{\text { CPSI }}$ Input may be either HIGH or LOW.

After the fourth clock transition, the four serial data bits are aligned in the four data flip-flops and the FC Flip-Flop is set, forcing $\overline{I R F}$ LOW (Input Register full) and internally inhibiting further $\overline{\text { CPSI }}$ clock pulses. Figure 4-15 illustrates the final positions resulting from a $64-$ bit serial bit train. BO is the first bit, B63 the last bit.


Figure 4-15. Final Positions in a FIFO Resulting from a 64-Bit Serial Train
4.2.3.3 TRANSFER TO THE FALL-THROUGH STACK.

The outputs of F1ip-F1ops FO - F3 feed the Stack. A LOW leve1 on the $\overline{T T S}$ Input attempts to initiate a "fall-through" action. If the top location of the Stack is empty, data is loaded into the Stack and the input register is re-initialized. Note that this initialization is postponed until PL is LOW again. Thus, automatic FIFO action is achieved by connecting the $\overline{\text { IRF }}$ output to the $\overline{T T S}$ input.

Data falls through the Stack automatically, pausing only when it is necessary to wait for the next empty location. The $\overline{\mathrm{MR}}$ input only initializes the Stack control section and does not clear the data.

Serial Entry. Data on the DS Input is serially entered into the F3, F2, F1, F0, FC Shift Register on each HIGH-to-LOW transition of the CPSI Clock Input, provided $\overline{I E S}$ and PL are LOW.
4.2.3.4 OUTPUT REGISTER (DATA EXTRACTION).

The Output Register receives 4-bit data words from the bottom Stack location, stores it and outputs data on a 3-state 4-bit paralle1 data bus or on a 3-state serial data bus. The output section generates and receives the necessary status and control signals. Figure 4-16 is a conceptual logic diagram of the output section.


Figure 4-16. FIFO Conceptual Output Section

Paralle1 Data Extraction. When the FIFO is empty after a LOW pulse is applied to $\overline{M R}$, the Output Register Empty ( $\overline{O R E}$ ) Output is LOW. After data has been entered into the FIFO and has fallen through to the bottom Stack location, it is transferred into the output register, provided the "Transfer Out Parallel" (TOP) Input is HIGH , and the $\overline{\mathrm{OES}}$ Input is LOW. As a result of the data transfer, $\overline{O R E}$ goes HIGH indicating valid data on the data outputs (provided the 3-state buffer is enabled). TOP can now be used to clock out the next word. When TOP goes LOW, $\overline{O R E}$ goes LOW indicating that the output data has been extracted; but the data itself remains on the output bus until the next LOW-to-HIGH transition of TOP transfers the next word (if available) into the output register as explained above. During parallel data extraction, $\overline{\mathrm{TOS}}, \overline{\mathrm{CPSO}}$, and $\overline{\mathrm{OES}}$ should be LOW.

Serial Data Extraction. When the FIFO is empty after a LOW pulse is applied to $\overline{\mathrm{MR}}$, the Output Register Empty ( $\overline{\mathrm{ORE}}$ ) output is LOW. After data has been entered into the FIFO and has fallen through to the bottom Stack location, it is transferred into the output shift register provided the "Transfer Out Serial" ( $\overline{T O S}$ ) is LOW. TOP must be HIGH, and $\overline{O E S}$ and $\overline{\text { CPSO must }}$ be LOW. As a result of the data transfer, $\overline{\mathrm{ORE}}$ goes HIGH indicating valid data in the shift register. The 3-state serial Data Output QS is automatically enabled and puts the first data bit on the output bus. Data is serially shifted out on the HIGH-to-LOW transition of $\overline{\mathrm{CPSO}}$. The fourth transition empties the shift register, forces $\overline{O R E} L O W$ and disables the serial output OS. For serial operation, the $\overline{O R E}$ output may be tied to the $\overline{T O S}$ input, requesting a new word from the Stack as soon as the previous one has been shifted out.

### 4.2.3.5 HORIZONTAL EXPANSION.

The device may be horizontally expanded to store long words (in multiples of four bits) without external logic. The interconnections necessary to form a 16 -word by 12 -bit FIFO are shown in Figure 4-17. Using the same technique, any FIFO of 16 words by 4 X n bits can be constructed. When expanding in the hori-. zontal direction, it is usual to connect the $\overline{\mathrm{IRF}}$ and $\overline{\mathrm{ORE}}$ outputs of the right most device (most significant device) to the TTS and $\overline{T O S}$ inputs respectively of all devices to the left (less significant devices) to guarantee that no operation is initiated before all devices are ready.


Figure 4-17. FIFO Horizontal Expansion Scheme

FIFO Buffer Timing of various configurations and conditions is shown in Figures 4-18 through 4-25.


Conditions: Stack not full, $\overline{\text { IES }}$, PL LOW

Figure 4-18. FIFO Timing -
Serial Input, Unexpanded or Master Operation


Conditions: Stack not full, $\overline{\text { IES }}$ HIGH when initialized, PL LOW

Figure 4-19. FIFO Timing -
Serial Input, Expanded Slave Operation


Conditions: Data in stack, TOP HIGH, IES LOW when initialized, $\overline{O E S}$ LOW

Figure 4-20. FIFO Timing -
Serial Output, Unexpanded or Master Operation


Conditions: Data in stack, TOP HIGH $\overline{\text { IES }}$ HIGH when initialized

Figure 4-21. FIFO Timing Serial Output, Slave Operation


Conditions: $\overline{\text { IES }}$ LOW when initialized, $\overline{\mathrm{EO}}, \overline{\mathrm{CPSO}}$ LOW. Data available in stack

Figure 4-22. FIFO Timing - Parallel Output, 4-Bit Word or Master in Parallel Expansion


# Conditions: $\overline{T T S}$ connected to $\overline{\text { IRF }}, \overline{T O S}$ connected to $\overline{\text { ORE }}, \overline{\mathrm{IES}}, \overline{\mathrm{OES}}, \overline{\mathrm{EO}}, \overline{\mathrm{CPSO}}$ LOW, TOP HIGH 

Figure 4-23. FIFO Fall Through Time


Conditions: Stack not full, $\overline{\text { IES }}$ LOW when initialized

Figure 4-24. FIFO Timing - Parallel Load Mode, 4-Bit Word (Unexpanded) or Master in Parallel Expansion


Conditions: Stack not full, device initialized (Note 1) with $\overline{\text { IES }} \mathrm{HIGH}$

Figure 4-25. FIFO Timing - Paralle1 Load, Slave Mode
4.2.4 CYCLIC REDUNDANCY CHECK (CRC) GENERATOR/CHECKER (U23). Cyc1ic Redundancy Check (CRC) Generator/Checker is a 16-bit programmable device which operates on serial data streams and provides a means of detecting transmission errors. Cyclic encoding and decoding schemes for error detection are based on polynomial manipulation in modulo arithmetic. For encoding, the data stream (message polynomial) is divided by a selected polynomial. This division results in a remainder which is appended to the message as check bits. For error checking, the bit stream containing both data and check bits is divided by the same selected polynomial. If there are no detectable errors, this division results in a zero remainder. Although it is possible to choose many generating polynomials of a given degree, standards exist that specify a small number of useful polynomials. This device implements the polynomials listed in Table 4-14 by applying the appropriate logic levels to the select pins S0, S1 and S2. The CRC connections are shown in Figure 4-26.

The device consists of a 16 -bit register, a Read Only Memory (ROM) and associated control circuitry as shown in the Block Diagram. The polynomial control code presented at inputs SO, S1 and S2 is decoded by the ROM, selecting the desired polynomial by establishing shift mode operation on the register with Exclusive OR gates at appropriate inputs. To generate the check bits, the data stream is entered via the Data Inputs (D), using the HIGH to LOW transition of the Clock Input ( $\overline{\mathrm{CP}}$ ). This data is gated with the most significant Output ( $Q$ ) of the register, and controls the Exculsive OR gates (Figure 4-27). The Check Word Enable (CWE) must be held HIGH while the data is being entered. After the last data bit is entered, the CWE is brought LOW and the check bits are shifted out of the register and appended to the data bits using external gating (Figure 4-28).

To check an incoming message for errors, both the data and check bits are entered through the D Input with the CWE Input held high. - The device is not in the data path, but only monitors the message. The Error Output becomes valid after the last check bit has been entered into the device by a HIGH to LOW transition of $\overline{\mathrm{CP}}$. If no detectable errors have occurred during the data transmission, the resultant internal register bits are all LOW and the Error Output (ER) is LOW. If a detectable error has occurred, ER is HIGH; ER remains valid until the next HIGH to LOW transition of $\overline{\mathrm{CP}}$ or until the device has been preset or reset.

A HIGH on the Master Reset Input (MR) asynchronously clears the register. A LOW on the Preset Input ( $\overline{\mathrm{P}}$ ) asynchronously sets the entire register if the control code inputs specify a 16 -bit polynomial; in the case of 12 or 8 -bit check polynomials, on1y the most significant 12 or 8 register bits are set, and the remaining bits are cleared.


CONNECTION DIAGRAM DIP (TOP VIEW)


## NOTE:

The Flatpak version has the same pinouts (Connection Diagram) as the pinours IConnection

| PIN NAMES |  |
| :--- | :--- |
| $S_{O}-S_{2}$ | Polynomial Select Inputs |
| $\bar{D}$ | Data Input |
| $\overline{C P}$ | Clock (Operates on HIGH to |
|  | LOW Transition) Input |
| $\bar{P} W E$ | Check Word Enable Input |
| $M R$ | Preset (Active LOW) Input |
| $M$ | Master Reset (Active HIGH) Input |
| $\mathbf{Q}$ | Data Output |
| ER |  |



Figure 4-26. CRC Checker Connections

Table 4-14. CRC Polynomial Select Codes

| SELECT COde |  |  | POLYNOMIAL | REMARKS |
| :---: | :---: | :---: | :---: | :---: |
| $S_{2}$ | $\mathbf{S}_{1}$ | $\mathrm{s}_{0}$ |  |  |
| L | L | L | $x^{16+}+{ }^{15}+x^{2+1}$ | CRC-16 |
| L | L | H | $x^{16+x^{14}+x^{14}}$ | CRC-16 REVERSE |
| $L$ | H | L | $x^{16+}+x^{15}+x^{13}+x^{7}+x^{4}+x^{2}+x^{1+1}$ |  |
| L | H | H | $x^{12+} x^{11}+x^{3}+x^{2}+x+1$ | CRC-12 |
| H | L | L | $x^{8}+x^{7}+x^{5}+x^{4}+x+1$ |  |
| H | L | H | $\mathrm{x}^{8+1}$ | LRC-8 |
| H | H | $L$ | $x^{16+x^{12}+x^{5}+1}$ | CRC-CCITT |
| H | H | H | $x^{16+x^{11}+x^{4}+1}$ | CRC-CCITT REVERSE |



Figure 4-27. CRC Equivalent Circuit For $X^{16}+X^{15}+X^{2}+1$


NOTES:
Check word Enable is HIGH while data is being clocked. LOW during transmission of check bits. 9401 must be reset or preset before each compuration.
CRC check bits are generated and appended to data bits.

Figure 4-28. CRC Check Word Generation

```
4.2.5 16K (2K x 8) UV ERASABLE PROM (U21).
```

This device is a 16,384-bit ultraviolet erasable and electrically programmable read-only memory (EPROM).

## Pin Configuration



PIN NAMES


Figure 4-29. PROM Connections

### 4.2.5.1 ERASURE CHARACTERISTICS.

The erasure characteristics of this device are such that erasure begins to occur when exposed to light with wavelengths shorter than approximately 4000 Angstroms. It should be noted that sunlight and certain types of fluorescent lamps have wavelengths in the 3000-4000 $\begin{aligned} & \text { range. Data show that constant exposure to }\end{aligned}$ room level fluorescent lighting could erase the typical device in approximately 3 years, while it would take approximately 1 week to cause erasure when exposed to direct sun1ight. If the device is to be exposed to these types of lighting conditions for extended periods of time, opaque labels are available which should be placed over the window to prevent unintentional erasure.
4.2.5.2 DEVICE OPERATION.

The five modes of operation of the device are listed in Table 4-15. It should be noted that all inputs for the five modes are at TTL levels. The power supplies required are a +5 V VCC and a Vpp. The Vpp power supply must be at 25 V during the three programming modes, and must be at 5 V in the other two modes.

Table 4-15. PROM Mode Selection

| Pins <br> mOOE | CEmam ( 8 ) | $\begin{gathered} \delta \\ 180 \end{gathered}$ |  | $\begin{aligned} & \text { Vec } \\ & 180 \end{aligned}$ | $\begin{aligned} & \text { CuTHuTs } \\ & \text { Mis.121n } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Hesd | $v_{12}$ | $v_{16}$ | - 5 | . 5 | Dout |
| Stander | $v_{\text {IN }}$ | Oonit Cone | . 5 | 4 | Mish 2 |
| Propen | Pulred $V_{\text {IL }}$ io $V_{\text {IM }}$ | $V_{\text {M }}$ | +75 | 4 | Dim |
| Progrem Verior | $V_{\text {IL }}$ | $v_{1 L}$ | $\cdot 25$ | 4 | Dour |
| Progeominniot | $V_{12}$ | $V_{\text {IM }}$ | $\cdot 25$ | $\cdot 5$ | Mind 2 |

Read Mode. The device has two control functions, both of which must be logically satisfied in order to obtain data at the outputs. Chip Enable ( $\overline{\mathrm{CE}}$ ) is the power control and should be used for device selection. Output Enable ( $\overline{O E}$ ) is the output control and should be used to gate data to the output pins, independent of device selection. Assuming that addresses are stable, address access time (tACC) is equal to the delay from $\overline{\mathrm{CE}}$ to output (tCE). Data is available at the outputs 120 ns (tOE) after the falling edge of $\overline{\mathrm{OE}}$, assuming that $\overline{\mathrm{CE}}$ has been low and addresses have been stable for at least tACC - tOE.

Standby Mode. The device has a standby mode which reduces the active power dissipation by $75 \%$, from 525 mW to 132 mW . The device is placed in the standby mode by applying a TTL high signal to the $\overline{\mathrm{CE}}$ input. When in standby mode, the outputs are in a high impedence state, independent of the $\overline{\mathrm{OE}}$ input.

Output Deselection. The outputs of two or more devices may be OR-tied together on the same data bus. Only one should have its output selected ( $\overline{O E}$ low) to prevent data bus contention
between devices in this configuration. The outputs of the others should be deselected by raising the OE input to a TTL high level.

Programming. Initially, and after each erasure, all bits of the device are in the "1" state. Data is introduced by selectively programming "0's" into the desired bit locations. Although only "0's" will be programmed, both "1's" and "0's" can be presented in the data word. The only way to change a "0" to a "1" is by ultraviolet light erasure.

The device is in the programing mode when the Vpp power supply is at 25 V and $\overline{\mathrm{OE}}$ is at VIH. The data to be programmed is applied 8 bits in parallel to the data output pins. The levels required for the address and data inputs are TTL.

When the address and data are stable, a 50 msec , active high, TTL program pulse is applied to the $\overline{C E} / P G M$ input. A program pulse must be applied at each address location to be programmed. One can program any location at any time -- either individually, sequentially, or at random. The program pulse has a maximum width of 55 msec . The device must not be programmed with a DC signal applied to the $\overline{\mathrm{CE}} / \mathrm{PGM}$ input.
4.2.6 1024 X 4 BIT STATIC RANDOM ACCESS MEMORY (U5, U6).

This device is a 4096-bit static Random Access Memory organized as 1024 words by 4 bits and requires no clocks or refreshing to operate. The data is read out nondestructively and has the same polarity as the input data. Common input/output pins are provided and a separate chip select ( $\overline{\mathrm{CS}}$ ) lead allows easy selection of an individual package.


Figure 4-30. RAM Connections

### 4.2.7 TRI-STATE BUFFERS (U39).

These devices provide six, two-input buffers in each package. One of the two inputs to each buffer is used as a control line to gate the output onto the high-impedance state, while the other input passes the data through the buffer. The outputs are placed in the tri-state condition by applying a high logic level to the control pins.

Logic and Connection Diagram


| INPUTS |  | OUTPUT |
| :--- | :--- | :---: |
| $\bar{G}$ | $A$ | $Y$ |
| $H$ | $X$ | $H \cdot Z$ |
| $L$ | $H$ | $H$ |
| $L$ | $L$ | $L$ |

Figure 4-31. Tri-State Buffer Connections

### 4.2.8 BUS COMPARATOR (U37).

This device compares two binary words of two to six bits in length and indicates matching (bit-for-bit) of the two words. Inputs for one word are TTL inputs, whereas inputs of the second word are high impedance receivers driven by a terminated data bus. The output has a latch that is strobe controlled. The transfer of information to the output occurs when the $\overline{\text { STROBE }}$ input goes from a logical 1 to a logical 0 state. Inputs may be changed while the $\overline{S T R O B E}$ is at the logical 1 level, without affecting the state of the output.

Logic Diagram


Connection Diagram


Truth Table

-Latched in previous state

Figure 4-32. Bus Comparator Connections

|  | $\frac{\text { Min }}{}$ | $\frac{\text { Typ }}{}$ | $\frac{\text { Max }}{1.9}$ |
| :--- | :--- | :--- | :--- |
| Positive-going Threshold Voltage | 1.4 | 1.6 | 1.9 |
| Negative-going Threshold Voltage | 0.5 | 0.8 | 1.0 |

$$
\mathrm{Y}=\overline{\mathrm{A}}
$$



Figure 4-33. Schmitt Trigger Connections

### 4.2.10 D POSITIVE-EDGE-TRIGGERED FLIP-FLOPS WITH PRESET AND CLEAR (U2, U16, U25, U26, U36).

| INPUTS |  |  |  | OUTPUTS |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| PR | CLR | CLK | D | 0 | ¢ |
| L | H | $x$ | x | H | $L$ |
| H | L | $x$ | $x$ | $L$ | H |
| L | $L$ | $x$ | $\mathbf{x}$ | $\mathrm{H}^{\text {- }}$ | $\mathrm{H}^{\circ}$ |
| H | H | 1 | H | H | L |
| H | H | 1 | L | 1 | H |
| H | H | $L$ | $\times$ | 00 | $\overline{0} 0$ |



Figure 4-34. D F1ip-F1op Connections
4.2.11 TRI-STATE BUFFERS (U14, U38).

This device provides eight, two-input buffers in each package that employ low power Schottky TTL technology. One of the two inputs to each buffer is used as a control line to gate the output into the high impedance state, while the other input passes the data through the buffer. The outputs are placed in the tri-state condition by applying a high logic level to the enable pins.

Logic and Connection Diagram


Truth Table

| INPUTS |  | OUTPUT |
| :---: | :---: | :---: |
| $\mathbf{G}$ | $A$ | $Y$ |
| $H$ | $X$ | $Z$ |
| $L$ | $H$ | $H$ |
| $L$ | $I$ | $L$ |

Figure 4-35. Tri-State Buffer Connections
4.2.12 DECODER (U10, U17).

These are Schottky-clamped circuits designed for memory-decoding or data-routing applications requiring very short propagation delay times. This DIP decodes one of eight lines, based on the conditions at the three binary select inputs and the three enable inputs.

Connection and Logic Diagrams


Truth Table

| INPUTS |  |  |  |  | OUTPUTS |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENABLE |  | SELECT |  |  |  |  |  |  |  |  |  |  |
| G1 | G2* | C | 8 | A | Yo | V1 | V2 | V3 | Y4 | Y5 | Y6 | Y 7 |
| $\times$ | H | x | $x$ | $x$ | H | H | H | H | H | H | H | H |
| $L$ | $x$ | x | $x$ | x | H | H | H | H | H | H | H | H |
| H | $L$ | 1 | 1 | 2 | $L$ | H | H | H | H | H | H | H |
| H | 1 | 1 | 1. | H | H | $L$ | H | H | H | H | H | H |
| H | 1 | $L$ | H | $L$ | H | H | L | H | H | H | H | H |
| H | $L$ | $L$ | H | H | H | H | H | L | H | H | H | H |
| H | $L$ | H | 1 | $L$ | H | H | H | H | L | H | H | H |
| H | $L$ | H | 1 | H | H | H | H | H | H | L | H | H |
| H | $L$ | H | H | L | H | H | H | H | H | H | L | H |
| H | 6 | H | H | H | H | H | H | H | H | H | H | 1 |
| G2-G2A + G2B |  |  |  |  |  |  |  |  |  |  |  |  |
| $H=$ High level, $L$ = low level, $X=$ don't care |  |  |  |  |  |  |  |  |  |  |  |  |



Figure 4-36. Decoder Connections

### 4.2.13 DECODER/DEMULTIPLEXER (U22).

These Schottky-clamped circuits are designed to be used in high performance memory decoding or data routing applications requiring very short propagation delay times. This device contains two separate two-1ine to four-line decoders in one package. The active-low enable input can be used as a data line in demultiplexing applications. The device features fully buffered inputs, presenting only one normalized load to its driving circuit. All inputs are clamped with high performance Schottky diodes to suppress line-ringing.

Connection Diagram Logic Diagram


Figure 4-37. Decoder/Demultiplexer Connections
4.2.14 QUAD 2-LINE TO 1-LINE DATA SELECTOR/MULTIPLEXER (U9). This data selector/multiplexer contains inverters and drivers to supply full on-chip data selection to the four output gates. A separate strobe input is provided. A four-bit word is selected from one of two sources and is routed to the four outputs.

Connection Diagram


Logic Diagram


Truth Table

| INPUTS |  |  |  | OUTPUT Y |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| STROBE | SELECT | A | B | $\begin{array}{c}\text { I57 LIS7A } \\ \text { LS157. S157 }\end{array}$ | LS158 |  |
| S158 |  |  |  |  |  |  |$]$

$\mathrm{H}=$ High Level. L = Low Leval. X = Don't Care

Figure 4-38. Quad 2-Line to 1-Line Data Selector/Multiplexer Connections.

### 4.2.15 D-TYPE TRANSPARENT LATCHES (U42).

These 8-bit registers feature totem-pole three state outputs designed specifically for driving highly capacitare or relatively low impedance loads. The high impedance third state and increased high-logic-level drive provide these registers with the capability of being connected directly to and driving bus lines. Transparent operation means that while the enable (G) is high, the $Q$ outputs follow the data (D) inputs. When the enable is low, the output is latched at the level of the data that was set up. The output control daes not affect the internal operation of the latches. That is, the old data can be retained or new data can be entered even while the outputs are off.

Connection Diagram


Truth Table

| OUTPUT <br> CONTROL | ENABLE <br> G | D | OUTPUT |
| :---: | :---: | :---: | :---: |
| L | H | H | H |
| L | H | L | L |
| L | L | X | $Q_{0}$ |
| H | X | X | Z |

Logic Diagram


Figure 4-39. D-Type Transparent Latch Connections

### 4.2.16 DUAL 4-BIT BINARY COUNTER (U45).

These devices contain eight master-slave flip-flops and additional gating to implement two individual four bit counters in a single package, each with a clear and clock input. Parallel outputs are available from each counter so that any submultiple of the input count frequency is available.


Logic Diagram


Truth Table

| COUNT SEQUENCE <br> (EACH COUNTER) |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| COUNT | OUTPUT |  |  |  |
|  | 0 | ${ }^{0}$ | $\mathrm{O}_{8}$ | $0_{A}$ |
| 0 | L | L | L | L |
| 1 | L | L | L | H |
| 2 | L | $L$ | H | L |
| 3 | L | L | H | H |
| 4 | L | H | L | L |
| 5 | L | H | L | H |
| 6 | L | H | H | L |
| 7 | L | H | H | H |
| 8 | H | L | L | L |
| 9 | H | L | L | H |
| 10 | H | L | H | L |
| 11 | H | L | H | H |
| 12 | H | H | L | L |
| 13 | H | H | $L$ | H |
| 14 | H | H | H | L |
| 15 | H | H | H | H |

Figure 4-40. 4-Bit Binary Counter Connections
4.2.17 TRI-STATE D FLIP-FLOPS (U4, U11, U12, U13, U30, U31, U41). These 8 -bit registers contain D-Type flip-flops with totem-pole tri-state outputs capable of driving highly capacitive or low impedance loads. When the output control is taken to a high logic level, the outputs go into the high impedance state. When a low logic level is applied to the output control, data at the $D$ inputs are loaded into their respective flip-flops on the next positivegoing transition of the clock.

Connection Diagram


Truth Table

| OUTPUT <br> CONTROL | CLOCK | D | OUTPUT |
| :---: | :---: | :---: | :---: |
| L | I | H | H |
| L | t | L | L |
| L | L | X | OO |
| H | $X$ | $X$ | $Z$ |

Figure 4-41. Tri-State D F1ip-F1op Connections

## SECTION 5 <br> MAINTENANCE AND TROUBLESHOOTING

### 5.0 INTRODUCTION.

The AM-500 circuit board performs to full capability with a minimum of maintenance. This Section describes maintenance and troubleshooting procedures and procedures for handing warranty returns.
5.1 CIRCUIT BOARD CHECKOUT.

The AM-500 circuit board was fully tested before it left Alpha Microsystems and will operate satisfactorily in the system if the hardware and software requirements of Sections Two and Three of this manual are met. Should a problem arise after the circuit card has been in operation, use the following procedures to identify and locate the fault.

1. Check all cabling for proper seating of connectors.
2. Check the circuit board for proper seating in the slot.
3. Check all power connections for correct voltages.
4. Check jumper options to ensure correctness of application.
5. Verify that the fault is in the $\mathrm{AM}-500$ and not in the system or in the peripherals. This can best be accomplished with substitution of a known good circuit board.
6. Perform the diagnostic tests included with the circuit board as described in paragraph 5.3.

### 5.2 WARRANTY PROCEDURES.

This circuit board is covered by warranty issued by Alpha Microsystems Inc., Irvine, California. Complete details of the warranty are included with the circuit board. Should a problem arise with this circuit board, call your dealer or the Alpha Micro International Support Services Group for information.
5.3 TROUBLESHOOTING PROCEDURES.

Diagnostic testing software is included with the AM-500 circuit board and should be used for troubleshooting and to verify proper operation. Test description and operating procedures are contained in the following paragraphs.

### 5.3.1 TEST DESCRIPTION - EXERCISE MODE.

1. Test-A. This test formats all specified tracks of the disk under test and then Cyclic Redundancy Check (CRC) verifies that all sector addresses and codes are written correctly.
2. Test-B, Test-C, Test-D, Test-E. These tests write to all specified tracks a data word supplied by the user. They then read all sectors written to verify correct data. Sectors are written and read sequentially.
3. Test-F. This test writes random data to all specified tracks. Addressing is also random. Each sector will be written to only once during this test. After all appropriate sectors have been written, a verify operation is performed using the same random addressing scheme.
4. Test-G. This test writes to all specified tracks using an incrementing data pattern. Addressing
alternates between two counters; one counting down from the maximum specified sector address, and the other counting up from the minimum specified sector address. Only odd numbered sectors are written to when addressing from the counter counting down, and only even numbered sectors are written to when addressing from the counter counting up. When all appropriate sectors have been written, a read operation begins using the same data and addressing scheme.
5. Test-H. This test forces error conditions which should be detected by the drive. First, it tries to access an illegal sector address. Secondly, it tries to seek an illegal cylinder address. Both cases should cause detectable errors.

Test-H error reporting. An error is recorded when the proper error code is not returned by either the Hawk or the AM-500. For example: During the "illegal cylinder" portion of the test, a cylinder address of 500 is sent to the Hawk. This address is illegal and should cause the Hawk to send the appropriate error status back to the program. If the actual status does not indicate that an illegal cylinder address was received by the Hawk, the program records an error.

## NOTE

Before attempting this test, one switch setting must be changed on the Hawk. The switch to be changed is located on the $I / 0$ board, which is the PC board directly below the 50 pin ribbon cable coming from the AM-500. The switch itself (S4) is located in the lower left hand corner (see Figure 1). When the switch has been located, set position 3 to the "ON" position as described
in Figure 1. If this change is not made, the program will not detect illegal cylinder addresses. Ensure that the switch is returned to its proper setting when testing is done.


In the figure above, S4, Position 3, is shown in the "OFF" position. Set it to the "ON" position by depressing the switch on the left hand side. If a daisy chain configuration is being used, this switch setting must be made on each unit that is to be tested.

Figure 5-1. Disk Drive I/O Circuit Board Switch Location

### 5.3.2 TEST DESCRIPTION - ADJUSTMENT MODE.

1. Test-A. Performs one "RESTORE" operation on the disk under test. This causes the disk to reset its internal logic and seek cylinder 0 .
2. Test-B. Performs an alternate seek between any two user specified cylinders without performing a read operation.
3. Test-C. Seeks any legal cylinder head and enables the drive select and read gate.

### 5.3.3 TEST OPERATION.

5.3.3.1 RUNNING THE PROGRAM.

Load HWKTST.PRG into memory and run by typing "HWKTST" followed by a carriage return.
5.3.3.2 OPERATOR INPUTS.

The operator must specify a disk number (0-7) to indicate which disk will be subjected to tests.

The operator must specify the test mode (exercisor adjustment).

If the adjustment mode was selected, the user must select a test from the "menu" presented by the program. If either test $B$ or $C$ is chosen, the user will be prompted to enter additional data regarding the cylinders and head to be used.

If the exercise mode was selected, the user may define a sequence of tests to be performed. Up to 20 test characters may be entered followed by a carriage return. A letter may be repeated, which will cause the test to be repeated. The standard default is all tests (A,B,C,D,E,F,G G H) and is selected by entering a carriage return only.

Some tests (A,B,C,D,E $\& F)$ require additional operator inputs to fully define them. If any of these tests are selected, the operator will be prompted to enter the appropriate data. Default values will be used if the operator enters only a carriage return in response to these prompts. Below is a list of parameters and their standard default values:

| PARAMETER/TEST | DATA |
| :--- | :---: |
| Maximum Track | 807 |
| Minimum Track | 0 |
| Test-A | 0 |
| Test-B | 0 |
| Test-C | FFFF |
| Test-D | 0333 |
| Test-E | $39 C E$ |
| Test-F | FFFF |

5.3.3.3 USER OPTIONS.

The user may suppress all error outputs by answering yes (Y) to the first user option.

If errors are not to be suppressed, the user may include a dump of the AM-500 data buffer after each data miscompare.

The user may elect to dump an error summary after each test by answering yes (Y) to the appropriate prompt.

The user may elect to pause after each test. To continue, the user types a carriage return on the controlling terminal.

The user may elect to continuously loop through the entire test sequence, or only loop $X$ times as specified by the user.

### 5.3.3.4 ERROR HANDLING.

All disk errors will be printed as they occur unless the user has selected the option suppressing error printouts. The user may also select an option which will cause the AM-500 to dump its data buffer after every data miscompare. Additionally, error summaries may be output after each test or just a final error summary after all tests are complete.

### 5.3.3.5 ERROR TYPES.

Errors come under two types. The first type would be disk errors, such as seek, sector-not-found, and sentinel. These types of errors will cause up to four re-trys to occur before giving up. If the error persists for three or more re-trys, it is considered a hard error and is marked as such in the error summary. If the error corrects itself in less than three re-trys, it is marked as a soft error in the error summary.

The second type would be data errors. These errors are due to data miscompares during read operations. These errors are always marked as "hard" errors in the error summary.
5.3.3.6 TEST TERMINATION.

The test program may be aborted at any time by typing "CTRL-C."

### 5.3.3.7 BUFFER PRESETS.

In order to make disk read transfers apparent, tests $A, B, C, D$ and E fill the AM-500 buffer with complement data prior to each read operation. The buffer is not preset in other tests since data will never repeat between any two consecutive reads.
5.3.4 WORST CASE DATA PATTERNS.

The following data patterns generate worst case data patterns for the disk:
0 Produces highest amplitude, lowest frequency.This is the default pattern for tests $A$ and $B$.
FFFF Produces lowest amplitude, highest frequency.This is the default pattern for test $C$.
333
Worst case for peak shift. This is the default pattern for test $D$.
39CE Worst case for one-bit amplitude. This is the default pattern for test $E$.
5.3.5 DISK ERROR CODES.
ERROR
NUMBER DESCRIPTION
Disk not ready or fault exists.
Write protected.
Seek error.
Sector not found.
CRC error.
Sentinel field error.
NOTEAll other error codes are illegal at this time.If any other codes appear, the AM- 500 board mayhave problems.

### 5.3.6 RECOMMENDED TEST PROCEDURE.

The following sequence is recommended as a short but comprehensive test and will take about 7 minutes:

Type HWKTST followed by a RETURN.

Enter DISK NUMBER when prompted by program.

Select the EXERCISE MODE (\#1).

When prompted to enter test sequence ( $\mathrm{A}-\mathrm{H}$ ), enter a carriage return. This will cause a predetermined sequence of tests which consists of tests $A, B, C, D$, E,F,G,H in that order.

When asked for "HIGHEST TRACK," enter 220. When asked for "LOWEST TRACK," enter 200.

Enter a return for each of the following:
Format Data Word= Test B Data Word=
Test C Data Word= Test D Data Word=
Test E Data Word=

Enter a "Y" for "Suppress all error messages."

Enter "N" for "Dump data and CRC error tallies after each test."

Enter "N" for "Pause after each test."

Enter "N" for "Loop continuously?"

Enter 2 for number of passes.

This test sequence will use all of the default data values which were described in paragraph 5.3.5.

Upon completion of the tests, the program will give an ERROR REPORT. If any ERRORS exist, restart the program and use the same test sequence that was used originally, with one exception: when prompted with "SUPPRESS ALL ERROR MESSAGES?" enter "N." This will cause the program to output an error message each time an error is detected, indicating the test that was being executed when the error was detected.

## SECTION 6

SCHEMATIC AND PARTS LIST

Table 6-1. Component Crossreference List

| DESIG. | MFR. TYPE NO . | $\begin{aligned} & \text { PAR. } \\ & \text { NO. . } \end{aligned}$ | REF. <br> DESIG. | MFR. TYPE NO . | $\begin{aligned} & \text { PAR } \\ & \text { NO. } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| U1 | $\begin{aligned} & 10 \mathrm{Mhz} \\ & \text { Oscillator } \end{aligned}$ | $4.2 .10$ | U22 | 74LS139 | 4.2 .13 |
|  |  |  | U23 | 9401 | 4.2.4 |
| U2 | 74LS74 |  | U25 | 74LS 74 | 4.2 .10 |
| U3 | 74LS0 8 | $4.2 .10$ | U26 | 74LS 74 | 4.2 .10 |
| U4 | 74LS374 | 4.2 .17 | U27 | 74LS14 | 4.2 .9 |
| U5 | 2114 | 4.2 .6 |  |  |  |
| U6 | 2114 | 4.2 .6 | U29 | 74LS32 | - |
| U7 | 9403 | 4.2 .3 | U30 | 74 LS 374 | 4.2 .17 |
| U8 | 9403 | $4.2 .3$ | U31 | 74LS 374 | 4.2 .17 |
| U9 | 74LS157 | $4.2 .14$ | U32 | 74LS00 | - |
| U10 | 74LS138 | 4.2 .12 | U33 | 74LS0 8 | - |
| U11 | 74LS374 | 4.2 .17 | U34 | 74LS04 | - |
| U12 | 74LS374 | 4.2 .17 | U35 | 75451 | - |
| U13 | 74LS374 | 4.2 .17 | U36 | 74 LS 74 | 4.2.10 |
| U14 | 81LS9 7 | 4.2 .11 | U37 | 8131 | 4.2 .8 |
| U15 | 220/330 |  | U3 8 | 81LS9 7 | 4.2 .11 |
|  | Resistors | - | U39 | 8097 | 4.2 .7 |
| U16 | 74LS 74 | 4.2 .10 | U40 | 74LS30 | - |
| U17 | 74LS138 | 4.2 .12 | U41 | 74 LS 374 | 4.2 .17 |
| U18 | 74LS04 | - | U42 | 74 LS 373 | 4.2 .15 |
| U19 | 280 | 4.2 .1 | U4 3 | 75451 | - |
| U20 | 9517 | 4.2.2 | U44 | 74 LS 32 | $4.2 .16$ |
| U21 | $\begin{aligned} & 2716 \\ & C 28498 \end{aligned}$ | 4.2 .5 | U45 | 74LS39 3 |  |
|  |  | - |  |  |  |

*Paragraph number in Section 4 where a description of integrated circuit logic can be found.











| LPHA нicro SYSTEMS IRVINE, CA 92714 |  |  |
| :---: | :---: | :---: |
| Revistow | ${ }^{\text {TTHE }}$ SHEMATIC AM-500 <br> DISK CONTROLLER |  |
|  |  |  |
| DTKWN \|CHECKED | SFEET NO. | braming no. |
| Fprobved - | 9 | DWL-00500-XX |





| RETAIL PRICE $\$$ | 0.00 |  |
| :--- | :--- | :--- |
| WHSLE PRICE | $\$$ | 0.00 |
| NET PRICE | $\$$ | 0.00 |

DESCRIPTION DISK CONTROLLER AM-500
PART NUMBER DESCRIPTION QTY

| 1 | DWF-00500-00 | PCB DISK CONTROLLER AM-500 |  |
| :---: | :---: | :---: | :---: |
| 2 | CNS-00040-00 | SOCKET 40 PIN DIP |  |
| 3 | CNS-00024-00 | SOCKET 24 PIN DIP | 3 |
| 4 | CNS-00020-00 | SOCKET 20 PIN DIP | 10 |
| 5 | CNS-00018-00 | SOCKET 18 PIN DIP |  |
| 6 | CNS-00016-00 | SOCKET 16 PIN DIP | 6 |
| 7 | CNS-00014-00 | SOCKET 14 PIN DIP | 19 |
| 8 | CNS-00008-00 | SOCKET 8 PIN DIP |  |
| 9 | IC1-74367-00 | IC HEX BUFFER |  |
| 10 | CPP-00156-01 | CAPACITOR 15 UF 20V |  |
| 11. | CPN-00473-01 | CAPACITOR .047UF | 11 |
| 12. | HDM-00001-00 | HEATSINK 1.500 X 1.810 . 000 LG |  |
| 13 | ICL-00323-00 | IC REGULATOR +5V TO-3 |  |
| 14 | RS2-00101-00 | RESISTOR 100 OHM $\frac{1}{4} \mathrm{~W} 5 \%$ CAR |  |
| 15. | RS2-00222-00 | RESISTOR $2.2 \mathrm{~K} \frac{1}{4} \mathrm{~W}$ 5\% CAR |  |
| 16. | ICl-74393-01 | IC DUAL 4 BIT BINARY COUNTER |  |
| 17. | ICI-75451-00 | IC DUAL INTERFACE DRIVER NI OC |  |
| 18. | ICl-07430-01 | IC 8 INPUT NAND GATE |  |
| 19. | CNF-00002-11 | CONN HEADER 50PIN .092LG RT ANGLE |  |
| 20. | ICl-08197-01 | IC BUFFER OCTAL |  |
| 21. | ICl-74373-01 | IC OCTAL D FLIPFLOP |  |
| 22. | ICl-74374-01 | IC OCTAL D FLIPFLOP |  |
| 23. | ICl-74374-02 | IC OCTAL D FLIPFLOP | 4 |
| 24. | ICl-07474-01 | IC DUAL D FLIPFLOP |  |
| 25. | ICS-09403-00 | IC 8 BIT FIFO |  |
| 26. | ICI-74138-01 | IC DECODER 3 TO 8 LINE |  |
| 27. | ICl-07404-01 | IC HEX INVERTER |  |
| 28. | ICl-07408-01 | IC QUAD 2 INPUT AND GATE |  |
| 29. | ICM-02114-04 | RAM 256 X 4 BIT STATIC |  |
| 30. | ICl-74139-01 | IC DECODER 2 TO 4 LINE DUAL |  |
| 31. | ICl-08131-00 | IC COMPARATOR 6 BIT |  |
| 32. | ICl-74157-00 | IC QUAD 2 TO 1 DATA SELECTOR |  |
| 33. | ICS-09401-00 | IC GENERATOR CRC |  |
| 34. | ICl-07414-01 | IC HEX INVERTER W/HYSTERESIS |  |
| 35. | ICl-07400-01 | IC QUAD 2 INPUT NAND GATE |  |
| 36. | ICl-07432-01 | IC QUAD 2 INPUT OR GATE | 2 |
| 37. | ICS-00001-01 | IC OSCILLATOR 10MHZ |  |
| 38. | RSN-00003-00 | RESISTOR PACK 14PIN DIP 220/3300HM |  |
| 39. | ICS-09517-00 | IC CHIP DMA |  |
| 40. | ICS-00080-00 | IC MICROPORCESSOR Z-80 |  |
| 41. | HDS-00632-01 | SCREW 6-32 X . 375 | 2 |
| 42. | HDN-00632-01 | NUT HEX 6-32 STL SM PATTERN | 2 |
| 43. | HDW-00632-01 | WASHER LOCK 6-32 |  |
| 44. | ICl-07474-02 | IC DUAL D FLIPFLOP | 1 |
| 45. | ICM-00501-01 | ROM 2K X 8 BIT |  |

